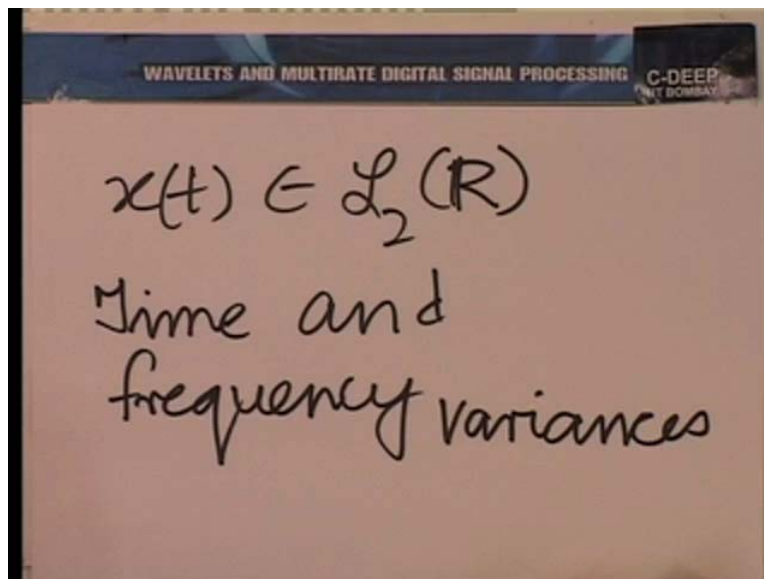


Advanced Digital Signal Processing
Wavelets and Multirate
Prof. V. M. Gadre
Department of Electrical Engineering
Indian Institute of Technology, Bombay

Lecture No # 18
Time Bandwidth Product Uncertainty Bound

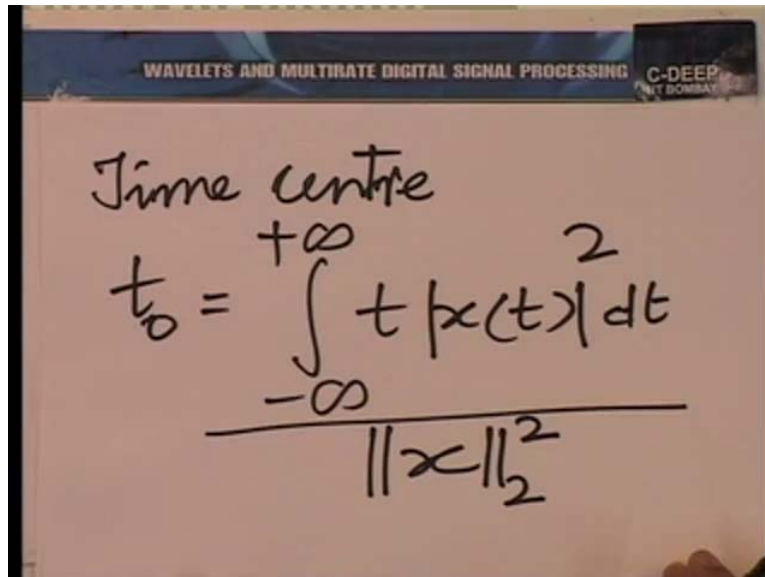
A very warm welcome to the eighteenth lecture, on the subject of wavelets and multirate digital signal processing. We build further in this lecture the idea of uncertainty and what is called the time band width product. The product of sigma t squared and sigma omega squared has been defined them in the last time. Now, we shall just recall a few ideas from the previous lecture to put our discussion in perspective. So, what we intend to talk about today is what is called the time bandwidth product, essentially as I said the product of the time variance and the frequency variance. And, we wish to build a lower bound on that product based on some very fundamental principles that could be explained when we look at functions as vectors. So, with that background, let us recall some of the definitions that we had brought out in the previous lecture.

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Last time, we had agreed to confine ourselves to a function $x(t)$ belonging to $L^2(\mathbb{R})$. And, we said we could define it as time and frequency variances. In fact, the time variance was first defined by choosing what is called the time centre.

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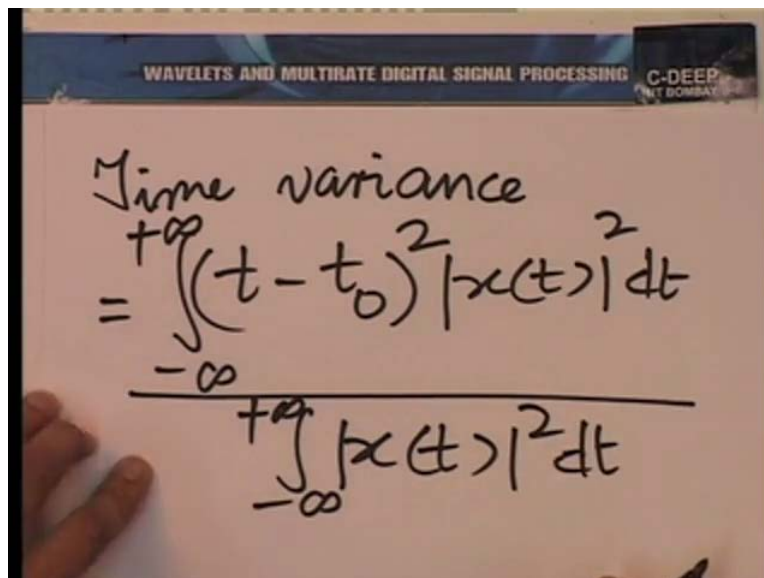


The image shows a whiteboard with the title "WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING" and "C-DEEP" in the top right corner. The handwritten text on the whiteboard is:

$$\text{Time centre}$$
$$t_0 = \frac{\int_{-\infty}^{+\infty} t |x(t)|^2 dt}{\|x\|_2^2}$$

The time centre t_0 is defined according to this. Essentially, treating $|x(t)|^2$ as a mass in terms of t and then looking for the centre of mass. That is the interpretation of this. And, we assumed t_0 the centre was a finite number. I mean, that is the reasonable assumption in almost all cases. That we will talk about.

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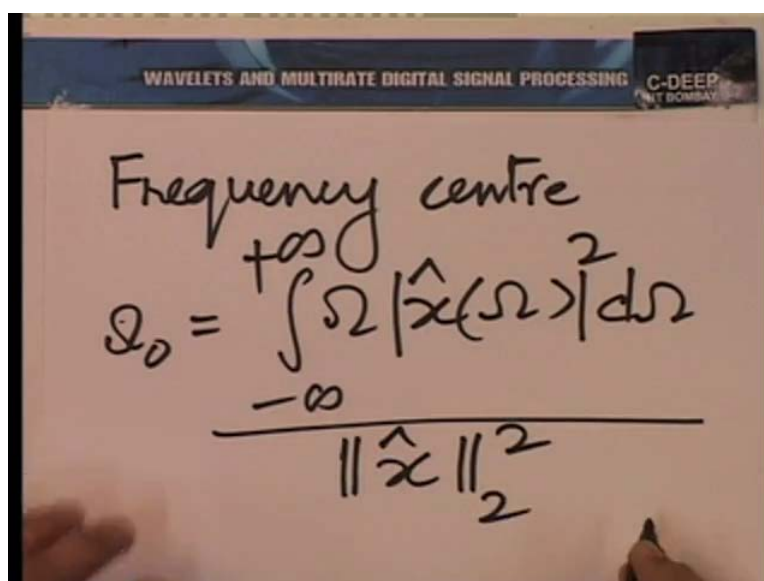


The image shows a handwritten formula for 'Time variance' on a presentation slide. The slide has a header that reads 'WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING' and 'C-DEEP IIT BOMBAY'. The formula is written in black ink on a light-colored background. It is as follows:

$$\text{Time variance} = \frac{\int_{-\infty}^{+\infty} (t - t_0)^2 |x(t)|^2 dt}{\int_{-\infty}^{+\infty} |x(t)|^2 dt}$$

From the time centre, we define the time variance. So, time variance is given by t minus t_0 the whole square mod x t square dt integrated over all time divided by the integral of mod x t squared. Essentially, this time variance was indeed the variance of a density, which we constructed out of that mass; mod x t squared divided by the norm of x the whole squared. Now, similarly we could talk about the frequency centre and the frequency variance.

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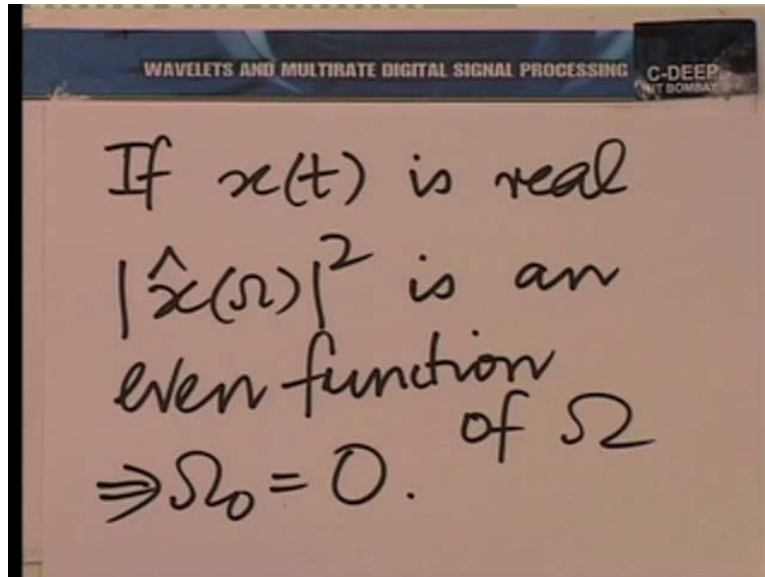


The image shows a handwritten formula for 'Frequency centre' on a presentation slide. The slide has a header that reads 'WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING' and 'C-DEEP IIT BOMBAY'. The formula is written in black ink on a light-colored background. It is as follows:

$$\Omega_0 = \frac{\int_{-\infty}^{+\infty} \Omega |\hat{x}(\Omega)|^2 d\Omega}{\|\hat{x}\|_2^2}$$

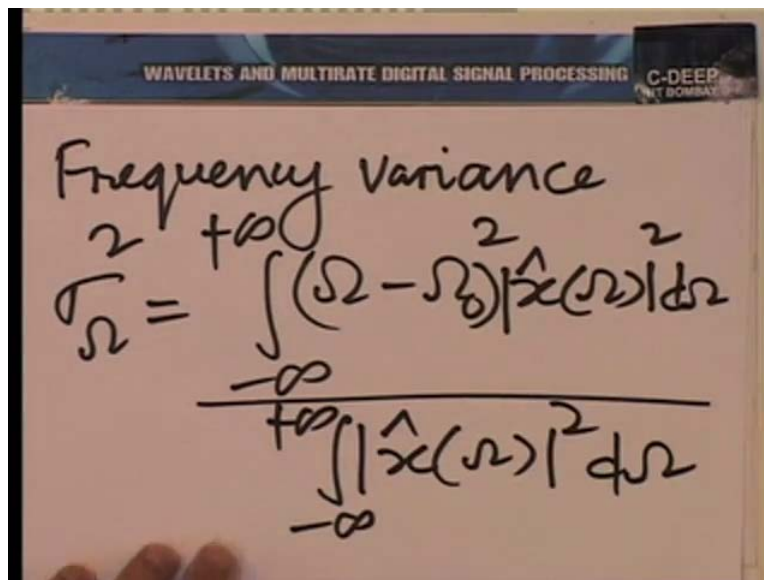
So, we defined the frequency centre to be capital omega 0 given by minus to plus infinity integral omega mod x cap omega the whole squared divided by the norm x cap squared. Again, we assume the frequency centre as finite, which is of course almost always the case. There is no problem and we also noted that for real x t, this frequency centre is 0.

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So, we said if x t is real, then mod x cap omega squared is an even function of omega. And, that implies that omega naught is 0 because of symmetry. Now, we could similarly define the frequency variance.

