

Linear Dynamical Systems
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Week - 3 and 4
Controllability and State Feedback
Lecture - 25
Test for Stabilizability

(Refer Slide Time: 00:13)


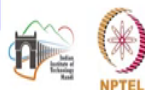
Popov-Belevitch-Hautus(PBH) Test for Stabilizability

For stabilizability, one can also reformulate the eigenvector test as a rank condition, similar to that for controllability.

Theorem (PBH test for stabilizability)

- 1 The continuous-time LTI system (AB-LTI) is stabilizable if and only if
$$\text{rank} \begin{bmatrix} A - \lambda I & B \end{bmatrix} = n, \quad \forall \lambda \in \mathbb{C} : \text{Re}[\lambda] \geq 0.$$
- 2 The discrete-time LTI system (AB-LTI) is stabilizable if and only if
$$\text{rank} \begin{bmatrix} A - \lambda I & B \end{bmatrix} = n, \quad \forall \lambda \in \mathbb{C} : |\lambda| \geq 1.$$

The proof of this theorem is analogous to the earlier proof, except that now we need to *restrict* our attention to only the "unstable" portion of \mathbb{C} .



So, the next test is the PBH test which is Popov Belevitch and Hautus Test for the stabilizability. So, for stabilizability one can also reformulate the in fact, this is what we have defined the PBH test as an elegant restatement of the eigenvector test. That the continuous time system is stabilizable if and only if this rank condition is satisfied for all lambdas belonging to the set of complex number such that the real part of those lambda is greater than equal to 0, right.

So, if you recall the result for the controllability this condition was not there while testing the rank condition, but for stabilizability as we had seen in the eigenvector test that we are only concerned with the eigenvalues which are on the right hand side. So, again for stabilizability we are concerned with the eigenvalues which are onto right hand side, ok. So, for the discrete time system we both have the same result, but those eigenvalues should be outside the unit circle or on the boundary of that unit circle.

So, the proof of this theorem is analogous to the earlier proof that is the controllability proof except that now, we need to restrict our attention to only the unstable portion of the set of complex numbers, where the eigenvalues lies, ok. So, we would not be going through the proof this PBH test. But, if we want to determine the stabilizability directly from the given AB pair first of all we will compute the eigenvalues of the A matrix, and then only for the eigenvalues which are on the right hand side we will carry out this test, ok.

(Refer Slide Time: 02:17)

Lyapunov Test for Stabilizability



Theorem (Lyapunov test for stabilizability)

The LTI system (A-B) is stabilizable if and only if there is a positive-definite solution P to the following Lyapunov matrix inequality

$$AP + PA' - BB' < 0 \quad / \quad APA' - P - BB' < 0 \quad (\text{LMI})$$

+



The last test is the Lyapunov test for stabilizability. It says that the system is stabilizable if and only if there is a positive definiteness solution P to the following Lyapunov matrix inequality, ok. So, this is in the continuous time and this is in the discrete time domain. If you are typing this pair AP coming from the discrete time system and we term this as an LMI which is a linear matrix inequality.

(Refer Slide Time: 02:51)

Proof: ((LMI) has positive-definite solution $P \implies$ (AB-LT) is stabilizable)



The simplest way to do this is by using the eigenvector test.

Assume that

- 1 (LMI) holds, and
- 2 $x \neq 0$ be an eigenvector of A' associated with the "unstable" eigenvalue λ ; i.e., $A'x = \lambda x$.

Then,

$$x^*(AP + PA')x < x^*BB'x = \|B'x\|^2,$$

where $(\cdot)^*$ denotes the complex conjugate transpose. But the left-hand side of this equation is equal to

$$(A'x^*)'Px + x^*PA'x = \lambda^*x^*Px + \lambda x^*Px = 2\operatorname{Re}[\lambda]x^*Px.$$

Since P is positive-definite and $\operatorname{Re}[\lambda] \geq 0$, we conclude that

$$0 \leq 2\operatorname{Re}[\lambda]x^*Px < \|B'x\|^2,$$

and therefore x must not belong to the kernel of B' .



