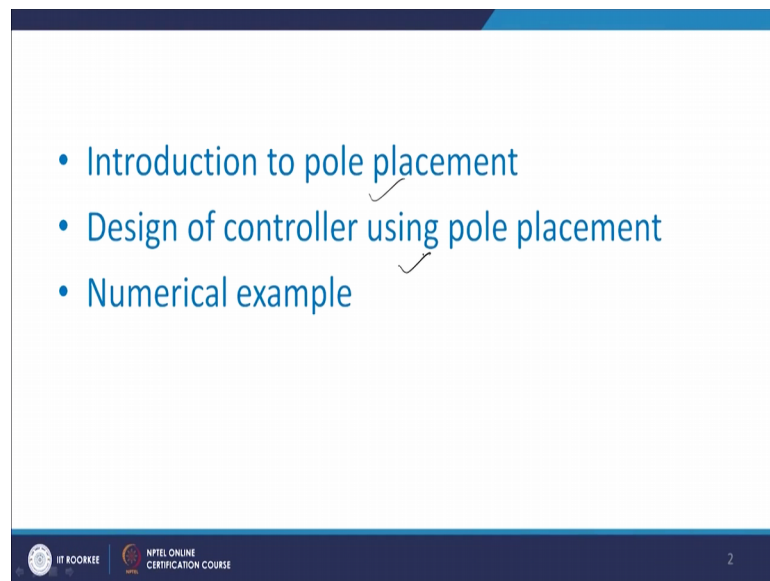


Advanced Linear Continuous Control Systems
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Lecture - 33
Pole Placement by State Feedback (Part-I)

Now we start with Pole Placement by State Feedback.

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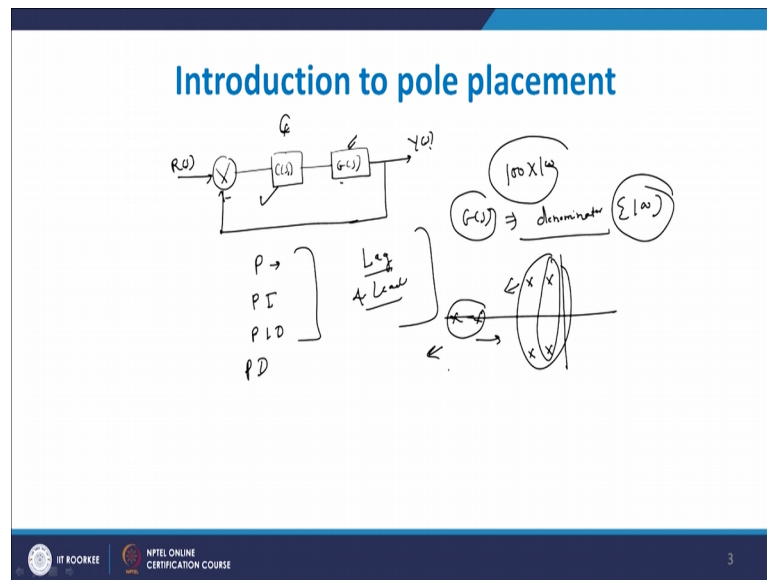


- Introduction to pole placement
- Design of controller using pole placement
- Numerical example

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In this we will study introduction to pole placement, design of controller using pole placement and we will see one numerical example now.

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Introduction 2 pole placement basically what is mean your pole placement pole placement means that we have design a controller. It is similar to the controller which we are design in a classical approach. For example, in classical approach what we are seen? We have a plant it is like this.