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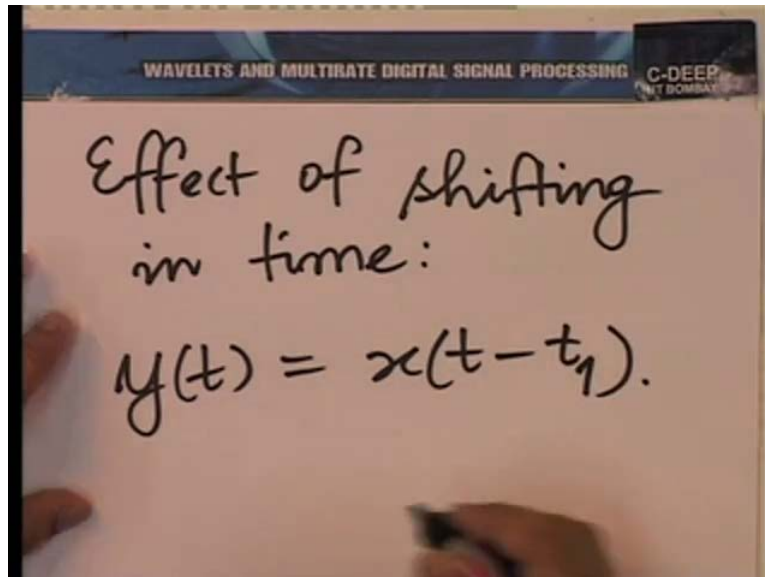
The image shows a handwritten formula on a presentation slide. The slide has a header that reads "WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING" and "C-DEEP" with "IIT BOMBAY" below it. The handwritten text "Frequency Variance" is at the top. The formula is:

$$\sigma_{\Omega}^2 = \frac{\int_{-\infty}^{+\infty} (\Omega - \Omega_0)^2 |\hat{x}(\Omega)|^2 d\Omega}{\int_{-\infty}^{+\infty} |\hat{x}(\Omega)|^2 d\Omega}$$

So, frequency variance sigma omega square; so, to speak, essentially integral omega minus omega naught squared modulus squared of x cap omega divided by a similar integral of mod x cap squared. So, once again the frequency variance is like a variance on the density constructed out of the modulus squared of the Fourier's transform.

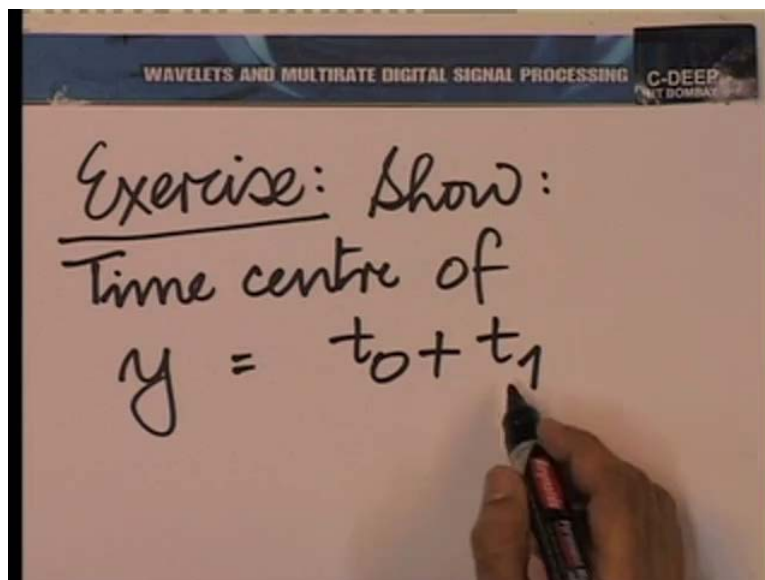
Now, let us see what happens when we shift a functioning time. As we expect, when we shift a function in time there is no change in the magnitude of the Fourier transform. In fact, all that happens is at the centre shifts.

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So, effect of shifting in time. So, let us consider y of t to be x of t minus t_1 . Now, we can easily write down the integrals corresponding to y of t . And, they are not very difficult to evaluate. I shall not go through all the details of working, but I shall straight away put down some of the important results here.

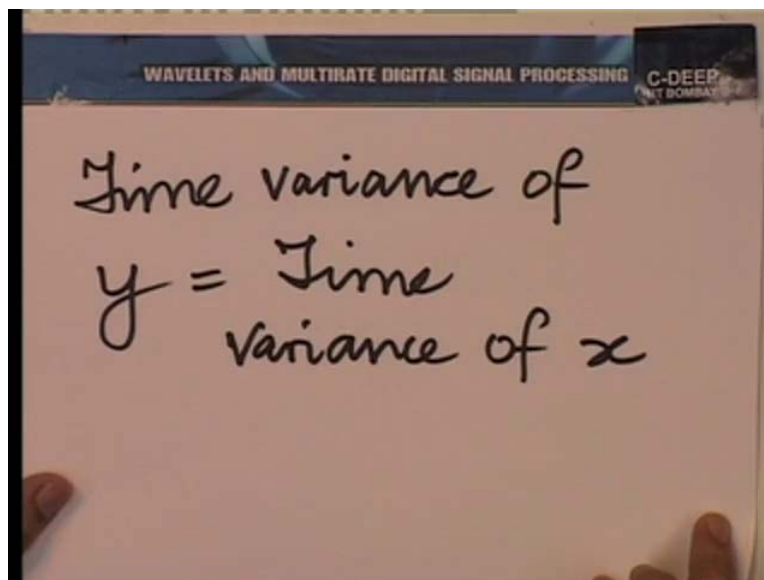
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In fact, I put this as an exercise. We can show that the time centre of y would essentially be... you see I take an example. Suppose, for example, you have a function centred at the point t equal to five. So, t_0 is equal to five. Now, if you shift the function backward by five units in time, you would get a function centred at zero. That is easy to visualise. And, if one just writes the integrals down carefully that is easy to prove.

So, essentially what it means is that, when you shift backwards by the centre, you are bringing it to be centred around zero. And, with that, we can generalise saying that the time centre of y ; essentially, t_0 plus t_1 . So, for example, if t_1 is equal to minus t_0 , then the centre of y would be zero. Anyway, so one can show the time centre of y 0.

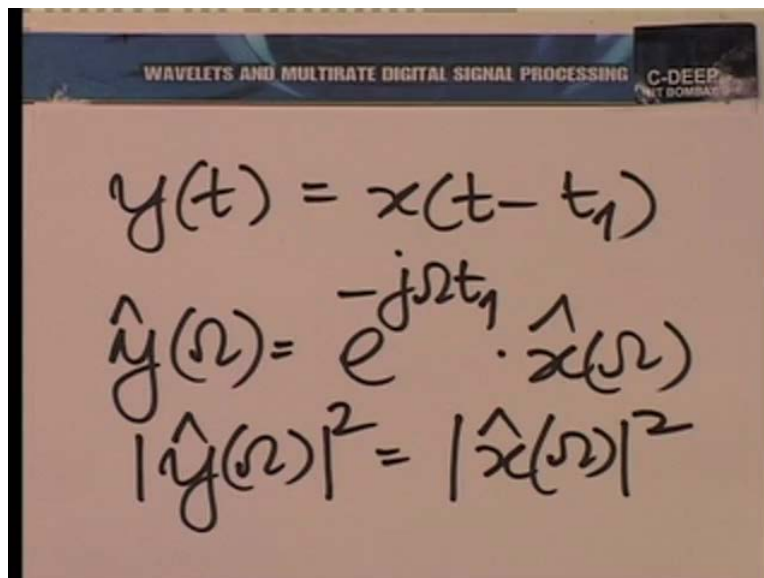
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The image shows a handwritten equation on a slide. The text is written in black ink on a light-colored background. The equation is:
$$\text{Time variance of } y = \text{Time variance of } x$$

The time variance of y is equal to the time variance of x . Again, this is a very easy result to show just amongst writing the integrals down carefully and making a transformation of variable. So, I would not go in to the full proof here. Similarly, we can now analyse what happens in frequency. What happens in the Fourier transform, let us look at that first.

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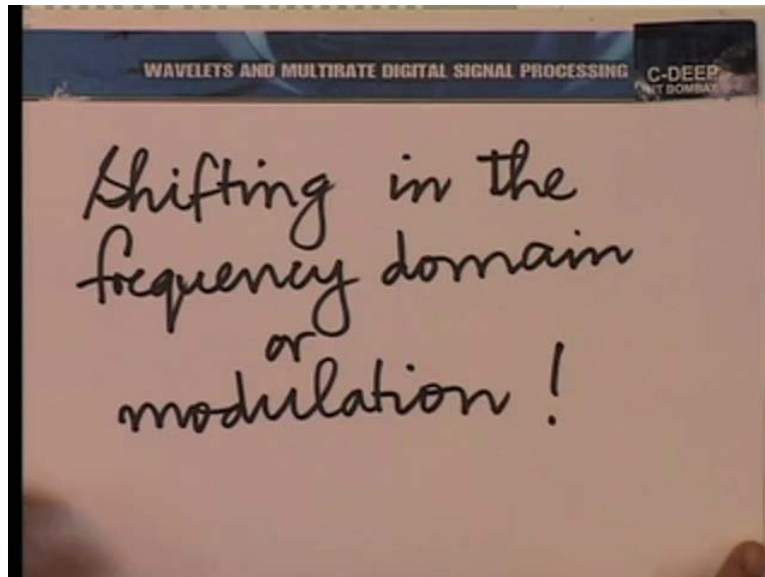
The image shows a slide with a header that reads "WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING" and "C-DEEP WIT BOMBAY". The slide contains three handwritten equations in black ink on a light brown background:

$$y(t) = x(t - t_1)$$
$$\hat{y}(\Omega) = e^{-j\Omega t_1} \cdot \hat{x}(\Omega)$$
$$|\hat{y}(\Omega)|^2 = |\hat{x}(\Omega)|^2$$

See, if y of t is equal to x of t minus t_1 , the Fourier transform y cap Ω is e raised to the power of minus j Ωt_1 times the Fourier transform of x . And, very clearly $\text{mod } y \text{ cap } \Omega$ the whole squared is equal to $\text{mod } x \text{ cap } \Omega$ the whole square. That is very easy to see here. So, you see under a shift in time, the magnitudes square, magnitude of the Fourier transform is unchanged, the mass **as** seen on the frequency axis remains unchanged in the Fourier domain.

So, you can shift as much as you desire and it has no effect on the Fourier transform magnitude squared. As you can see it effects the phase, but not the magnitude. And, the phase is not of consequence to us, as far as the centre and the variance are concerned. So, anyway with that, you understood what happens when we shift. Now, let us take the dual operation of shifting, namely modulation.

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So, in fact let us look at it, as I shift in the frequency domain. So, let us consider $y(t)$ to be a modulated version of $x(t)$.

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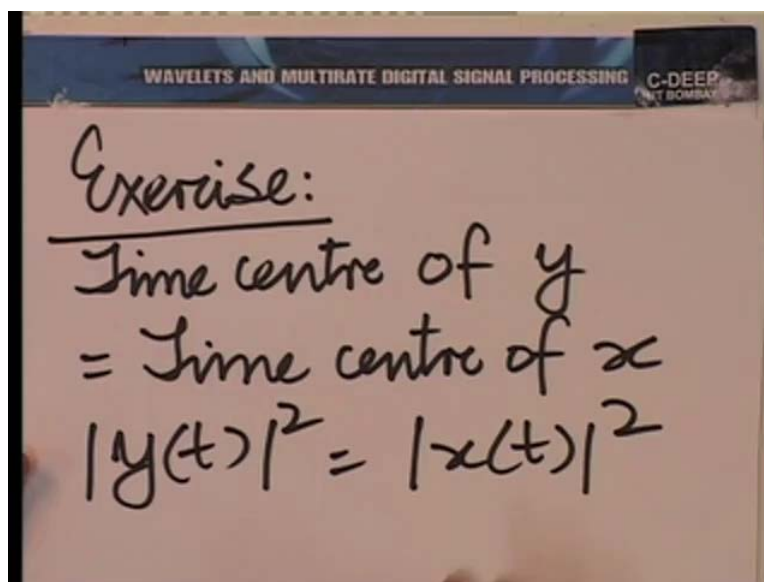
A photograph of a presentation slide. The slide has a dark blue header with the text 'WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING' and 'C-DEEP' on the right. The main content consists of two handwritten equations in black ink: $y(t) = e^{j\omega_1 t} x(t)$ and $\hat{y}(\omega) = \hat{x}(\omega - \omega_1)$.