So, let us see the quick proof for this result. So, the first implication in that the LMI has positive definite solution matrix P which implies that the pair AB is stabilizable. So, again we would be using this eigenvector test to prove this part. So, there are two assumptions here which the first one is already given here that the LMI has a positive definite solution P , it means that disc that LMI holds. Ok, the second part is that there is nonzero vector x which happens to be the eigenvector of A transpose associated with the unstable eigenvalue λ that is A transpose x is equal to λx , ok.

So, forming the quadratic form of the left hand side of the Lyapunov equation so, this was we were having on the left hand side and then using that eigenvector. We form this x^* this matrix into x which should be less than or in fact, is less than x^*BB' transpose x and using

the property of the norms I can write this as the squared norm of $b^T x$, ok. So, we have stated we already know that it denotes the complex conjugate transpose.

So, see now proceeding in a similar way what we had done for the Lyapunov test in the controllability, this right hand side part we have represented by A^* , $A^T x^*$ transpose which would be nothing, but your x^* and $AP x$ ok. And on this side since $A^T x^*$ is would be equal to $\lambda^* x^*$. Similarly, here we would have $A^T x^*$ is equal to λx and since λ being scalar I can take him onto the left to the I can compute them which finally, I would have the twice of real part of λ into $x^* P x$.

So, see this part which is pretty much important. So, since P is positive definite and we know that the real part of λ is either greater than or equal to 0. We can conclude that this twice of real part of $\lambda x^* P x$ would definitely be less than of squared norm because squared norm is positive; P is a positive definite already.

So, this part would be positive definite, but since λ we are having onto the right hand side which is also positive, but from this inequality. This linear matrix inequality we would have this less than part and which would always be greater than or equal to 0, ok. And therefore, x must not belong to the kernel of B^T , if x happens to belong to the kernel of B^T we would have $b^T x = 0$ ok, but it is not equal to 0.

(Refer Slide Time: 06:19)

Proof: ((AB-LTI) is stabilizable \implies (LMI) has positive-definite solution P)

We saw earlier that controllability of the pair (A_c, B_c) guarantees the existence of a positive-definite P_c such that

$$A_c P_c + P_c A_c' - B_c B_c' = -Q_c < 0.$$

On the other hand, since A_u is a stability matrix, we conclude from the **Lyapunov stability theorem** that there exists a positive-definite matrix P_u such that

$$A_u P_u + P_u A_u' = -Q_u < 0.$$

Defining

$$\bar{P} = \begin{bmatrix} P_c & 0 \\ 0 & \rho P_u \end{bmatrix}$$

for some scalar $\rho > 0$ to be determined shortly, we conclude that

$$\begin{aligned} \bar{A}\bar{P} + \bar{P}\bar{A}' - \bar{B}\bar{B}' &= \begin{bmatrix} A_c & A_{12} \\ 0 & A_u \end{bmatrix} \begin{bmatrix} P_c & 0 \\ 0 & \rho P_u \end{bmatrix} + \begin{bmatrix} P_c & 0 \\ 0 & \rho P_u \end{bmatrix} \begin{bmatrix} A_c' & 0 \\ A_{12}' & A_u' \end{bmatrix} - \begin{bmatrix} B_c \\ 0 \end{bmatrix} \begin{bmatrix} B_c' & 0 \end{bmatrix} \\ &= - \begin{bmatrix} Q_c & -\rho A_{12} P_u \\ -\rho P_u A_{12}' & \rho Q_u \end{bmatrix}. \end{aligned}$$



The other direction of this implication that the AB is stabilizable implies that LMI has a positive definite solution. So, we already know that the here AB is stabilizable and if AB is stabilizable we can decompose it into the controllable part and the uncontrollable part. And stabilizable meaning to say that all the eigenvalues are on the left hand side.