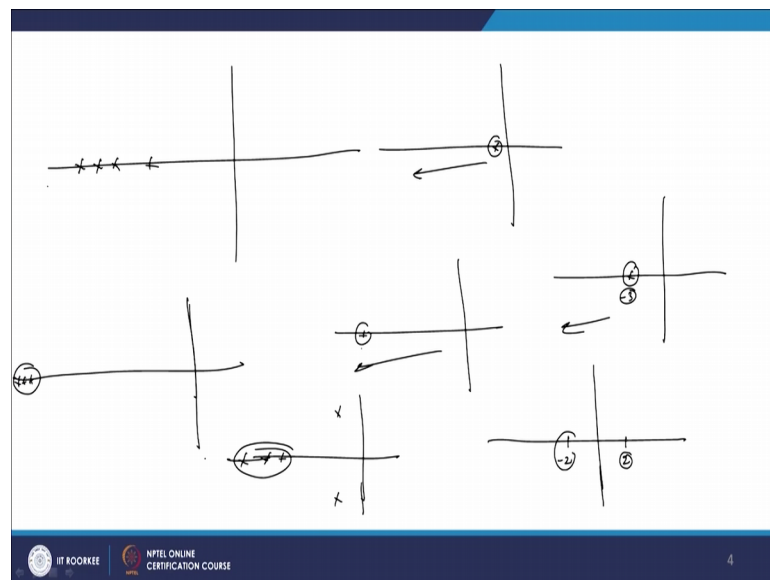
Let us say this is controller C of s , then this is a plant G of s now this is feedback control system its negative say R s and Y s first of all why this need for controller? A controller is needed because this is a plant G s in open loop response is not satisfactory, then without controller means without taking this distance. We have taken a feedback control system; still performance is not satisfactory and because of that we have designed a this controller C of s . We know that the classical controllers are P controller there is proportional, then proportional integral then PID , P PI PID sometimes even a PD also other controllers are say lag and lead.

These are the classical controller. So, what is the problem with this controller is that as our purpose is to get a desired result whatever is our objective is that need to be achieved. So, we will find that if this G s all the plant of is a very high order, let us say 100 order plant 100 band rate or the G s G s the denominator order is say 100 denominator say order is a 100. So, in that case if you design a PID controller can we get the satisfactory results result means what? Let us say let us say there are this poles are here these are poles, and here is somewhere these are 2 poles.

Now, here are problem is that the poles which are nearer to imaginary axis that need to be shifted far away. So, if you what do shift this using the PID controller. So, what will happen these poles may shift this may shift, but what about these 2 poles? So, if may possible that this pole may more like this or may more like this, this no guarantees therefore, if you want a desired response. So, desired performance for higher order system this is not possible by means of a conventional classical controller.

So, here we can use a state feedback controller.

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So, what is the advantage of this state feedback controller is that, you can place pole anywhere in the; you can place pole here, you can place here you can place here then you can place here someone we say no I we want to place the pole at the (Refer Time: 04:26) only here. So, this is possible because poles governed a performance. So, you have a desired object that you can place the pole anywhere it means that, you can get the desired response desired performance can be achievable.

But there are some issues with respect to this. Problem is that, if you move rgese poles let us say original pole is at here. So, the pole which is here it has say will last time constant, but if you moving here this pole this side. So, what is the time constant time constant is reduces and when we are doing a some measurement, because our approach is state feedback state we have to measured. So, when we are present pole far away where

time constant is very very small, we required high quality sensors means acquire censored, then required actuator the large actuator that is the motors.

So, all these involve a more cost and again there should be again we needed to means we have to make it available, sometime it may not be available. Therefore, there is an issue about the pole placement. We got place the pole anywhere all though we can resign theoretically practically, there are some issues. Again the problem is this pole placement is that, that we have to measure the all the states. If some states cannot be measurable, then there are some issues we cannot design a controller. Although in future we will see design of the observer, but in actual practice if you want to design a controller we have to measure the all the states, and second as we have seen that as we move the poles far away from imaginary axis, then in that case we required a very high quality actuators so, there is some the problem.

Therefore, whenever we design a pole placement first of all we have to see the requirement. Suppose if you have a pole at here now let us say as minus 3. So, if you know if the pole at a minus 3, if it at proper positions. So, there is no need and to move this pole left side because our requirement is satisfactory; that means, you have to move the poles left side when there is a requirement.