So, let us consider $y(t)$ equal to $e^{j\omega_1 t}$ times $x(t)$. Now, of course it is very easy to see the Fourier transform, $\hat{y}(\omega)$ is $\hat{x}(\omega - \omega_1)$. You will recall

that this idea is used in amplitude modulation or for that matter in frequency modulation in Communication Engineering.

When we shift a signal on the frequency axis, we are modulating in time. Now of course here we are talking about modulation by a complex exponential. If we modulate by a sine wave, we are essentially modulating by two complex exponentials and adding up these two modulated signals. Just, to put the idea of modulation in perspective. Anyway, coming back to this; so, when we modulate with $e^{j\omega_1 t}$, we essentially shift on the ω axis as we have done here. So, y cap ω is x cap ω minus ω_1 .

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And, of course it is again easy to work out the following exercise. The time centre of y is equal to the time centre of x . In fact, that is very easy to see mod y t squared is equal to mod x t squared. And, first as far as that time domain is concerned, as far as the variable t is concerned, there is no change in the mass that is placed on t . So, it is not surprising that the time centre is unchanged and also the time variance, which we will now show.

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP MIT BOMBAY

$$\text{Time variance of } y = \text{Time variance of } x!$$

The time variance of y is equal to the time variance of x . What about the frequency variance and the frequency centre? That goes exactly dual to what happened, when we shifted in time.

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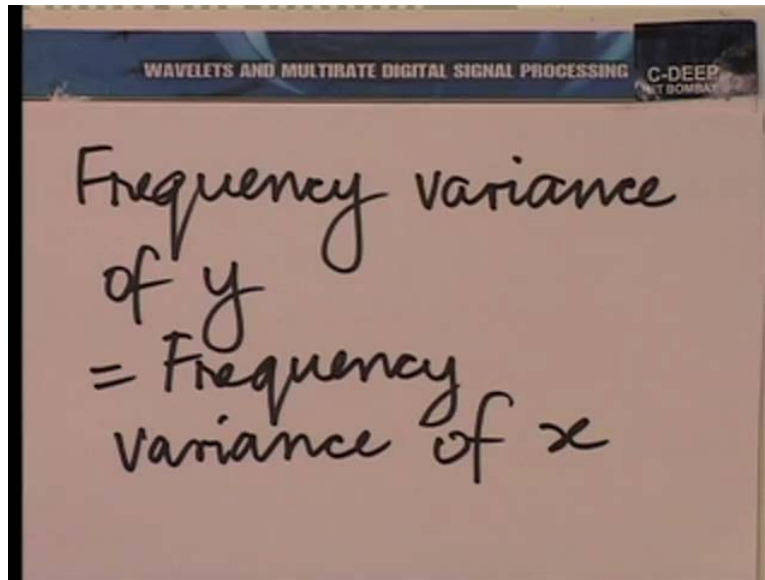
WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP MIT BOMBAY

$$\text{Frequency centre of } y = (\text{Frequency centre of } x) + \Omega_1$$

So, frequency centre, the frequency centre of y is now going to be the frequency centre of x plus ω_1 . And, put a bracket there. And, what happens to the frequency variance? While shifting

on the frequency axis, has no effect on the variance. Again, I leave it to you to show this by writing down the integral carefully. It is a simple exercise. I shall put down the final result.

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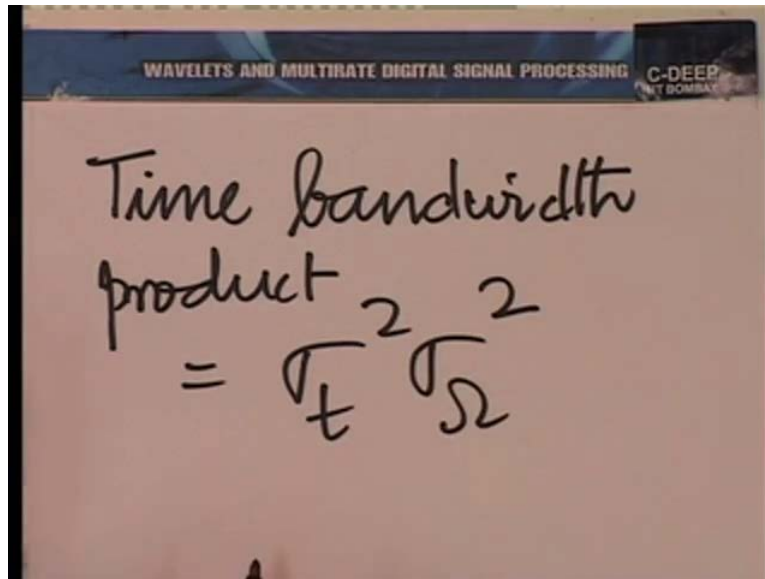


The image shows a handwritten equation on a slide. The slide has a header that reads "WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING" and "C-DEEP" on the right. The handwritten text in black ink says: "Frequency variance of y = Frequency variance of x".

$$\text{Frequency variance of } y = \text{Frequency variance of } x$$

So, the frequency variance of y is equal to the frequency variance of x . So far, so good. So, we have taken two dual operations; the time shift and the frequency shift. And, both of them we have had no change in the variances. And, obviously when there is no change in the variances, there is going to be no change in the product of the variances. So, the $\sigma_t^2 \sigma_\omega^2$ product, which we shall now give a name; we call it as the time bandwidth product.

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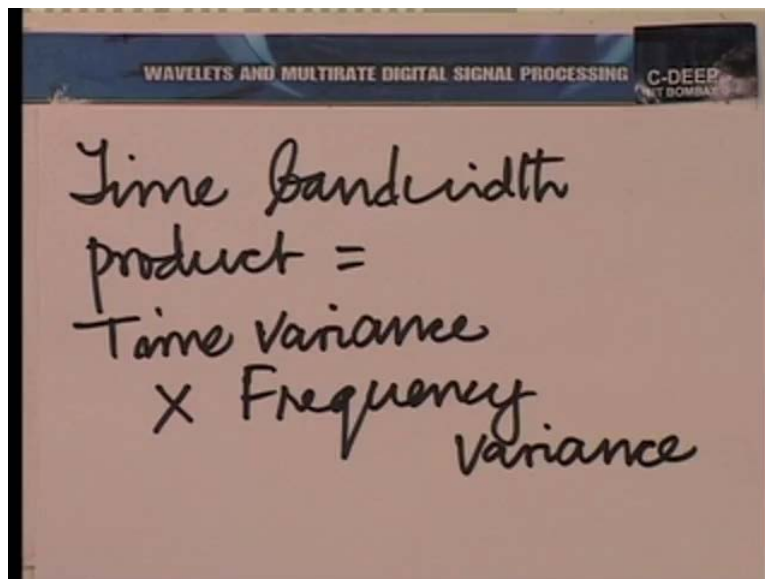


A photograph of a whiteboard with a blue header that reads "WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING" and "C-DEEP" on the right. The whiteboard contains the handwritten text "Time bandwidth product" followed by the equation $= \sigma_t^2 \sigma_\Omega^2$.

$$\text{Time bandwidth product} = \sigma_t^2 \sigma_\Omega^2$$

So, we introduce the term for time bandwidth product, given by the product sigma t squared times sigma omega squared.

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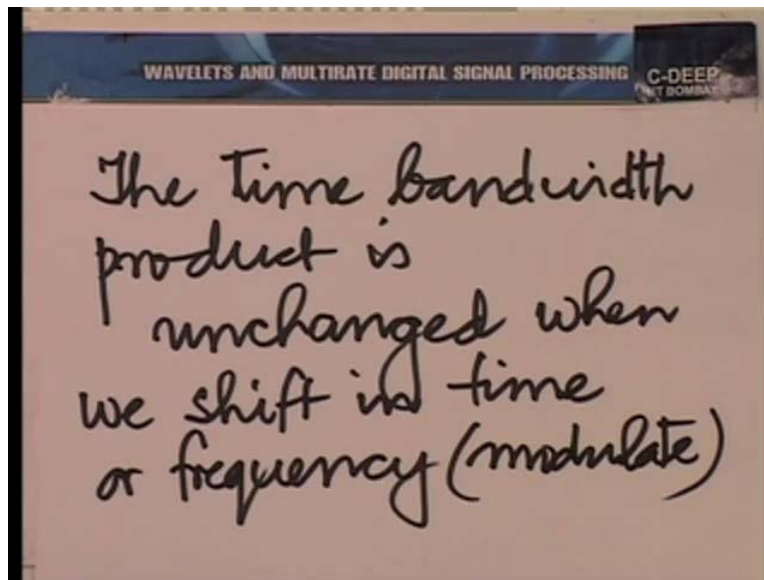
A photograph of a whiteboard with a blue header that reads "WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING" and "C-DEEP" on the right. The whiteboard contains the handwritten text "Time bandwidth product =" followed by "Time variance" and "X Frequency variance" on separate lines.

$$\text{Time bandwidth product} = \text{Time variance} \times \text{Frequency variance}$$

Essentially, the time bandwidth product is the product of the time variance and the frequency variance. And, this product is very important from a number of perspectives. As you can see,

there is a strong invariance that this product exhibits. When you shift a function in time or when you shift it in frequency; in other words, modulated in time, this product is unchanged.

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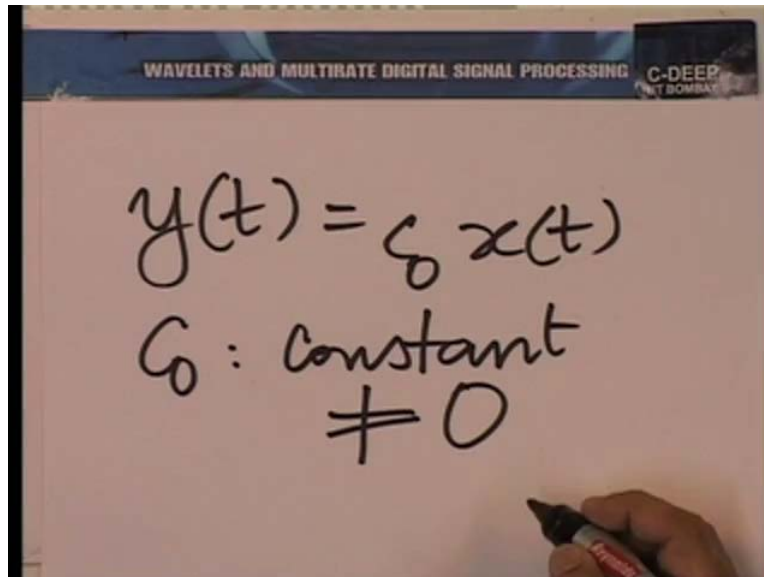
Let us make a note of that. The time bandwidth product is unchanged, when we shift in time or frequency. And remember, shifting in frequency is modulation in time. Now, we **would** like to look at this product in a little greater **depth**. What happens to this product when we stretch or compress. You see in the wavelet transform **or** for that matter in all the discussion that we had in some of the previous lectures, we have talked about moving along the frequency axis by a process of dilation.

We saw that, when **we stretch the compressed function**, the band pass function, it is equivalent to moving that function on the frequency axis; moving in a constant quality factor fashion keeping the ratio of the centred frequency to the bandwidth **a** constant. Now, in this process of compression or expansion together called dilation, what is happening to this fundamental quantity the time bandwidth product; is the question that we would now like to ask and answer.

So, the first thing that we observe is that you know, if you look at these let us... let us look at that process. So, suppose you have a function, you multiplied by a construct; that is a trivial thing. We will settle that matter first. So, we had, see here, we are talking about scaling the

independent variable. When we stretch or compress, we are talking about scaling the independent variable. What happens when you scale the dependent variable? So, if you multiply the function by a constant. You must answer that question first.

