

Analog Electronic Circuits
Prof. Pradip Mandal
Department of Electronics and Electrical Communication Engineering
Indian Institute of Technology, Kharagpur

Lecture – 78
Differential Amplifier: Analysis and Numerical Examples (Contd.)

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Small signal analysis of differential amplifier realized by MOSFETs
Common mode of stimulus:

The slide contains several diagrams and equations. On the left, there is a schematic of a differential amplifier with inputs $v_{in,c}$ and outputs v_{o1} and v_{o2} . In the center, a small-signal equivalent circuit is shown with two MOSFETs, their gates tied together to a common-mode input $v_{in,c}$, and their sources connected to a common source resistor of $2R_T$. The drains are connected to load resistors R_{D1} and R_{D2} . The outputs are v_{o1} and v_{o2} . Handwritten notes include: $v_{o,c} = \frac{v_{o1} + v_{o2}}{2} = -v_{in,c} \frac{R_{D1} || R_{D2}}{2R_T}$; $v_{o,c} = -v_{in,c} \frac{R_{D2}}{2R_T}$; and $v_{o2} = v_{in,c} \frac{g_{m1} R_{D1}}{1 + g_{m1} R_T}$. On the bottom left, there is a diagram of a common source (CS) stage with a source resistor $2R_T$ and a gate input $v_{in,c}$.

Yeah. So, welcome back after the short break. So, we are talking about the common mode stimulus. And let us see what happens to the circuit, when we stimulate the circuit with identical signal at the 2 inputs. And so, here we do have the small signal equivalent circuit and here, we like to feed the signal small signal. So, $v_{in,1}$ equals to $v_{in,c}$. So, same thing same signal we are feeding here at the other input.

So, $v_{in,2}$ equals to $v_{in,c}$. Now for our understanding of the circuit, again we are keeping the circuit disconnected here. And we like to see what kind of signal we do get with this

stimulus. So, if we are keeping this is disconnected and if you refer to the circuit here, at the transistor level, this is common source amplifier with degenerator, source degenerator. So, this is the source degenerator and we know its consequence namely the signal coming at its output.

It will be v_{o2} equals to the input v_{in} ; with a minus sign here and then $g_{M1} R_{D1}$ divided by $1 + g_{M1} r_{e2}$. Or you can approximate this by minus $v_{in} g_{M1} R_{D1}$. In fact, this one part you can remove. So, we can simply consider g_{M1} and r_{e2} so, this g_{M1} and this g_{M1} this getting cancelled. So, same thing for the other output, namely v_{o1} equals to again minus sign, the corresponding input multiplied by R_{D1} .

So, this is R_{D1} , this is R_{D2} and then $2 R_{T}$. So, note that the signal here and the signal here they are identical. Now if I consider on the other hand the signal at the emitter, if I consider the signal at the emitter it is working as similar to our previous discussion. Before we connect the resistor and these two resistors; we do have signal coming here very close to the applied input voltage, which is v_{in} and then thevenin equivalent resistance it is approximately $1/g_{M1}$.

So, likewise if I consider the other side, this side and what we get it is similar kind of equivalent circuit. Namely, the signal source, which is close to v_{in} and then, thevenin equivalent resistance of $1/g_{M2}$ and note that unlike for differential case since these two signals the applied signal at the 2 inputs they are in phase then this signal and this signal they are in phase.

Now, even if we make this connection since these two signals they are identical there will not be any change. So, the signal coming at this point it will be v_{in} . So, before we make this connection whatever the signal we are having here and here they are remaining same even if you make this connection. And that is mainly because the input signal it is the applied input signal they are identical and they are in phase. So, I should say that the common mode output if I say, if I take average of v_{o2} and this v_{o1} .

So, that gives us the common mode output $v_{o.c}$. So, that is v_{o1} plus v_{o2} by 2. So, that is remaining same as individual one namely $v_{in.c}$ multiplied by R_D divided by $2R_T$. In fact, that gives us the common mode gain common mode gain A_c defined as of course, with a minus sign do you have a minus sign here. So, A_c is which is defined as $v_{o.c}$ divided by $v_{in.c}$ equals to minus R_D by $2R_T$ ok.

So, that is that is the common mode gain. In fact, similar thing you can get for the other differential amplifier realized by BJT.