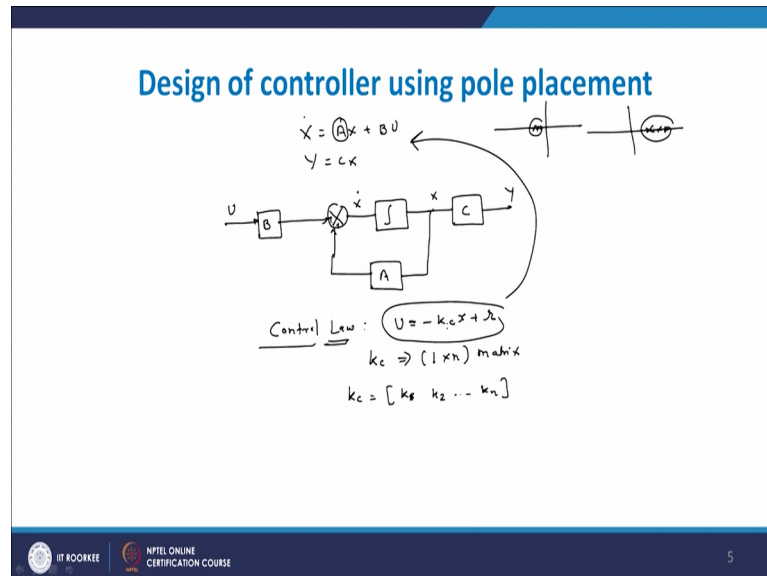
If you want to reduce a some control energy let us say one pole on the right hand side there is 2 you can shift at say minus2. So, if you make it like this that is another system has pole has 2 if we move at minus 2 we required very less control energy. So, that is some say thumb rule we can also use. Other thing which can be used that, because the system sometimes required transient response. So, we can place the pole that is some poles are here. So, that the transients issues resolved and other poles we can place somewhat here on the real axis.

So, depend upon the problem, depend upon the availability, depend upon the cost associated and mainly the availability. So, we can design a controller, but the most important thing is that if we having the all facilities everything is available then, in that case this pole placement is better technique than the classical controller, but in most of the times we do not have the resources. So, in that case that this classical control is far from better. Because in a classical control it depend upon the input $R(s)$ and output $Y(s)$,

but whereas, if you (Refer Time: 08:05) takes place we have to measure the all the state which are involved in the system.

Now, we will see, how to design a controller using pole placement technique.

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Now we are designing the controller using pole placement techniques so, how our system is basically \dot{X} is equal to $Ax + Bu$ this is our system Y equal to Cx ; and here we are designing controller because A is the system matrix, the system matrix A has pole maybe on the right hand side, it may have poles which are near to imaginary axis; that means, this is the conditions, A matrix may have eigenvalues which are nearer or it may possible that eigenvalue of the right hand side that system is unstable.

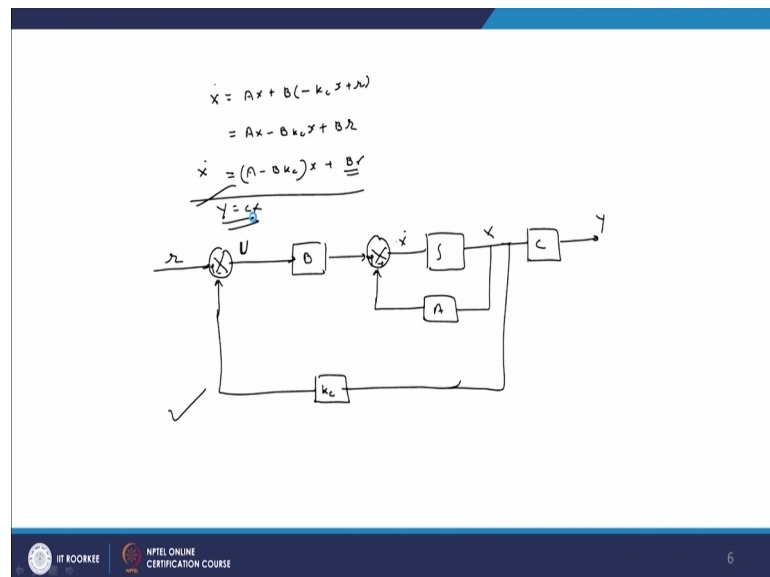
So, we have to stabilize this A or we have to improve the performers of the A ; now if you draw the block diagram for this. So, you can relate now here is one integrator, now this x now here this is C this is output y now here is A matrix now here \dot{x} this is \dot{x} now here is B this is u . So, you will find that for this case your plant is this \dot{X} that is equal to $Ax + Bu$ and y equals to Cx .

So, this is the broader on this step. Now we have to design; a controller the controller we design; that means, you have to design a control law and this control law is through input; that means, u is place the important role how much input we have to provide? Therefore, you must be taken into considerations and this u must have relationship with

the controller therefore, the control law which we can consider as the control law which we can considered as U equal to say minus $k_c x$ plus r this is a control law, where this k_c is belong to $1 \times n$ matrix; that is this k_c is nothing but k_1, k_2 up to say k_n .