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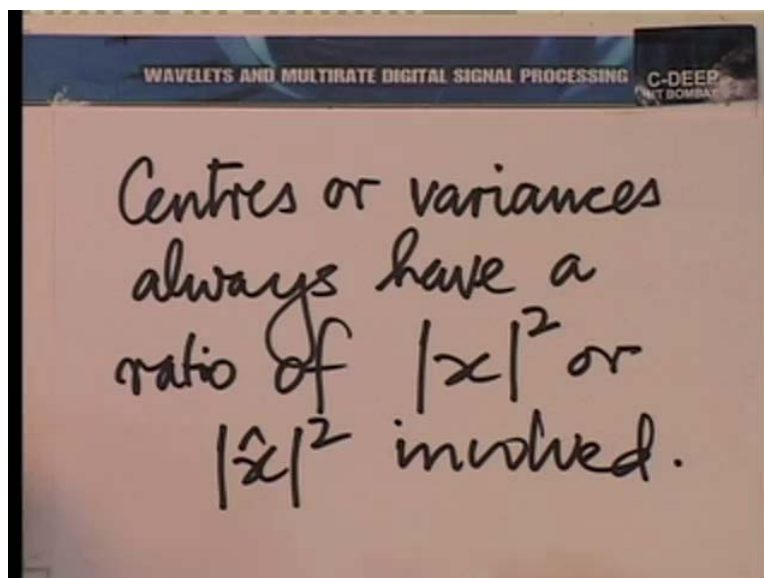
A photograph of a whiteboard with handwritten text. The top of the whiteboard has a blue header with the text "WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING" and "C-DEEP" on the right. The main content is handwritten in black marker. It shows the equation $y(t) = c_0 x(t)$ followed by the text " c_0 : constant" and " $\neq 0$ ". A hand holding a black marker is visible at the bottom right of the whiteboard.

$$y(t) = c_0 x(t)$$

c_0 : constant
 $\neq 0$

So, again it is very easy to show that if $y(t)$ is equal to some constant, let us say C_0 times $x(t)$, C_0 is the constant. Of course, unequal to zero; that is straight forward. So, C_0 is a constant.

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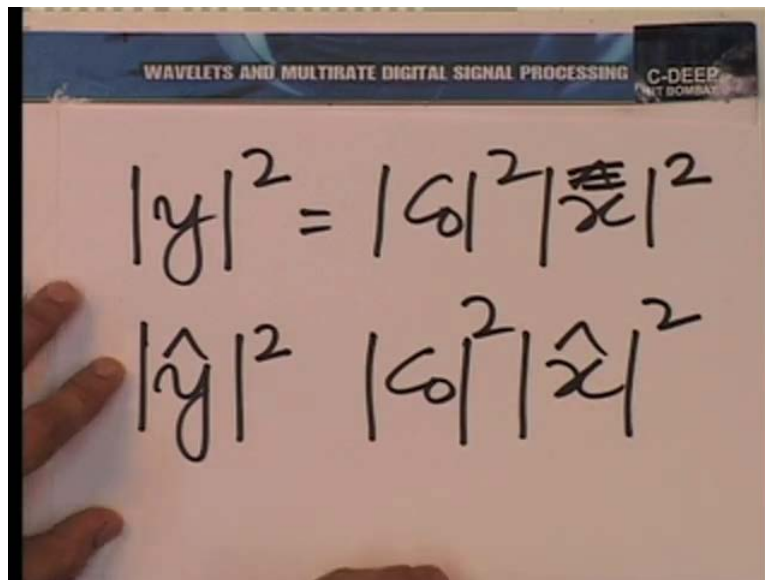


A photograph of a whiteboard with handwritten text. The top of the whiteboard has a blue header with the text "WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING" and "C-DEEP" on the right. The main content is handwritten in black marker. It says "Centres or variances" followed by "always have a" followed by "ratio of $|x|^2$ or" followed by " $|\hat{x}|^2$ involved."

Centres or variances
always have a
ratio of $|x|^2$ or
 $|\hat{x}|^2$ involved.

It is very easy to see that in the process of defining the centre or the variance, the centres or variance always have a ratio of mod x squared or mod x cap squared, involved.

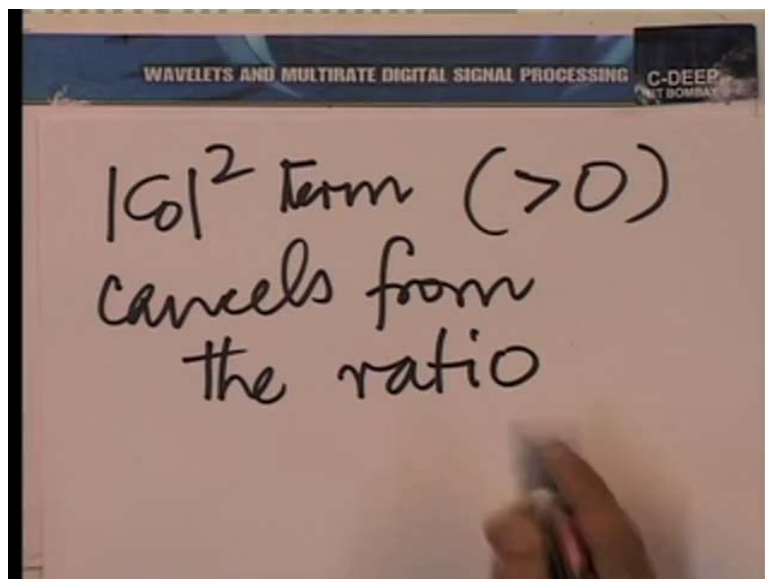
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The image shows a whiteboard with handwritten mathematical expressions. At the top, there is a header that reads "WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING" and "C-DEEP MIT BOMBAY". The main content consists of two equations. The first equation is $|y|^2 = |C_0|^2 |\hat{x}|^2$. The second equation is $|\hat{y}|^2 = |C_0|^2 |\hat{x}|^2$. A hand is visible on the left side of the whiteboard, pointing towards the equations.

Now, you know, if y is C_0 times x then $|y|^2$ is $|C_0|^2$ times $|x|^2$ or for that matter, $|y|^2$ is going to be $|C_0|^2$ times $|x|^2$. I am sorry, here it is just x .

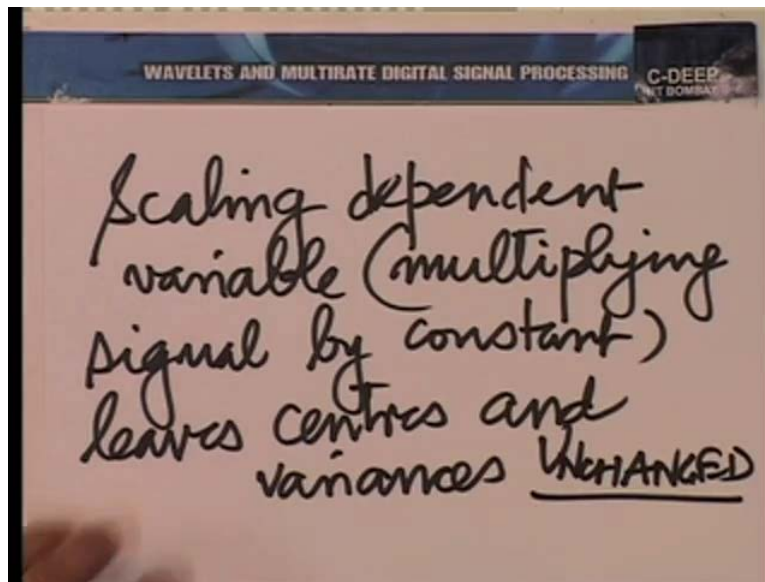
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The image shows a whiteboard with handwritten text. At the top, there is a header that reads "WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING" and "C-DEEP MIT BOMBAY". The main content consists of the text " $|C_0|^2$ term (> 0)" followed by "cancels from the ratio". A hand is visible at the bottom right of the whiteboard, holding a pen.

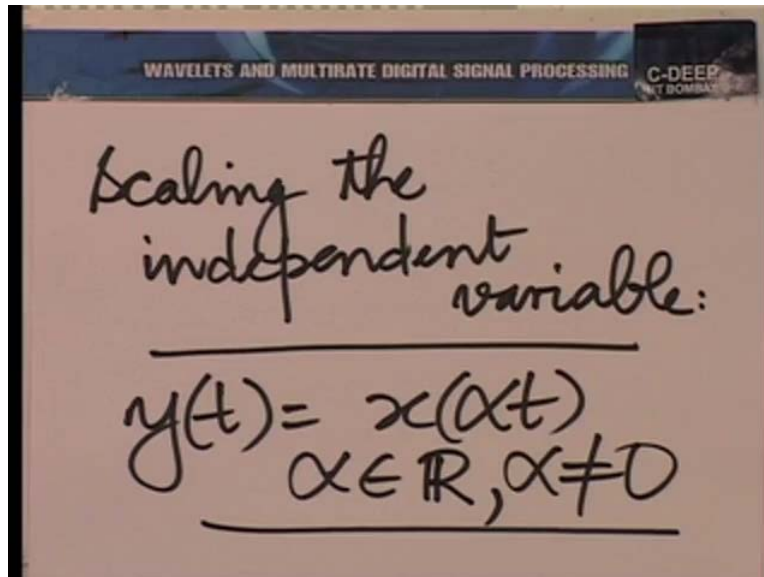
So, here to its mod $C \neq 0$ squared mod x cap squared. Now, when we take a ratio, this mod $C \neq 0$ squared being non-zero is going to cancel from the numerator and denominator. And therefore, when we scale the dependent variable there is no effect on the time centre or the frequency centre. Let us make a note of that. The mod $C \neq 0$ square term, of course greater than 0 strictly, cancels from the ratio.

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And therefore, scaling the dependent variable or essentially multiplication by a constant, multiplying the signal by a constant leaves the centres and variances unchanged. This is important. Scaling the dependent variable leaves the centres and the variances unchanged. And, now the most difficult of **them, all in** yet elegant and beautiful. What happens when we scale the independent variable?

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP
M T BOMBAY

Scaling the independent variable:

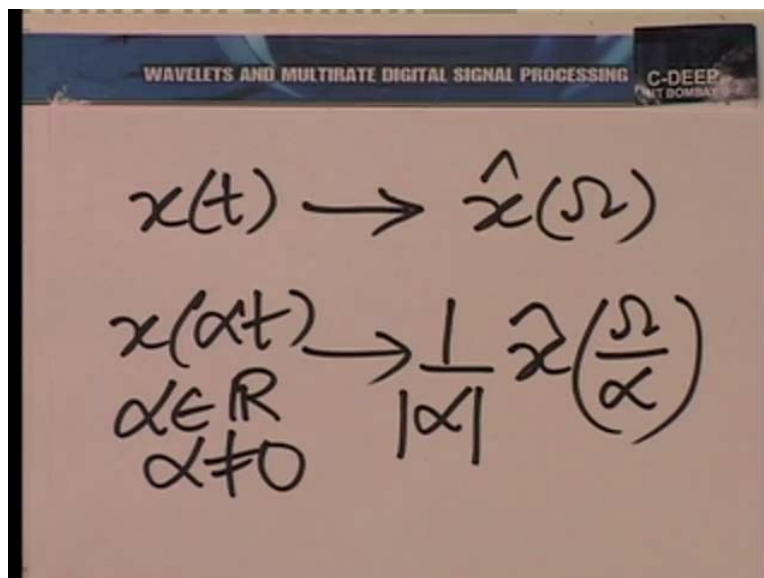
$$y(t) = x(\alpha t)$$

$\alpha \in \mathbb{R}, \alpha \neq 0$

In other words, let us consider $y(t)$ equal to $x(\alpha t)$ and, α is the real number. α is not equal to 0. Now, we know what happens to the Fourier transform in this.

So, you will recall that when we scale the t variable by α , the Fourier variable ω is scaled by $1/\alpha$, but $|1/\alpha|$ also one by $|\alpha|$ factor that emerges outside.