So, we will go step by step. So, first of all we from the controllability of the pair A_c and B_c . We have this the Lyapunov equation in the context of the controllability this is one of the results where Q_c , where we have this $B_c B_c$ transpose and PC is a positive definite solution satisfying this equation.

Now, on the other hand since A_u is a stability matrix. So, it is already assumed in this implication that the pair AB is stabilizable. So, all the eigenvalues are already onto the left hand side and A_u is a stability matrix.

Now, using the Lyapunov stability theorem we can write this simply we can write this LMI, ok in terms of A_u and P_u and P_u is a positive definite solution of this equation. So, this equation we basically have written from the stability result and this equation we have result from the controllability result with different $A_c P_c$ and $A_u P_u$ pair, ok.

So, now we defined another matrix \bar{P} which is given by the elements P_c and row P_u on the diagonals and 0 on the half diagonal so, where ρ is some positive scalar or parameter. So, let us see this part for the complete equation which is in the result. In fact, ok

(Refer Slide Time: 08:29)

Proof: ((AB-LTI) is stabilizable \implies (LMI) has positive-definite solution P)

We saw earlier that controllability of the pair (A_c, B_c) guarantees the existence of a positive-definite P_c such that

$$A_c P_c + P_c A_c' - B_c B_c' = -Q_c < 0.$$

On the other hand, since A_u is a stability matrix, we conclude from the **Lyapunov stability theorem** that there exists a positive-definite matrix P_u such that

$$A_u P_u + P_u A_u' = -Q_u < 0.$$

Defining

$$\bar{P} = \begin{bmatrix} P_c & 0 \\ 0 & \rho P_u \end{bmatrix}$$

for some scalar $\rho > 0$ to be determined shortly, we conclude that

$$\begin{aligned} \bar{A}\bar{P} + \bar{P}\bar{A}' - \bar{B}\bar{B}' &= \begin{bmatrix} A_c & A_{12} \\ 0 & A_u \end{bmatrix} \begin{bmatrix} P_c & 0 \\ 0 & \rho P_u \end{bmatrix} + \begin{bmatrix} P_c & 0 \\ 0 & \rho P_u \end{bmatrix} \begin{bmatrix} A_c' & A_{12}' \\ A_{12}' & A_u' \end{bmatrix} - \begin{bmatrix} B_c \\ 0 \end{bmatrix} \begin{bmatrix} B_c' & 0 \end{bmatrix} \\ &= - \begin{bmatrix} Q_c & -\rho A_{12} P_u \\ -\rho P_u A_{12}' & \rho Q_u \end{bmatrix}. \end{aligned}$$

It turns out that by making ρ positive, but sufficiently small, the right-hand side can be made negative-definite. The proof is completed by verifying that the matrix.

$$P = T \begin{bmatrix} P_c & 0 \\ 0 & \rho P_u \end{bmatrix} T'$$

satisfies (LMI).



So, this is in the result. So, we need to show that this element has a positive definite solution P , ok. So, let us take the left hand side of that equation and start putting those matrices what we had obtained \bar{A} is basically this one if I write all these matrices explicitly, \bar{P} we are introducing we are defining this matrix \bar{P} .

And similarly \bar{B} I can write this Bc . Now simplifying them I would have $A c$ into $P c$ plus $P c$ into $A c$ transpose minus $B c$ into $B c$ transpose which is nothing, but minus of $Q c$. So, here I would have $Q c$ and I taken the minus sign outside of this matrix, ok.

So, similarly see this part we would have A_{12} row $P u$ here and the rest would be 0 and with the negative sign because we have taken the negative sign outside. Another half diagonal element we would have $A u$ rho into P sorry, this one. So, here we would have this would be 0 and this would be rho $P u$ into A transpose 12 , ok and seeing the last element we would have rho is a scalar. So, I can write $A u P u$ plus $P u$ into $A u$ transpose which is nothing, but minus $Q u$. So, I obtained this matrix with rho multiplication.