So, this k_1, k_2, k_n ; that means, this one k_1 for one state, k_2 for other state k_n for another state because this basically graded to the input and with the state that is why the U equal to minus $k_c x$ plus r . So this is the control law, now we have to use this control law in this equations \dot{x} equal to $Ax + Bu$. Now if you replace u equal to this in this particular equations.

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So, we will get we will get this \dot{x} equal to Ax plus this is B , u is minus $k_c x$ plus r now if you solve it. So, it will be the Ax minus B into k_c into x plus B into r and now if you solve it, it will be A minus say B to k_c into x plus B r now this is \dot{x} .

So, though this is the equation using the control law whereas, original equation is like this. So now, we have draw on the block diagram for this model now as our system has changed. So, you come into existence our controller is come into existence therefore, our block diagram is changed. So, now, we redraw the block diagram as. So, here let us say this is integrator C output y .

So, here \dot{x} this is state x this is the integrator, now here this is A , this is B , here has like this is r , now it is positive and now here we have got \dot{x} equal to $Ax + Bu$ and

now what is this is u you will find that, this u equal to minus k c x; that means, here we have to use a control law as this k c is a controller. So, it is negative and it is coming to the state x. So, if you see is x dot equal to A x + B u where B u, u equal to this equations has been computed and here Y equal to c x.

So, this is the block diagram of the system and now it become a close loop, because the state has been control here there is no access control in effect. Now we will see how to design a controller using state feedback control approach and whereas, or we can say the control law which we are taking here u equals to minus k c x so, here the plant which you will take.

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$$G(s) = \frac{1}{s^n + a_{n-1}s^{n-1} + \dots + a_1s + a_0}$$

$$\dot{x} = Ax + Bu$$

$$[A] = \begin{bmatrix} 0 & 1 & 0 & \dots & 0 \\ 0 & 0 & 1 & \dots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & 0 & \dots & 1 \\ -a_0 & -a_1 & -a_2 & \dots & -a_{n-1} \end{bmatrix} \quad [B] = \begin{bmatrix} 0 \\ \vdots \\ 1 \end{bmatrix} \quad k_c = [k_1 \ k_2 \ \dots \ k_n]$$

$$[A - BK] = \begin{bmatrix} 0 & 1 & 0 & \dots & 0 \\ 0 & 0 & 1 & \dots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & 0 & \dots & 1 \\ -a_0 & -a_1 & -a_2 & \dots & -a_{n-1} \end{bmatrix} - \begin{bmatrix} 0 \\ \vdots \\ 1 \end{bmatrix} [k_1 \ k_2 \ \dots \ k_n] = \begin{bmatrix} 0 & 1 & 0 & \dots & 0 \\ 0 & 0 & 1 & \dots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & 0 & \dots & 1 \\ -a_0 - k_1 & -a_1 - k_2 & \dots & \dots & -a_{n-1} - k_n \end{bmatrix}$$

Suppose the plant G of s is taken has 1 divided by S raise to n plus a n minus 1, s raise to n minus 1 plus a n minus 2, s n minus 2 plus a 1 s plus a naught now, this is a G of s.

But now here we are designing a controller for a system which is represented in a states space form. Therefore, the state space model of the system it is in terms of x dot equal to A x plus B u and therefore, here we will take this A matrix as, this A matrix will write in a companion form that is 0 1 0 0 then here 0 0 1 0 and here this process repeated it is still 0 0 1. And last row you are write a naught minus a 1 minus a 2 up to minus a n minus 1 this is a matrix, and now here b matrices 0 0 say 1, and here this k c is nothing but k 1 k 2 up to say k n, this is controller k c a we have written b.

Now here we are designing a controller because the performer of this system is not satisfactory therefore, we are designing the controller. So, here as per the control law, u equals to $-\mathbf{k}x + r$. So, we have got the system matrix as $A - B\mathbf{k}$. Original system matrix is A , A has some problem. So, the modified system matrix is $A - B\mathbf{k}$ therefore, here we can write down this matrix as $A - B\mathbf{k}$ that is equal to $\begin{bmatrix} 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & 1 & 0 \end{bmatrix} - \begin{bmatrix} 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix} \mathbf{k}$ and here we are get $\begin{bmatrix} 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & 1 & 0 \end{bmatrix} - \begin{bmatrix} 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix} \mathbf{k}$ that is like this $\begin{bmatrix} 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & 1 & 0 \end{bmatrix} - \begin{bmatrix} 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix} \mathbf{k}$ and here it has been multiplied by \mathbf{k} $\begin{bmatrix} k_1 & k_2 & \dots & k_n \end{bmatrix}$.