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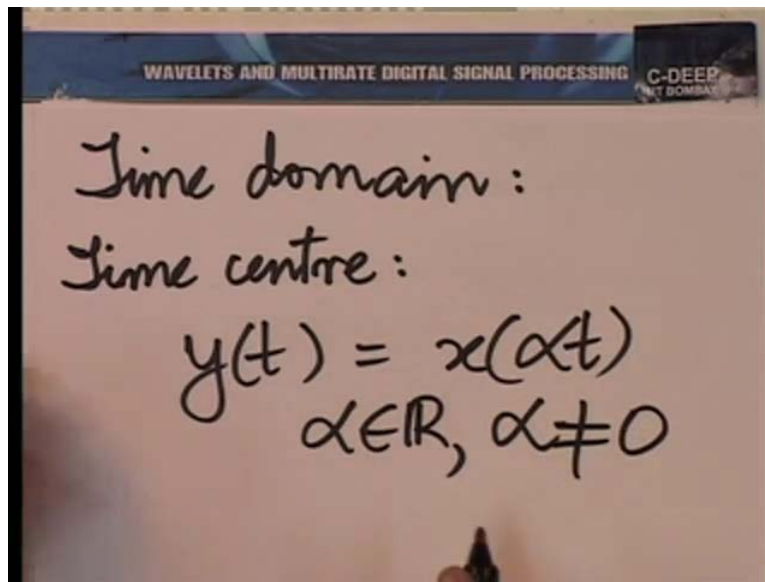
WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP
M T BOMBAY

$$x(t) \rightarrow \hat{x}(\omega)$$
$$x(\alpha t) \rightarrow \frac{1}{|\alpha|} \hat{x}\left(\frac{\omega}{\alpha}\right)$$

$\alpha \in \mathbb{R}$
 $\alpha \neq 0$

So, what we saying is, if $x(t)$ has the Fourier transform $X(\omega)$ and $x(\alpha t)$ with α real number, $\alpha \neq 0$ has the Fourier transform $\frac{1}{|\alpha|} X(\frac{\omega}{\alpha})$. Here, it is α ; here, it is $\frac{\omega}{\alpha}$. So, in fact it is adequate for us to see what happens to one of them, the time or the frequency domain. And, the other one can be interpreted. Let us take the time domain for simplicity.

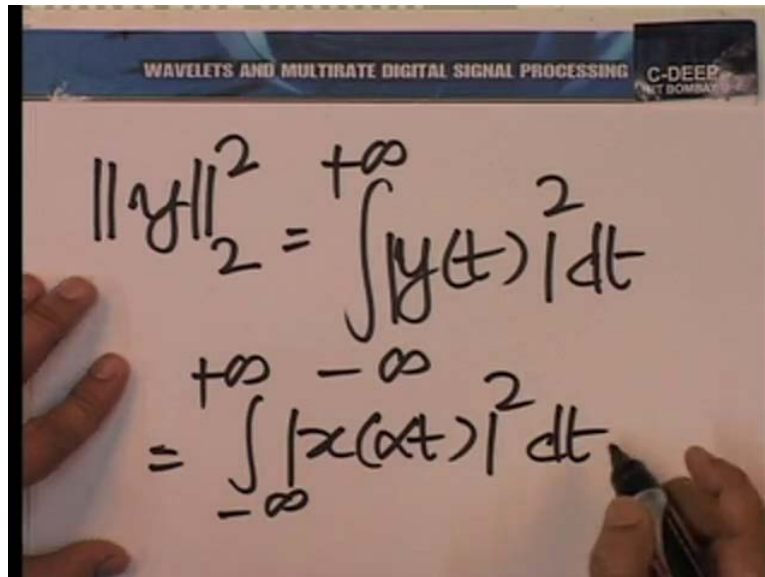
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Time domain:
Time centre:
$$y(t) = x(\alpha t)$$
$$\alpha \in \mathbb{R}, \alpha \neq 0$$

So, consider the time domain. Let us see what happens to the time centre. Let $y(t)$ be equal to $x(\alpha t)$; as usual α real, $\alpha \neq 0$. And, let us ask **what is the norm of $y(t)$, norm squared of $y(t)$ in the $L^2(\mathbb{R})$ sense against the norm of x .**

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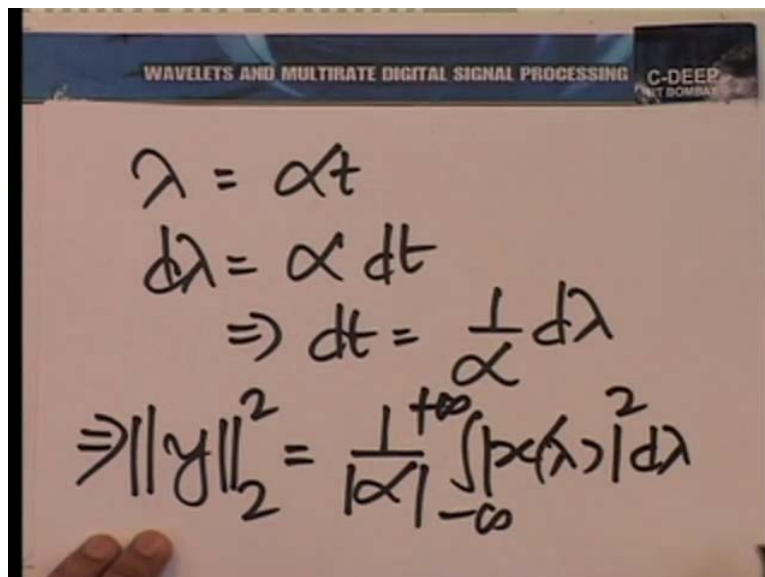


WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP
M T BOMBAY

$$\begin{aligned}\|y\|_2^2 &= \int_{-\infty}^{+\infty} |y(t)|^2 dt \\ &= \int_{-\infty}^{+\infty} |x(\alpha t)|^2 dt\end{aligned}$$

So, the L 2 norm of y is this. L 2 norm squared I mean and of course, this is easily seen to be mod x alpha t squared d t. Now, it is an easy integral to evaluate. All that we need to do is to make a transformation of variable.

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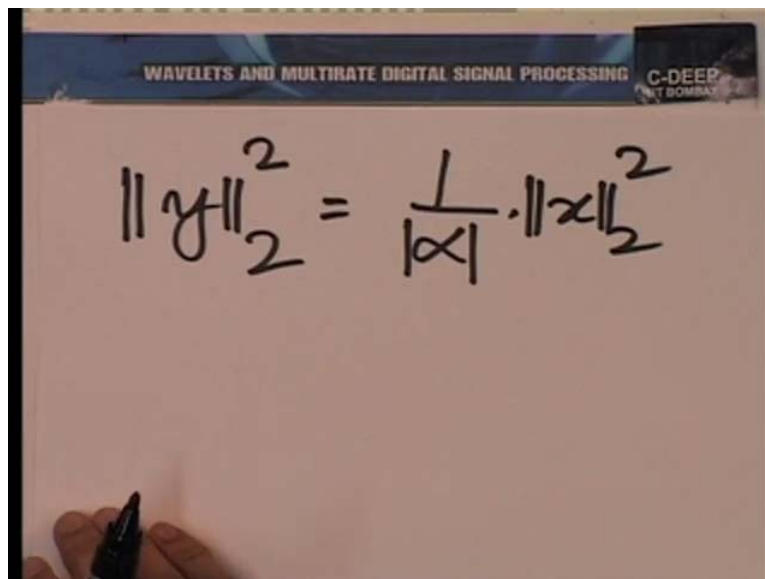
WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP
M T BOMBAY

$$\begin{aligned}\lambda &= \alpha t \\ d\lambda &= \alpha dt \\ \Rightarrow dt &= \frac{1}{\alpha} d\lambda \\ \Rightarrow \|y\|_2^2 &= \frac{1}{|\alpha|} \int_{-\infty}^{+\infty} |x(\lambda)|^2 d\lambda\end{aligned}$$

So, we have $\lambda = \alpha t$, whereupon $d\lambda = \alpha dt$. And therefore, dt is $1/\alpha$ by $d\lambda$. Now, here in the integral you see, let me just bring the integral back here, the relevant integral.

So, we are going to replace dt by $d\lambda/\alpha$. And, depending upon whether α is positive or negative. If α is positive, the limits still remain from minus to plus infinity. If α is negative, the limits go from plus infinity to minus infinity, but then there is also $1/\alpha$ negative there. So, $1/\alpha$ is negative. And therefore, if you take the negative and the reverse integral together, all in all we always have the following. The norm of y in the L^2 sense is always $1/|\alpha|$ times integral from minus to plus infinity $|x|^2 d\lambda$.

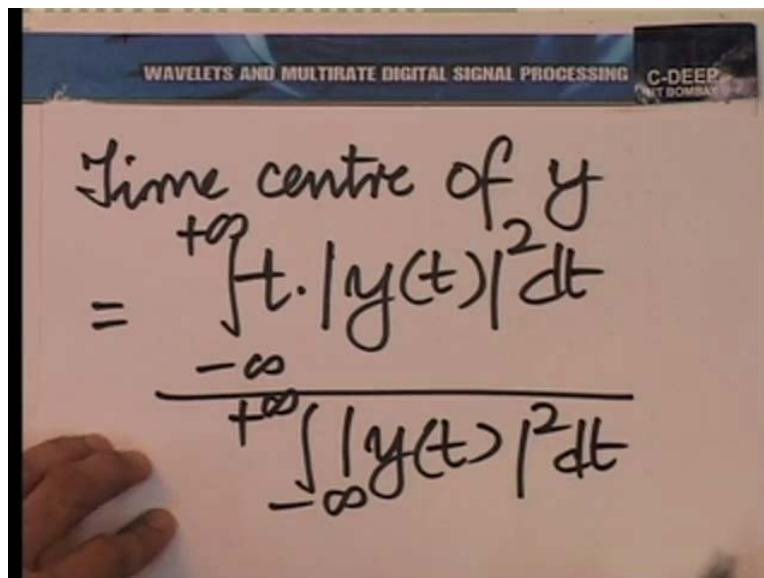
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$$\|y\|_2^2 = \frac{1}{|\alpha|} \cdot \|x\|_2^2$$

Whereupon, what we have concluded is norm y in the L^2 sense squared is $1/|\alpha|$ times the norm of x in the L^2 sense squared. Now, of course we can use the same reasoning. I have illustrated the **thought** behind the reasoning. And, if we write down the time centre and the frequency centre expressions, let us write down the time centre. . As we said we are going to focus on time.

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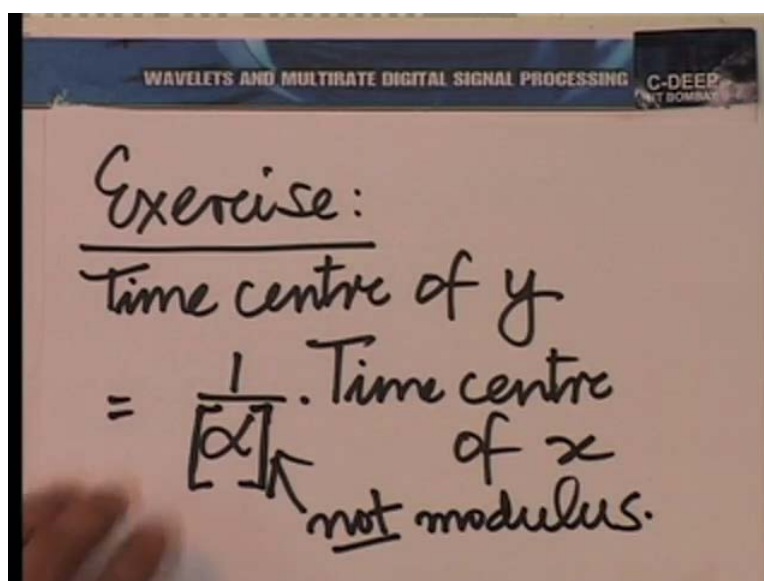


WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP
M T BOMBAY

$$\text{Time centre of } y = \frac{\int_{-\infty}^{+\infty} t \cdot |y(t)|^2 dt}{\int_{-\infty}^{+\infty} |y(t)|^2 dt}$$

So, let us write down the time centre expressions. The time centre of y is going to be given by t times mod y t squared dt **integrated**, divided by integral mod y t square dt . Now, again I shall not repeat all the working that I have done to relate the norms of y and x . Essentially, the central idea there is a transformation of variable. Put λ equal to αt and do this all throughout the integral. And, one can straight away write down the final result.