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Small signal analysis of differential amplifier realized by BJTs
Common mode of stimulus:

The slide contains the following equations and labels:

- Common-mode gain: $A_c = -\frac{R_c}{2R_T}$
- Output voltages: $v_{o2} = \frac{g_m R_{c1} v_{in.c}}{1 + g_m 2R_T}$ and $v_{o1} = -\frac{g_m R_{c2} v_{in.c}}{1 + g_m 2R_T}$
- Common-mode output: $v_{o.c} = -\frac{g_m R_c v_{in.c}}{1 + g_m 2R_T} \approx -\frac{R_c v_{in.c}}{2R_T}$

The circuit diagram shows two BJTs with a common emitter resistor of $2R_T$ and collector resistors R_{c1} and R_{c2} . The common-mode input $v_{in.c}$ is applied to both bases. The simplified equivalent circuit shows the common emitter resistor as $2R_T$ and the collector load as R_c .

So, in the next slide we do have the corresponding circuit and here again even though we do have small change small difference, in the circuit namely we do have r_{π} and r_{π} here. But all

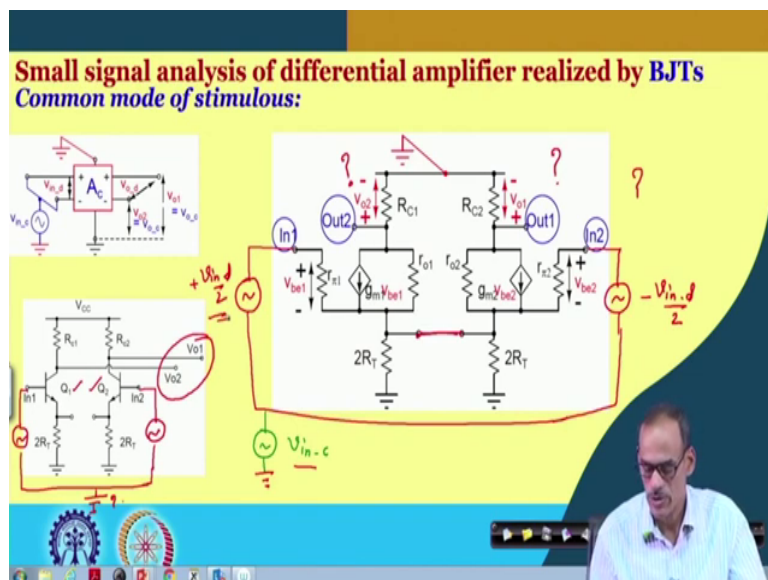
practical purposes when you stimulate the circuit with identical signal here $v_{in,c}$ at input 1 and input 2.

Then the signal coming here and here they are identical. And the signal at then $v_{o,2}$ equals to minus yeah, minus $g_{M1} R_{c1}$ divided by $1 + g_{M1} R_{T1}$ into $v_{in,c}$. And same thing we will be getting here also namely $v_{o,1}$ is equal to minus $g_{M2} R_{c2}$ divided by $1 + g_{M2} R_{T2}$ into $v_{in,c}$. Now since the signal here and here they are in phase even if we make this connection the signal coming here it will be remaining unchanged.

Namely this will be approximately equal to $v_{in,c}$ and as a result these two output after even after making this connection they are remaining unchanged. And hence $v_{o,c}$ it is equal to approximately $g_{M1} R_{c1}$ divided by $1 + g_{M1} R_{T1}$ into $v_{in,c}$ with a minus sign. So, you may ignore this one and then you can remove this g_{M1} part. So, that will be equal to minus R_{c1} divided by $2 R_{T1}$ into $v_{in,c}$. In fact, that gives us the same expression of common mode gain equals to minus R_{c1} divided by $2 R_{T1}$ right.

So, in summary we got the expression of the common mode gain and differential mode gain, whenever we will be going into numerical circuit then we will see their corresponding values. Now so, far we are talking about the small signal situation. Now you may recall that while we are applying the signal at the input definitely we are also applying a meaningful DC.