Now in order to see this matrix, this matrix first of all is a symmetric matrix because the off diagonal elements are equal or let us say in fact, if I take the transpose of this matrix I would obtain in fact, this equivalent, ok.

Now, the second part $Q c$ we have defined already is a positive definite matrix, ok. $Q u$ which we have obtained from the uncontrollable part is a positive definite matrix, ok. Now, I can choose rho a very small value which is already a positive such that the diagonal elements would be positive. If it is a positive and it is symmetric then the entire matrix would be a positive definite matrix, and this positive definite matrix can be written which was given in the main result being this matrix a positive definite matrix we would have the right hand side the negative definite, ok.

Now, since this \bar{P} is the transformed matrix of the transformed system. So, in order to obtain the P matrix of the original pair $A B$, again I using this T transformation matrix I could

obtain this P and you can verify that this P matrix satisfy that linear matrix inequality which is given in the result, ok. So, this completes the proof.

(Refer Slide Time: 12:07)

Controllability after sampling

Consider a continuous-time state equation

$$\dot{x} = Ax + Bu \quad (\text{AB-LTI})$$


If the input is piecewise constant or

$$u(k) := u(kT) = u(t) \quad \text{for } kT \leq t < (k+1)T$$




then the equation can be described by

$$\bar{x}(k+1) = \bar{A}\bar{x}(k) + \bar{B}u(k) \quad + \quad (\text{C2D})$$

with $\bar{A} = e^{AT}$, $\bar{B} = \left(\int_0^T e^{At} dt\right) B \equiv MB$.

 If (AB-LTI) is controllable, will its sampled equation (C2D) is controllable?

This problem is important in designing so-called dead-beat sampled-data systems and in computer control of continuous-time systems.

So, additional result we have on the controllability. That so far we have; so, in the first week we discussed the continuous to discrete time transformation given a continuous time system. So, we need to sample our system sampling time T to finally, obtain this discrete time system if the discrete time system is already given to us, ok. Then we had studied different test to carry out whether the system is controllable or whether the system is stabilizable.

Now, we had separate test for the continuous time system also, but now the here the problem is if I use the continuous to discrete transformation given a continuous time system. So, what can I comment on to it controllability, ok. So, if you recall from the first week. So, we had studied two ways of doing the discretization; the first one is using the Euler method and the

second one by assuming that u_k is some piecewise constant signal between this time interval, ok.

So, but when we assume that u as piecewise constant given by this, one we obtain this discrete time representation where \bar{A} and \bar{B} . We have explicitly computing using this \bar{A} is given by $e^{A T}$ to the power A into the sampling time and \bar{B} is given by M into B , where m is the this part, ok.

So, now the problem here is which we want to address that if the pair AB is controllable will its sampled equation which is given by after doing the C2D continuous discrete conversion is controllable, ok. So, we will see the answer to this problem. In fact, this problem is quite important in designing so called dead-beat sampled-data systems in the computer control of continuous-time systems.

(Refer Slide Time: 14:23)

Controllability after sampling



Let λ_i and $\bar{\lambda}_i$ be, respectively, the eigenvalues of A and \bar{A} .

Theorem (Controllability after sampling)

Suppose $(AB-LTI)$ is controllable. A sufficient condition for $(C2D)$, with sampling time T , to be controllable is that

$$|\text{Imag}(\lambda_i - \lambda_j)| \neq \frac{2\pi m}{T}, \quad m = 1, 2, \dots$$

whenever $\text{Re}(\lambda_i - \lambda_j) = 0$.

For the single input case, the condition is necessary as well.



So, let λ_i and λ_i^* be respectively the eigenvalues of A and A^* . So, here note that when. So, this A^* I should have written this A_d which is the discrete time matrix of the continuous time matrix. So, here A^* is not the transform matrix algebraically equivalent transform matrix. So, whenever I am speaking of this A^* and the context of the controllability after sampling, this A^* is the discrete time counter part of the continuous time matrix A . So, λ_i is the eigenvalue of this and λ_i^* is the eigen value of A^* .