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP
M T BOMBAY

Exercise:

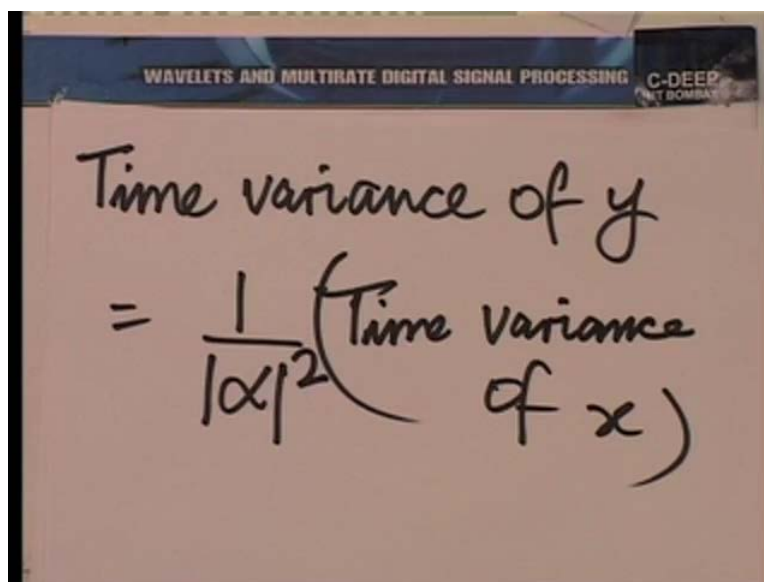
$$\text{Time centre of } y = \frac{1}{[\alpha]} \cdot \text{Time centre of } x$$

\nwarrow not modulus.

It is very easy to show by making that substitution that the time centre of y is equal to 1 by mod α times the time centre of x . Actually not mod α , I am sorry, it should be just one by α because we must take in to account the sign as well, not modulus. You must know where it is modulus and where it is not. Here, it is not modulus.

So, for example, I will take the example of α equal to minus 1. If α is minus 1, then the time centre also reflects it. So, the time centre of x was t_0 ; the time centre of y becomes minus t_0 . So, it is not mod α , it is just α . Sometimes, simple intuitive reasoning can also clarify certain points for us. Anyway, here it is not modulus of α .

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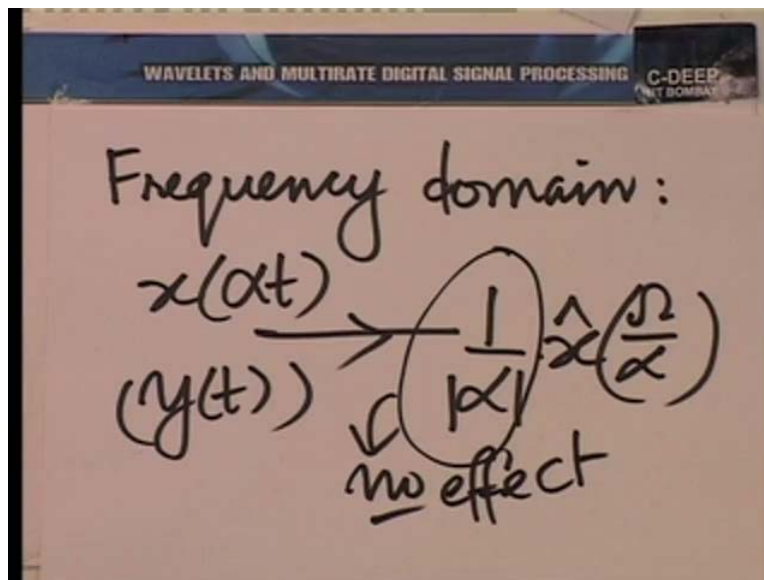


The image shows a handwritten equation on a slide. The slide has a header that reads "WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING" and "C-DEEP" on the right. The equation is written in black ink on a light-colored background. It states: "Time variance of y " followed by an equals sign, then a fraction $\frac{1}{|\alpha|^2}$ multiplied by "(Time variance of x)".

$$\text{Time variance of } y = \frac{1}{|\alpha|^2} (\text{Time variance of } x)$$

Similarly, one could use the same transformation and show the time variance of y . Now, here it is modulus. So, it will be one by modulus squared of the time variance of x . And in fact, if we now note that what we are doing in time is reversed in frequency. So, in time, we have t being replaced by αt ; in frequency, capital ω gets replaced by capital ω divided by α with the scaling of one by mod α . But, please remember the mod α scaling is not going to affect either the time centre or the time variance or the frequency centre or the frequency variance. So, scaling the independent variable, scaling the dependent variable has no effect. We saw that few minutes ago. Scaling the independent variable has an effect. Let us make a note of this.

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP
M T BOMBAY

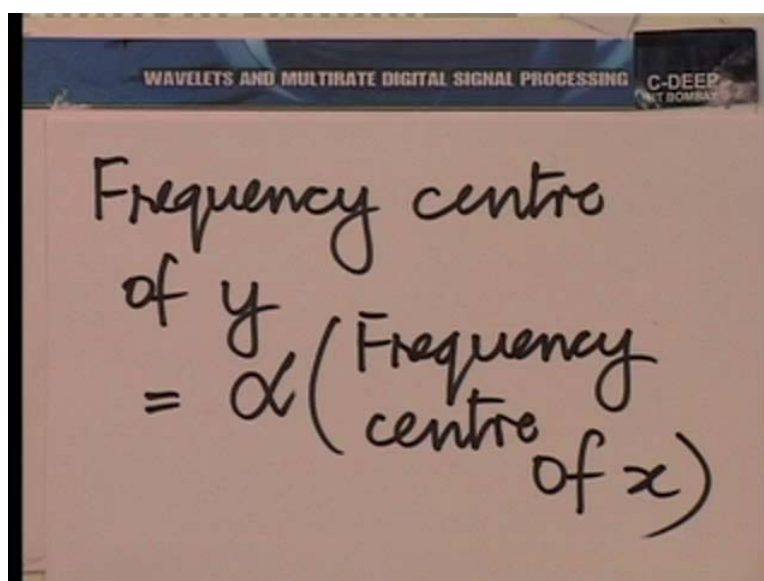
Frequency domain:

$$x(\alpha t) \rightarrow \frac{1}{|\alpha|} \hat{x}\left(\frac{\Omega}{\alpha}\right)$$

$(y(t))$ \downarrow no effect

So, we have, in the frequency domain x of αt , which is essentially $y t$ corresponds to 1 by $\text{mod } \alpha \times \text{cap } \omega$ by α . And, this has no effect. No effect on the variances or the centre as we saw a minute ago. It is essentially, this that we need to look at carefully; this one, the one by α scaling. And, one can use set of steps very similar to what we did in the time domain and arrive at the following conclusions.

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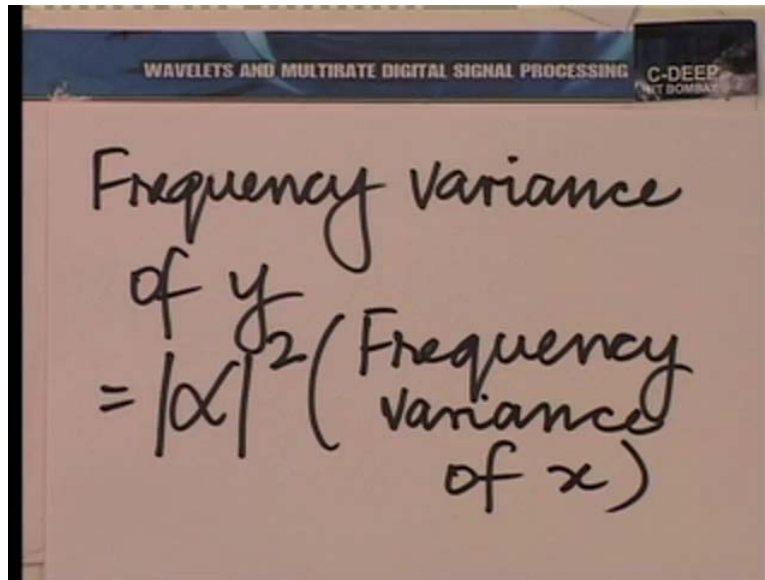


WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP
M T BOMBAY

Frequency centre
of y
 $= \alpha$ (Frequency
centre of x)

So, the conclusions are the frequency centre of y is equal to α times the frequency centre of x .

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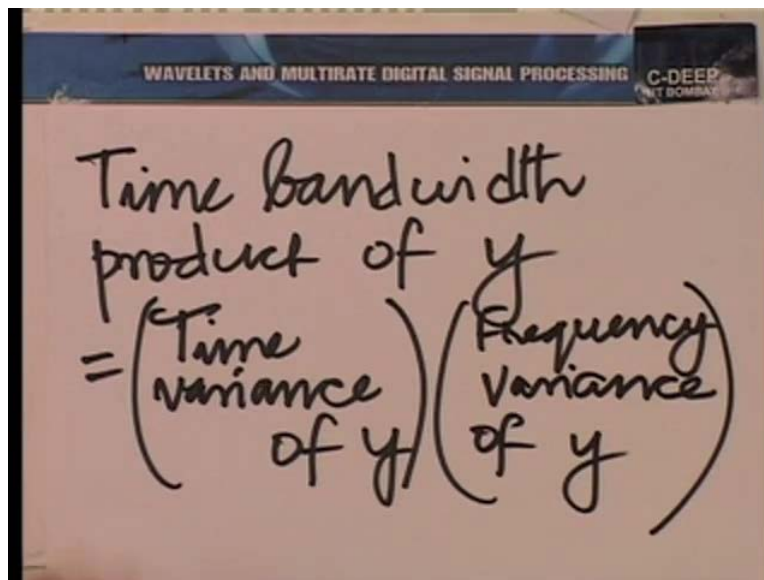


The image shows a handwritten equation on a slide. The slide has a header that reads "WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING" and "C-DEEP". The equation is written in black ink on a light-colored background. It states: "Frequency variance of y = $|\alpha|^2$ (Frequency variance of x)".

$$\text{Frequency variance of } y = |\alpha|^2 (\text{Frequency variance of } x)$$

And, as far as the frequency variance **course**, the frequency variance of y is more α times the frequency variance of x . I leave it to you as an exercise to prove these. It is easy to do by making the same kind of substitution of variable. Anyway, now let us look at the time bandwidth product.

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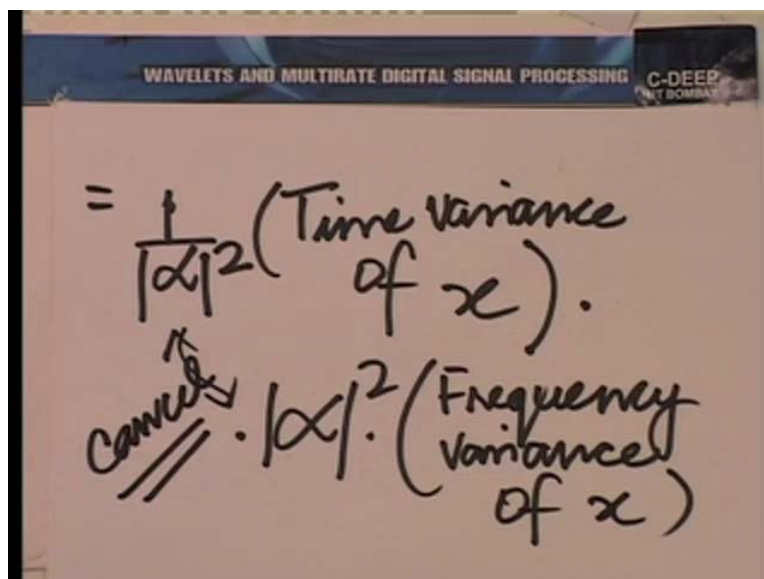


WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP
M.T. BOMBAY

$$\text{Time Bandwidth product of } y = \left(\begin{matrix} \text{Time} \\ \text{variance} \\ \text{of } y \end{matrix} \right) \left(\begin{matrix} \text{Frequency} \\ \text{variance} \\ \text{of } y \end{matrix} \right)$$

The time bandwidth product of y is essentially the time variance of y multiplied by the frequency variance of y .