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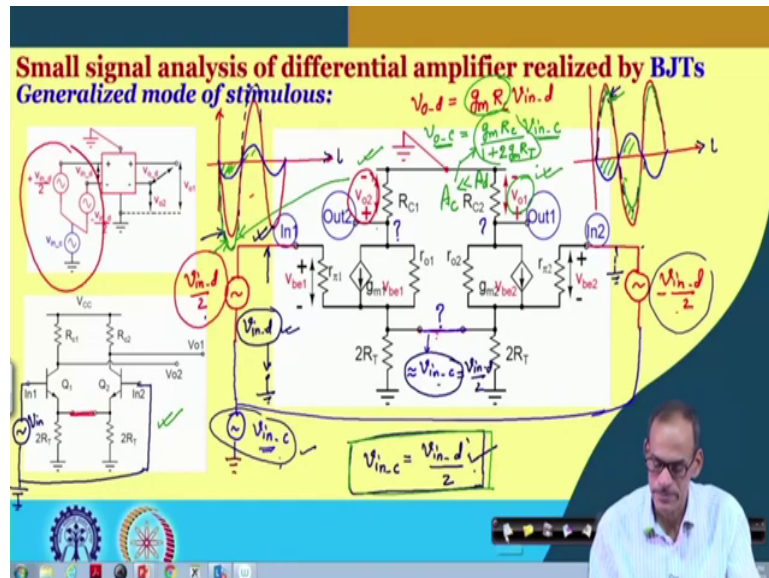
So, now next question is that what may be the meaningful DC quote and unquote meaningful DC? That can be that can be analysed by considering large signal behaviour of this entire circuit. In fact, not only this voltage, but also we like to know what may be the DC voltage coming at the 2 outputs. And we like to see whether the two transistors really in good condition or not. So, to understand that, we need to have large signal analysis ok.

So, before we go into the large signal analysis of course, so, I need to say one more thing that. So, far we are we are talking about small signal one at a time, namely differential part and then the common mode part. But then in case you have say both the signals coming together namely if $v_{in,c}$ and the differential part namely $+v_{in,d}/2$ and $-v_{in,d}/2$.

If they are coming together, then what happens? So, naturally before we make this connection the situation it was something, but again instead of repeating that I may directly get into this

connection. And then we like to see what may be the condition here and here. So, before you go for large signal analysis, let me do one more small signal analysis, where we are considering both the common mode and differential mode signal together. And we like to see the corresponding output.

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So, in the next slide we do have the generalized mode of stimulus. So, as I said that at the input we like to give both differential and common mode part together. And probably we can make this connection we can make this connection. Now let us see if we apply a differential part here $v_{in,d}$ by 2. So, here we do have $v_{in,d}$ by 2 and then we do have on the common mode part, let me use blue colour. So, $v_{in,c}$, which is going to going to the to both the inputs.

So, it is going here as well as here. Now let us see once we make this connection then what may be the signal here what may be the signal here and what may be the signal here. Now due

to due to this perfect differential sorry this is perfect differential; that means, we do have a minus sign here. So, these two differential signal is not having any influence; in fact, for differential component this is this supposed to be ground, but then we do have the common mode component.

So, the signal coming here it is only $v_{in c}$; I should say very close to $v_{in c}$ it is slightly less, but it is very close to that. And then what happens to this output, we do have we do have the this signal it is coming in amplified form. So, this differential part it is it is showing its effect it is in opposite phase. So, likewise here we do have the effect of differential part, which is which is in phase of the differential input ok.

And then we do have the effect of the common mode part which is of course, having smaller amplitude and. So, here we do have the common mode input which is again getting flipped. Note that for our simplicity we have considered both the signals are having the same frequency. So, and then if that is the case then at the 2 output what will get in the combined effect of these two it may be something like this.

So, this is the net output we do get at this point. And if I consider the other input it will be having the corresponding signal like this. So, you can see that this amplitude it is slightly higher, because of the common mode component is coming there and this signal it is slightly less than the red colour. So, we can say that at the output we do have both differential as well as common mode part they are present.

So, $v_{o d}$ if I take the difference of these two signal will be getting $g_M \cdot 1$ or g_M I should say g_M into R_c into $v_{in d}$. And if I take average of these two signals average of say this signal and this signal then what will be getting is only this part; because they are getting added up and the red portion it is getting cancelled out. So, we can say that $v_{o c}$ if I take average of the 2 signal then $v_{o c}$ equals to g_M into R_c into $v_{in c}$.