So, this is the result like suppose that the continuous time pair AB is controllable. So, a sufficient condition for the discretized equation with sampling time T to be controllable is that, that the imaginary part of the difference of two eigenvalues λ_i . And, λ_j of the continuous time system is not equal to $2\pi m/T$ where m is a positive integer, whenever the real part of the difference of two eigenvalues is 0, ok.

So, for the single input case this condition is necessary as well, but for the multiple input case this condition is only a sufficient condition, ok. So, we would not be going through the proof of this result, but we will see various implication of this result.

So, first of all notice here that if the eigenvalues of the A matrix are only on the lets say what would be the possible cases. So, this is my S plane, now if all the eigenvalues are on to the real axis, ok. Let us say we have the eigenvalues here or it we have the eigenvalues anywhere, ok. If we have the eigenvalues here then it means that the system would always be controllable right because there would not be any binary part, ok.

Now, this imaginary condition is for those eigenvalues which have a complex conjugate let us say I have one eigenvalue here and one eigenvalue here or one eigenvalue here and another eigenvalue here, right. Because only in this and in this case we could have the difference, the real part of the difference of two eigenvalues equal to 0, right. And in only those cases we would check the imaginary part of that difference, if that imaginary part of the difference is not equal to $2\pi m/T$ where T is a sampling time then we say that the system is the discrete time system is also controllable, ok.

(Refer Slide Time: 17:43)

Controllability after sampling



Let λ_i and $\bar{\lambda}_i$ be, respectively, the eigenvalues of A and \bar{A} .

Theorem (Controllability after sampling)

Suppose (AB-LTI) is controllable. A sufficient condition for (C2D), with sampling time T , to be controllable is that

$$|\text{Imag}(\lambda_i - \lambda_j)| \neq \frac{2\pi m}{T}, \quad m = 1, 2, \dots$$

whenever $\text{Re}(\lambda_i - \lambda_j) = 0$.

For the single input case, the condition is necessary as well.

It is straightforward to verify that if A has only real eigenvalues, then the discretized equation with any sampling period $T > 0$ is always controllable.



So, it is straightforward to verify that if the matrix A has only real eigenvalues then the discretized equation with any sampling period T greater than 0 is always controllable, right.

(Refer Slide Time: 18:01)

Controllability after sampling

Further remarks:

Suppose A has complex conjugate eigenvalues $\alpha \pm j\beta$.

- If the sampling period T does not equal any integer multiple of π/β , then the discretized state equation is controllable.
- If $T = m\pi/\beta$ for some integer m , then the discretized equation may not be controllable.

$$\begin{aligned} 2\beta &\neq \frac{2\pi m}{T} \\ 2\beta &\neq \frac{2\pi m}{T} \\ T &\neq \frac{\pi}{\beta}, m \neq 1 \end{aligned}$$



So, let us see we have complex conjugate pair of the matrix A . Now, if the sampling period T does not equal to any integer multiple of π by β . Let us recall which is the imaginary part of λ_i minus λ_j should not be equal to $2\pi m$ by T .

Now, the difference here the imaginary part we would have twice β from here and this should not be equal to $2\pi m$ by T which is equivalent to same that t is not equal to π by β or for m is equal to 1, ok. So, if it is not equal to the integer multiple of π by β then the discretized state equation is controllable, ok. But if it becomes equal to π by β for some integer m then it only says that the, for the discretized equation may not be controllable. this is the because of that this is only a sufficient condition.

So, if it satisfies, if the results satisfy it means that the discrete time version is also controllable, if it does not satisfy it does not mean that the system is not controllable, ok; it means that the system may not be controllable or be controllable.

(Refer Slide Time: 19:35)

Controllability after sampling



Further remarks:
Suppose A has complex conjugate eigenvalues $\alpha \pm j\beta$.