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP
M.T. BOMBAY

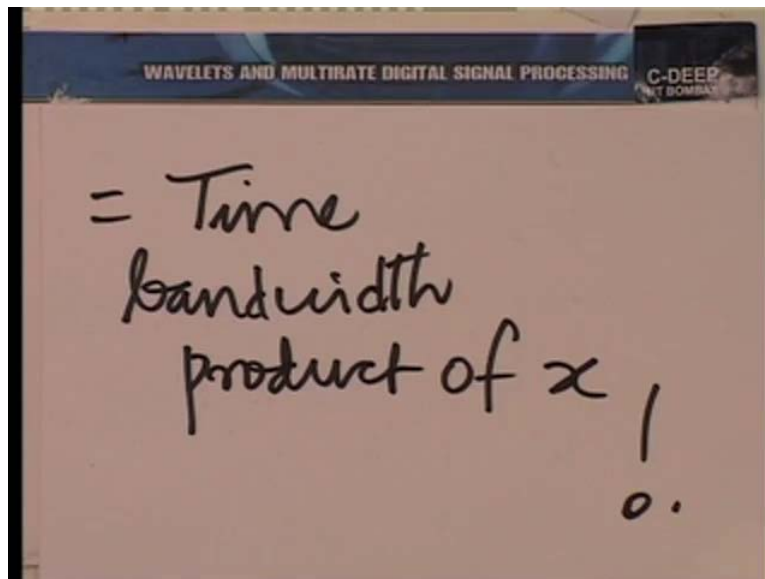
$$= \frac{1}{|\alpha|^2} (\text{Time variance of } x) \cdot \cancel{\alpha} \cdot |\alpha| \cdot (\text{Frequency variance of } x)$$

But, the time variance of y , by this argument is clearly seen to be 1 by mod α squared times the time variance of x multiplied by... So, as I am continuing this as a product here, multiplied by the frequency variance of y is mod α squared times the frequency variance of x . That is

very interesting. The time variance is multiplied by 1 by mod alpha squared and the frequency variance is multiplied by mod alpha squared.

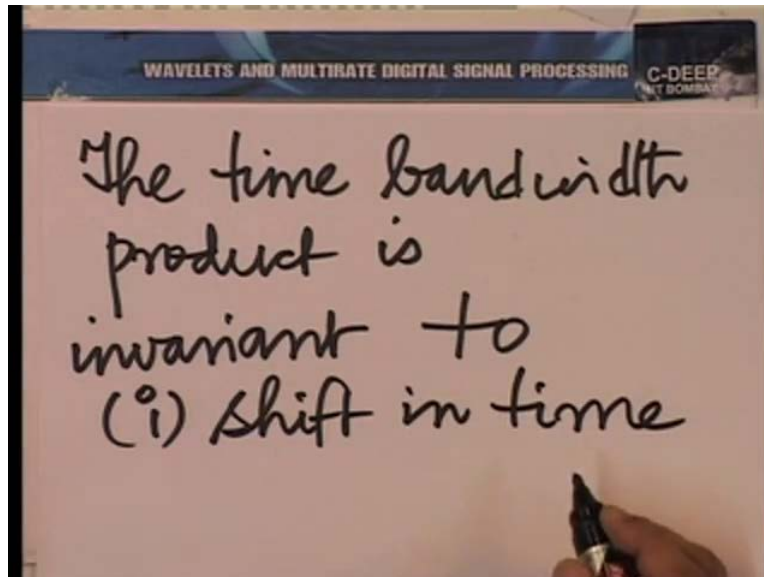
So, when you take a product, these two cancel. And, because they cancel, you get just the time variance of x times the frequency variance of x , which is essentially the time bandwidth product of x .

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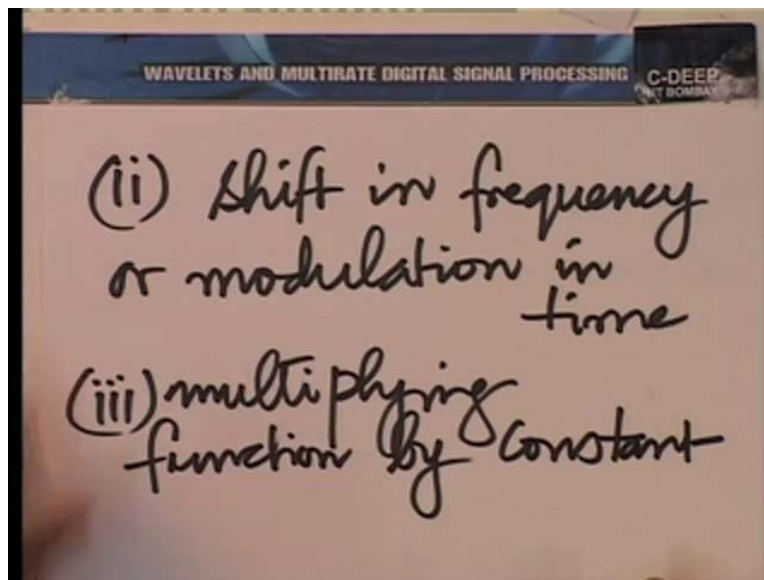
This is a very significant statement that we have made. What we have just shown is that the time bandwidth product is unaffected by scaling of the independent variable as well. It is very strongly invariant. Thus, we can see it is a very serious statement. Let us write it down.

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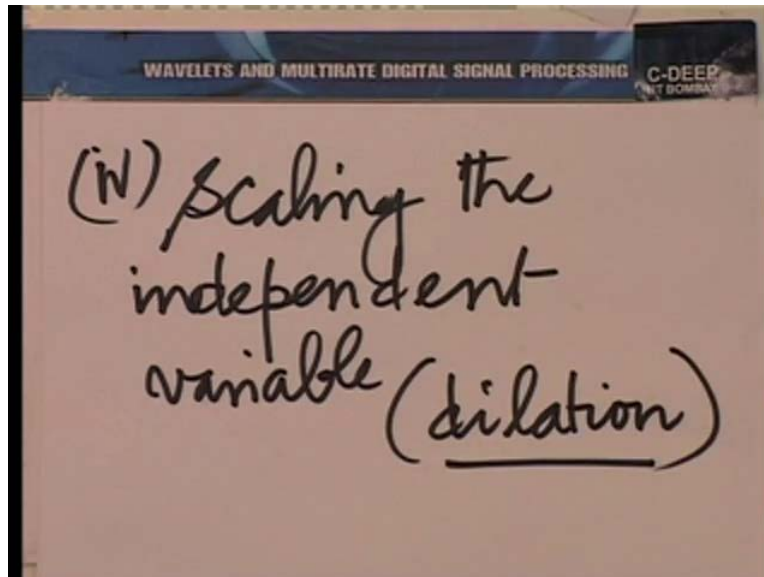
The time bandwidth product is invariant to, number 1: shift in time.

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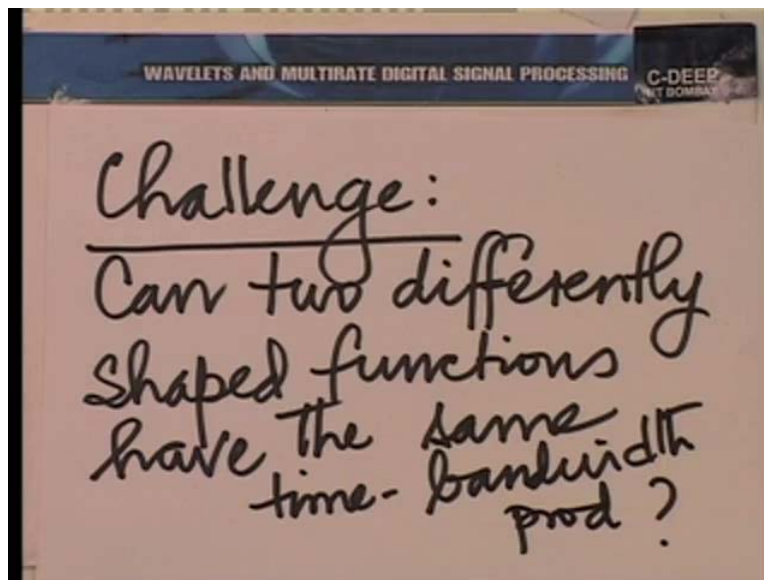
Number 2: shift in frequency or modulation in time. Number three: multiplication by a constant, multiplying the function by a constant.

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And finally, number four: the most interesting of them all. Scaling the independent variable, essentially what we have been calling dilation, **all these while** dilation by alpha. So, it is invariant to dilation. So, the time bandwidth product is something very fundamental about a function. What is it then variant **two**, what is it change **with**. A change is essentially with the shape. So, different shapes have certain time bandwidth product associated with them. That is interesting. Now, I put before you a question for thought. I like to put certain questions to challenge your imagination. And, this is one of them.

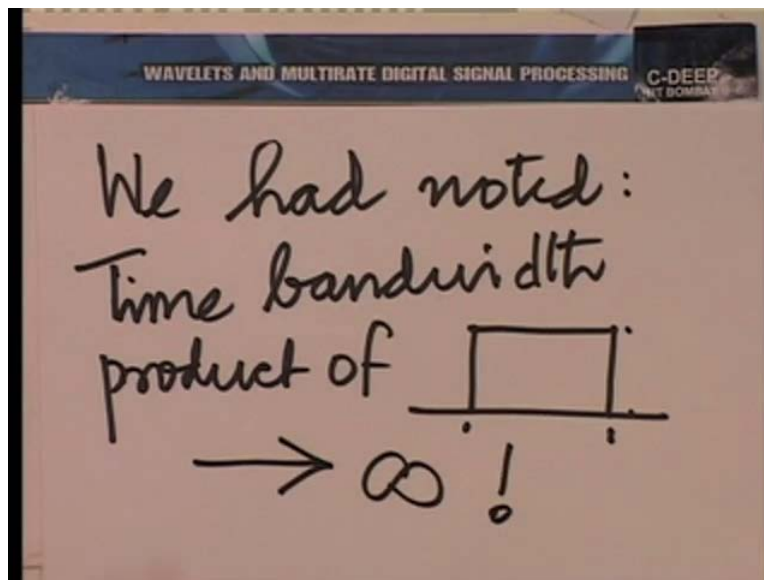
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Can two different shaped functions have the same time-bandwidth product? And, so I will not leave you entirely in the dark, I will give you a hint. What could happen, if you took the Fourier transform of a function? Can one employ the property of duality somewhere and with that slight hint, which I am sure should take you rather far, I shall leave the rest of the reasoning to you to answer this question. Anyway, that could also bring forth one more kind of invariance.

So, it is in some sense, giving you two gifts all at once. And, answer to this question and one more kind of invariance at this time-bandwidth product exhibits. Anyway, so, now what we have seen is that this time-bandwidth product is a very important quantity in the context of functions. It, in some sense, is characteristic of the shape. Though, I must remark not unique to the shape, but characteristic of the shape. Now, we saw what was the time-bandwidth product of this pulse in the previous lecture.

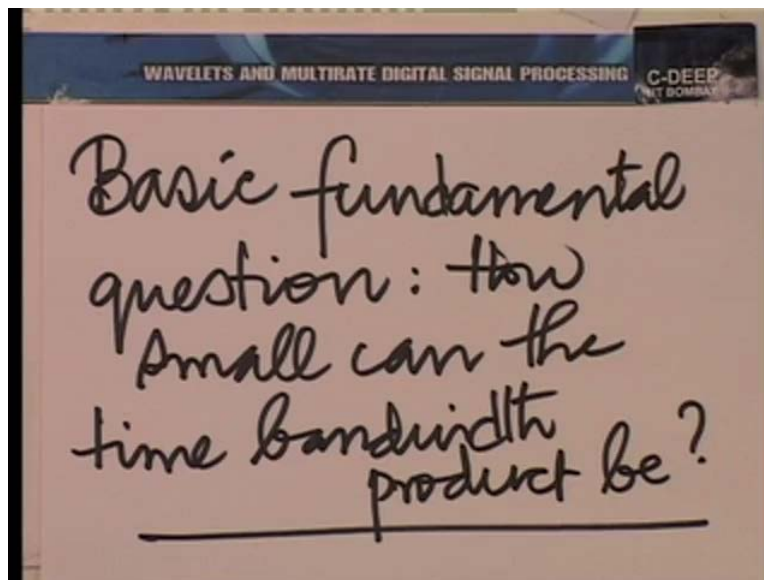
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Last time we had noted that the time bandwidth product of this pulse, now you will agree with me that I do not really need to write the limits here, nor do I need to give you the extent of this nor even, I need to give you the height of this. Just the shape would do because whatever be this extent, whatever be this height, the time bandwidth product is all the same. But, as you noted, the time bandwidth of this product tends to infinity. And, we had asked why this is terrible... told us why we are not contented with the Haar.