Now, this we have to multiply we have to subtract from this. So, after doing it we will get. So, we will get as $\begin{bmatrix} 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & 1 & 0 \end{bmatrix} - \begin{bmatrix} 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix} \mathbf{k}$ and lastly here $\begin{bmatrix} 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & 1 & 0 \end{bmatrix} - \begin{bmatrix} 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix} \mathbf{k}$ and in the last row we will get $-\mathbf{k}_1$, $-\mathbf{k}_2$ and here $-\mathbf{k}_n$. Because when this has been multiplied to this equation is nothing but $\begin{bmatrix} 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix} \mathbf{k}$ and last you will get \mathbf{k}_1 , \mathbf{k}_2 up to \mathbf{k}_n therefore, when we subtract this from this you will get $-\mathbf{k}_1$, $-\mathbf{k}_2$ minus $-\mathbf{k}_n$.

Now, for this we have to determine the eigenvalues because the eigenvalues govern the performance.

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$$\begin{aligned}
 & |\lambda I - (A - B\mathbf{k})| \\
 &= \lambda^n + (a_{n-1} + k_n)\lambda^{n-1} + (a_{n-2} + k_{n-1})\lambda^{n-2} + \dots + (a_0 + k_1) = 0 \quad \text{--- (1)} \\
 & \text{Suppose desired poles are at } \lambda_1, \lambda_2, \dots, \lambda_n \\
 &= \lambda^n + \alpha_{n-1}\lambda^{n-1} + \alpha_{n-2}\lambda^{n-2} + \dots + \alpha_0 = 0 \quad \text{--- (2)} \\
 & \alpha_{n-1} = a_{n-1} + k_n \Rightarrow k_n = \alpha_{n-1} - a_{n-1} \\
 & \alpha_{n-2} = a_{n-2} + k_{n-1} \Rightarrow k_{n-1} = \alpha_{n-2} - a_{n-2} \\
 & \vdots \\
 & \alpha_0 = a_0 + k_1 \Rightarrow k_1 = \alpha_0 - a_0 \\
 & \mathbf{k} = [k_1 \quad k_2 \quad \dots \quad k_n]
 \end{aligned}$$

So, the eigenvalues of this system we can write down as λ_i minus $A - B\mathbf{k}$ that is equal to. So, we will get $\lambda^n + (a_{n-1} + k_n)\lambda^{n-1} + \dots + (a_0 + k_1) = 0$.

$s^{n-1} + a_{n-2}s^{n-2} + k_{n-1}s^{n-1} + \lambda^{n-2}$ and this process repeated plus $a_0 + k_1$.

So, if you see the equations. So, this is in companion form. So, what is the equation companion form; we can directly write down the equation like this. So, $\lambda^n + a_{n-1}\lambda^{n-1} + a_{n-2}\lambda^{n-2} + \dots + a_0 + k_1$. Now here this is the equation this we want, but here we know the though the original system matrix say, but this k_1, k_2, \dots, k_n we do not know. So, they need to calculate this they are how to get it. So, here in this case we have to decide where we want to place the poles.

So, we assume that our poles which are required to be place let us say it suppose the desired poles are at say λ_1, λ_2 say up to λ_n these are my desired poles. So, it can plus anyone in they explain. So, it has any values, but these are my poles. Now if you based on the desired poles we can also obtain the caustic equation. So, using these desired poles the caustic equation can be written as $\lambda^n + \alpha_{n-1}\lambda^{n-1} + \alpha_{n-2}\lambda^{n-2} + \dots + \alpha_0 = 0$.