- If the sampling period T does not equal any integer multiple of π/β , then the discretized state equation is controllable.
- If $T = m\pi/\beta$ for some integer m , then the discretized equation *may not* be controllable.

Note (take it as an exercise)
Since $\bar{A} = e^{AT}$, if λ_i is an eigenvalue of A , then $\bar{\lambda}_i := e^{\lambda_i T}$ is an eigenvalue of \bar{A} .

If $T = m\pi/\beta$ the two distinct eigenvalues $\lambda_1 = \alpha + j\beta$ and $\lambda_2 = \alpha - j\beta$ of A become a repeated eigenvalue $-e^{\alpha T}$ or $e^{\alpha T}$ of \bar{A} . This will cause the discretized equation to be uncontrollable.

$\lambda = \alpha + j\beta$, $e^{(\alpha + j\beta)T} = e^{\alpha T} \cdot e^{j\beta T}$
 $= e^{\alpha T} \cdot e^{jm\pi}$
 $= e^{\alpha T} \cdot (-1)^m$



So, we can see the implication of the second point more precisely. So, first of all is note that since \bar{A} the discrete time state matrix is equal to e^{AT} where A is the continuous time matrix and T is the sample time. If λ_i is an eigenvalue of A then $\bar{\lambda}_i$ which is equal to $e^{\lambda_i T}$ is an eigenvalue of \bar{A} , ok. So, you can take this as an exercise if you have any difficulty in visualizing this because both the matrices A and \bar{A} are related by this one. So, the λ_i , the eigen values are also written in a similar way.

So, the see if T becomes equal to $m\pi/\beta$ for some integer m or and given these two distinct eigenvalues $\alpha + j\beta$ and $\alpha - j\beta$, ok. In fact, become a repeated eigenvalue either of $e^{-\alpha T}$ or $e^{\alpha T}$ of \bar{A} , ok. So, we can quickly visualize this say for example, take λ is equal to $\alpha + j\beta$. So, the eigen value of the discrete time would be $e^{\alpha T + j\beta T}$ into the sample time.

So, this i can write as $e^{\alpha T} e^{j\beta T}$, sorry it should be $j e^{\alpha T}$ times $e^{\alpha T} e^{j\beta T}$, and T is already known is a multiple of $m\pi/\beta$. So, this would become $e^{\alpha T} e^{jm\pi}$ and this part is nothing, but either it would be plus 1 or it would be minus 1. So, that is why it becomes a repeated eigenvalue of the discrete time state matrix and this will cause the discrete time equation to be uncontrollable.

(Refer Slide Time: 22:09)

Controllability after sampling

Further remarks:

Suppose A has complex conjugate eigenvalues $\alpha \pm j\beta$.

- If the sampling period T does not equal any integer multiple of π/β , then the discretized state equation is controllable.
- If $T = m\pi/\beta$ for some integer m , then the discretized equation may not be controllable.

Note (take it as an exercise)

Since $\bar{A} = e^{AT}$, if λ_i is an eigenvalue of A , then $\bar{\lambda}_i := e^{\lambda_i T}$ is an eigenvalue of \bar{A} .

If $T = m\pi/\beta$ the two distinct eigenvalues $\lambda_1 = \alpha + j\beta$ and $\lambda_2 = \alpha - j\beta$ of A become a repeated eigenvalue $-e^{\alpha T}$ or $e^{\alpha T}$ of \bar{A} . This will cause the discretized equation to be uncontrollable.

Theorem

If the continuous time LTI state equation is not controllable, then its discretized state equation with any sampling period, is not controllable.



So, the last result we have that if the continuous time system is not controllable then its discrete state equation with any sampling period will never be controllable matrix, right. If the system the LTI system is controllable then we had a test or we, you could compute the sampling time also; so, that is your discrete time version of the state equation becomes controllable. Now, in the original pair is not controllable then for any sampling period your discrete time version would never be a controllable.