We are dealing with Haar in terms of the scaling function or the wavelet function. You are essentially working with functions of an infinite time bandwidth product. Having the large time bandwidth product is bad. It means that, I cannot localise nicely in time and frequency. So, the whole objective, the whole game is to see how small you can make this time bandwidth product.

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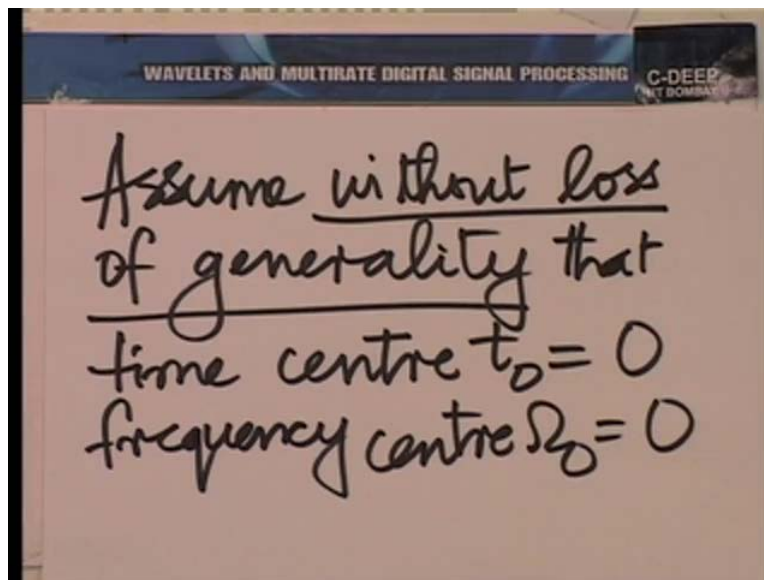


So, that is the question that the answer that the principle is going to ask. And, we shall formulate that question today; a basic fundamental question. How small can the time bandwidth product be? The Haar terrible infinity. Well, we shall have good news very soon.

We shall be able to come up with functions, whose time bandwidth product is not bad at all. But then, we will again see what nature often does to us, poor scientists and engineers. You can reach within a certain level of this bound. That, we are going to come too soon. But, reaching the bound itself is impossible. Anyway, that is just a prelude. Now, what we need to do is to establish at bound. And, to establish at bound, let us first simplify our work as we should always do. Try and take away unnecessary trappings from the problem, let us identify the essence of the core of the problem and then try and solve the problem.

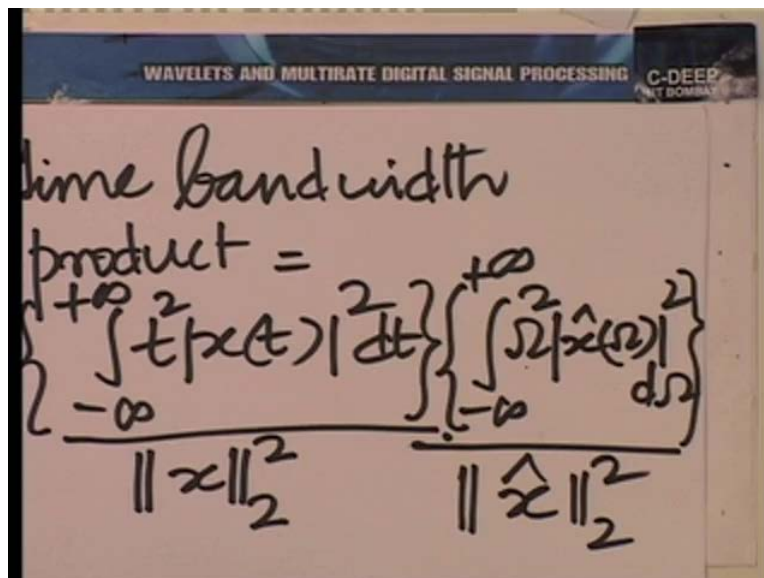
So, what we are trying to establish is the fundamental, know a bound on the time bandwidth product. We already noted the invariance of the time bandwidth product to a number of different operations namely shifting in time, shifting in frequency, multiplication by a constant and scaling the independent variable. Now, what we shall do therefore? Is to do away with the requirement of time and frequency centre.

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So, let us first, to answer this question, assume without loss of generality. And, this without loss of generality is because of all that I just said that the time centre and the frequency centres are both 0. After all, if the time centre is not 0, we could always shift and make the time centre 0 in time, without affecting a time bandwidth product. If the frequency centre is not 0, we can always shift in frequency; which means modulate in time and bring this frequency centre to 0. Of course if the function is real, we do not even need to do that. So, this is without loss of generality.

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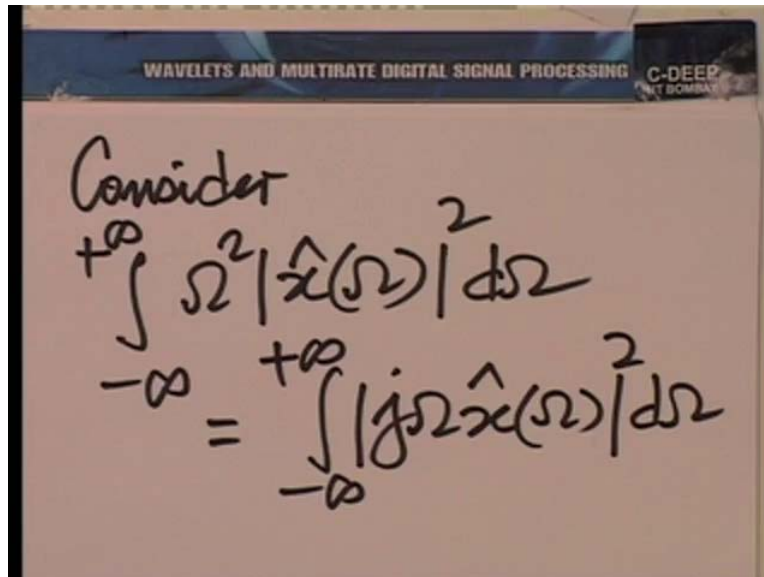
The slide shows a handwritten formula for the time-bandwidth product. The title at the top is "WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING" with "C-DEEP" and "IIT BOMBAY" on the right. The text "time bandwidth product =" is written. The formula is:

$$\left\{ \int_{-\infty}^{+\infty} t^2 |x(t)|^2 dt \right\} \left\{ \int_{-\infty}^{+\infty} \omega^2 |\hat{x}(\omega)|^2 d\omega \right\}$$

Below the first integral is the norm $\|x\|_2^2$, and below the second integral is the norm $\|\hat{x}\|_2^2$. The entire expression is a product of these two fractions.

Let us make the time centre and the frequency centre zero; where upon, we then have the time bandwidth product to be essentially the following. It is integral minus to plus infinity the whole squared of x multiplied by t squared $d t$ divided by the norm of x squared times something **similar in frequency**. So, we will keep this here; minus to plus infinity omega squared mod x cap omega squared d omega. And, here again divided by the norm of x cap the whole squared. Now, you have already made an observation about this. And, we formally make that observation again.

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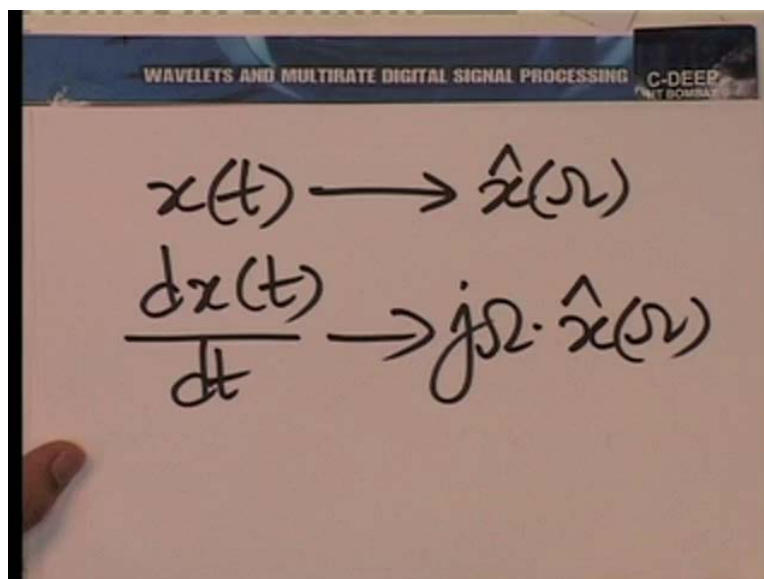


The slide shows a handwritten derivation. It starts with the word "Consider" followed by the integral $\int_{-\infty}^{+\infty} \omega^2 |\hat{x}(\omega)|^2 d\omega$. This is then equated to $\int_{-\infty}^{+\infty} |j\omega \hat{x}(\omega)|^2 d\omega$.

$$\text{Consider } \int_{-\infty}^{+\infty} \omega^2 |\hat{x}(\omega)|^2 d\omega = \int_{-\infty}^{+\infty} |j\omega \hat{x}(\omega)|^2 d\omega$$

We already seen that, this essentially is minus to plus infinity j omega x cap omega mod whole squared d omega.

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The slide shows two handwritten Fourier transform pairs. The first is $x(t) \rightarrow \hat{x}(\omega)$. The second is $\frac{dx(t)}{dt} \rightarrow j\omega \cdot \hat{x}(\omega)$.

$$x(t) \rightarrow \hat{x}(\omega)$$
$$\frac{dx(t)}{dt} \rightarrow j\omega \cdot \hat{x}(\omega)$$

And that, we have noted is essentially, you see, noting that if $x(t)$ has the Fourier transform $\hat{x}(\omega)$, then $\frac{dx(t)}{dt}$ has the Fourier transform $j\omega \hat{x}(\omega)$.

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$$\int_{-\infty}^{+\infty} |\hat{x}(\omega)|^2 d\omega = 2\pi \int_{-\infty}^{+\infty} \left| \frac{dx(t)}{dt} \right|^2 dt$$

Where upon, integral minus to plus infinity $\int_{-\infty}^{+\infty} |\hat{x}(\omega)|^2 d\omega$ is essentially integral from minus to plus infinity $\int_{-\infty}^{+\infty} \left| \frac{dx(t)}{dt} \right|^2 dt$, but thus a factor of 2π . So, it is 2π times this.

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$$\text{Time bandwidth product} = \frac{\|tx(t)\|_2^2}{\|x\|_2^2} \cdot \frac{\left\| \frac{dx(t)}{dt} \right\|_2^2}{\|x\|_2^2}$$

And, of course we also see that norm of x cap squared is 2π times the norm of x squared. And therefore, if we now put back all together here; when I put all these together, I keep these as the

R in the time bandwidth product. And, I note that this is essentially the norm squared or the energy in the derivative divided by the energy in the function, the factor of 2π cancels from the numerator and denominator here.

So, we could write down all in all; thus the time bandwidth product can be written as essentially the energy, the L^2 norm of $t \times t$ whole squared divided by the L^2 norm of x multiplied by the L^2 norm of $d x / dt$ whole squared divided by the L^2 norm of x the whole squared. A very interesting result. I will just remind you that when we put back the expression for the time bandwidth product, this is essentially the L^2 norm of $t \times t$. and, that is how we come to this conclusion.

So, here of course, there is the bit of abuse of notation. We are talking about a function $t \times t$ and then taking its L^2 norm. So, this is ratio of two products of L^2 norms. And, now here, we have all set to minimise this product. In fact, that would be our next step in the beginning of the next lecture. What is the minimum value of this product, minimum overall x that can exist in $L^2 \mathbb{R}$. And, that would give us a very fundamental bound in nature, which is called the uncertainty bound. We shall derive that uncertainty bound in the coming lecture. Thank you.