So, now this is the caustic equation this is our equations. So, these 2 equations are there and here we have find that this is my desired one; and this and this they have the same form $\lambda^n + \lambda^n$. These are the non quantities and here $a_{n-1}, a_{n-2}, \dots, a_0$ these are non quantities, but which is are unknown is k_1, k_2 up to k_n and this is nothing but the controller. So, if we compare these equation with respect to this equation this 2 equations here, this less the equation over 1. And let us say this is equation number 2 if we compare this 2 equations we can easily get the balance of k_1, k_2 up to k_n now we compare this equation.

So, if you compare this equation let us say α_{n-1} that is equal to $a_{n-1} + k_n$. So, if you solve it will get this $k_n = \alpha_{n-1} - a_{n-1}$. this we get then about this α_{n-2} that is equal to $a_{n-2} + k_{n-1}$. So, after solving it will get $k_{n-1} = \alpha_{n-2} - a_{n-2}$ and now we come to this equation α_0 and this. So, if we compare this part. So, we will get α_0 that is equal to $a_0 + k_1$ and again if you solve it, we will get $k_1 = \alpha_0 - a_0$. So, here k_1 we will get similarly get k_2, k_{n-1} and k_n .

So, in this way we have got the controller as $k_1 k_2$ up to k_n . So, this controller k will stabilize the system as per our requirement. As per the simulation is considered you can desire the pole anywhere whatever we required that desired pole space model is possible, desired controller is possible, desired performance is possible. So, this is where we can see the beauty of the advanced control approach or a state feedback approach where this is not possible in case of the classical approach. But although, I will tell that all that theoretically is ok, but practically there are some issues. As I told you that we required high quality actuator even a sensor because we have to sense.

So, if you are far away from the control input is more and that time it may get some problems that is we required very high quality actuators. So, now, this is finally, controller and our main purpose is to design this controller k and we have already designed it, but it has been designed through a mathematical approach that is for n th order. Now we will see how to get the volume of the controller from a numerical example. So, we will solve a numerical example.

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Numerical example

$\dot{X} = AX + BU$

$A = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & -3 & -4 \end{bmatrix} X + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} U$

Desired poles = $-2, -1+j, -1-j$

$(\lambda+2)(\lambda+1+j)(\lambda+1-j) = 0$

$(\lambda+2)(\lambda^2+2\lambda+2) = 0$

$(\lambda+2)(\lambda^2+2\lambda+2) = 0$

$\lambda^3 + 4\lambda^2 + 6\lambda + 4 = 0$

Handwritten notes: $1-2-j$, 1 , 1

So, your problem is $\dot{X} = AX + BU$, now here A matrix is $0 \ 1 \ 0 \ 0 \ 0 \ 1 \ 0$ minus 3 minus 4 x plus B is $0 \ 0 \ 1$ into u .

And our desired poles that the desired poles are at -2 minus 1 plus $-j$ 1 . These are our desired poles you will find that the desired poles in what we were taken the desired pole we have taken in terms of -1 to j 1 that is basically for the satisfied by

the transient issues and minus 2 if the real poles. You can take desired poles as minus 1 minus 2 as well as minus 3 also.

But in that case your system is completely odd and then there are no transients, sometimes for faster response we have to make this type of arrangement also now these are my desired poles, you can take other poles also. And now we want this poles so, what will be the value of the controller? So, for desired poles we solve the equation as lambda plus 2, lambda plus 1 plus j 1 lambda plus 1 j 1 equal to 0 then here lambda plus 2 lambda one whole square plus 1 equal to 0.

Now, if you solve it lambda plus 2 lambda square plus 2, lambda plus 2 equal to 0 and after solving it we will get lambda cube 4 lambda square plus 6 lambda plus 4 equal to 0. Now this is my desired caustic equation, this is we have got that is this is similar to our equation lambda raise to n plus alpha 0. So, this equation we have got here. Now this is desired one, but when we are using the control law that is u equals to minus k x in the system. So, we have got the equation in terms of a minus b into k c.

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$$\begin{aligned}
 & \lambda I - (A - BK_c) \\
 &= \begin{bmatrix} \lambda & 1 & 0 \\ 0 & \lambda & 1 \\ 0 & -3 & -4 \end{bmatrix} - \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} \begin{bmatrix} k_1 & k_2 & k_3 \end{bmatrix} \\
 &= \begin{bmatrix} \lambda & 1 & 0 \\ 0 & \lambda & 1 \\ 0 & -3 & -4 \end{bmatrix} - \begin{bmatrix} 0 & 0 & k_3 \\ 0 & 0 & k_2 \\ k_1 & k_2 & k_3 \end{bmatrix} \\
 &= \begin{bmatrix} \lambda & 1 & 0 \\ 0 & \lambda & 1 \\ -k_1 & -3-k_2 & -4-k_3 \end{bmatrix} \\
 & \underline{| \lambda I - (A - BK_c) |} = \lambda^3 + (4+k_3)\lambda^2 + (3+k_2)\lambda + k_1 = 0 \quad \text{(2)} \\
 & \qquad \qquad \qquad \begin{matrix} k_1 = 4 & ; & 3+k_2 = 6 & & 4+k_3 = 4 \\ & & \boxed{k_2 = 3} & & k_3 = 0 \end{matrix} \\
 & \qquad \qquad \qquad \underline{K_c = [k_1 \ k_2 \ k_3]} \\
 & \qquad \qquad \qquad \underline{= [4 \ 3 \ 0]}
 \end{aligned}$$

So, we determined this equation as a A minus B to k c that is equal to 0 1 0 0 0 1 0, minus 3, minus 4, 0 0 1 k 1 k 2 k 3. Now, we have to solve it. So, here we will get 0 1 0 0 0 1 0 minus 3 minus 4 and here after solving it we will get 0 0 0, 0 0 0 and we will get k 1 k 2 k 3 now we subtract this separation from this. So, we will get 0 1 0 0 0 1 because

all the zeros. So, after subtracting we have same $0 \ 1 \ 0 \ 0 \ 0 \ 1$, here we will get 0 minus k_1 , that is minus k_1 then here minus 3 minus k_2 and here minus 4 minus k_3 .

Now, here A minus B into k c equation we have got, now we have to determine its caustic equation in terms of λ ; so λ^3 minus A minus B into k c . So, it gives λ^3 plus 4 plus k_3 into λ^2 plus 3 plus k_2 into λ plus k_1 equal to 0 . So, this equation in a companion form reverse way you see say minus k_1 minus 3 minus k_2 minus 4 plus k_3 . So, this equation we have got.

So, this is the equation we have got from a from the original plant and in that case we will find that k_1 k_2 and k_3 are there. So, now, this equation we obtain and now we see the this equation, this is the desired equations now if we compares let us say this equation number 1 and this equation number 2, now if we compare these 2 equations. So, what will get? We will get this k_1 equal to. So, we will just see what is k_1 here; so k_1 equal to 4 . So, we will get k_1 equal to 4 then about another term this λ term and here is a λ term. So, you have 6.

So, this 3 plus k_2 this equal to 6 , so here k_2 equal to 3 , so k_2 equal to 3 and about this last one: so here 4 plus 4 into λ^2 square, so here 4 plus k_3 into λ^2 square. So, if you compare its. So, 4 plus k_3 equal to 4 So, we will get k_3 equal to 0 therefore, your controller k it is k_1 k_2 and k_3 . So, that gives you 4 this k_1 , the k_2 is 3 , and k_3 is 0 now this is the controller k .

So, this controller k we stabilize the system in such a manner that we are getting the desired performance. If you change the pole press band you will get the different values of the k . Here we will find that this given system matrix x dot equal to a x the view, that is in terms of a companion form that is basically a transfer function form and when the transfer function is there directly or when the system is given in the companion form, we can directly write the caustic equations there are no such issues.

But when the system is in the in any of the general form, then this type of things it is not easily possible we cannot direct write down with the correct equations. Therefore, we required some alternative method by which we can design a control room, that is our next expertise that how to design a controller when the given system matrix say it is in general form. And other most important point is that whenever we design a controller we have to check the controllability observability of the system. Here we are directly design

a controller, but actually when we designed a controller for any system we have to check the controllability and observability of the system; that means, particularly this case we will find that we have to basically important the controllability that is where you determine $BAB^T A^{-1} B$.

So, all this operation is based on the rank of this matrix. If the rank of this matrix or a determinant of matrix let us say 0 or whether rank is false, then this process we cannot use this process we cannot design a controller. So, controller design when this Q is non zero and the as I told you that this is the problem or these analysis we have done only when, you having a G that is in a transfer function form. But now our main task is to how to design a controller when a given system matrix in a general form.

So, that part we will see in the next class.

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So, now you see the some references.

Thank you.