So, that is how the circuit it operates; now as a special case if you consider that let you consider a special case where you like to give a 0 signal here; that means, if I apply say $v_{in c}$ equals to $v_{in d}$ by 2. So, what happens? So, in this case, this part if it is $v_{in d}$ by 2 and then,

we do have minus v in d by 2 ; so, together that gives us a 0 signal here on the other hand if I am having this signal equals to v in d by 2 . So, this v in d by 2 and this v in d by 2 together it gives us v in d .

So, the signal at this point with respect to ground it is v in d by 2 . So, such kind of stimulus namely if we have a signal only at one side and then other side it is say ground, it is referred as pseudo differential stimulus. Many a times suppose you do have an amplifier like this we like to give signal only at one end and then of course, along with the DC. While at the other input we like to give only this DC and the signal here it is whatever the complete signal you may say that this v in incidentally that is this v in d .

For such case; obviously, you will be getting a signal here which is v in c and this signal it is if you see here it is v in d by 2 . And the corresponding signal here because the common mode gain and differential mode gain they are quite different sorry, I have committed a mistake here. $v_o c$ it will be v in c multiplied by g_M into R_c divided by $1 + 2 g_M$ into R_T sorry yeah. So, since this gain is low that is why we are getting smaller signal. So, now, in this pseudo differential case, whenever we are considering, this common mode signal it is half of the differential signal.

Then since this signal it is small the in even in this case the net voltage which is this v_o 2 and the net voltage here. So, this net voltage here which is v_o 1 ; since this part it is very small. So, we can approximate that this output and this output they are almost like a differential. So, based on the logic if the common mode gain this A_c it is very small compared to this A_d .

Then you can say that even if we stimulate the circuit in this pseudo differential form, the corresponding output there very close to like a differential operation. And that is why it is referred as pseudo differential mode of operation right. And now, we are in a position to go for the large signal analysis.

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Large signal analysis of differential amplifier realized by BJTs

D.C. operating point

$$V_C = V_{a,DC} = V_{CC} - R_C \cdot \frac{I_{R_T}}{2}$$

$$I_C = f(V_{INC})$$

$A_{c,d}, A_{d,c} = 0$

Q_1, Q_2

$R_{C1} = R_{C2}$

$V_B = \frac{V_{INC} - V_{BE(Q_1)}}{(Q_1 + Q_2)}$

$I_{R_T} = \frac{V_{INC} - V_{BE(Q_1)}}{R_T}$

So, in the next slide we are going to talk about large signal analysis and we like to see the DC operating point of the amplifier.

So, again going back to the circuit here, the basic model here, where we do have the differential amplifier; which is getting stimulated by a pair of signal accompanying same amount of DC voltage. And here we do have the corresponding implementation. Now we are going to talk about what is the role of this DC voltage and what may be the range of this DC voltage and if we vary this DC voltage what may be the situation at the output.

Now, since this DC voltage it is applied to both the input terminal, to understand this circuit operation we do not require to split this circuit. In fact, we can keep the circuit like this. And at the two inputs so, at both the inputs we shall apply a DC voltage called V_{inC} and then we

will we can observe the corresponding DC voltage here and DC voltage here. So, we can vary this voltage and then we can see what it may happen to DC voltage here as well as here.

Now, we have, so, we do not require this kind of split. So, we are considering this is connected and hence our subsequent discussion it will be with this circuit we may not be going back to the split one. Now coming back to what we said is that, if I vary this voltage and then if we observe the voltage at the two outputs, definitely they are also defining the condition of transistor 1 and transistor 2. And for good operation of the circuit we want both the transistor should be in active region of operation.

So, the range of this voltage it should be such that both the transistor should be in active region of operation. Now we also have said that to make A_{cd} and A_{dc} to 0 we say that Q_1 and Q_2 are identical and R_{C1} equals to R_{C2} right. And since they are identical probably, we can consider the entire circuit together. And in fact, since these two are identical and these two are identical, we can say that voltage here and voltage here they will be same.

So, we can simply consider it is a folded circuit where Q_1 and Q_2 we can overlap together and then we do have R_T here. So, likewise we do have R_{C1} and R_{C2} they are coming in parallel. And here you do have Q_1 and Q_2 they are coming in parallel. And the voltage you are applying here it is this V_{INC} input common mode voltage. Now, if you if you analyze this circuit of course, depending on this voltage will be getting a voltage here and then that will define this current and that current it is flowing here.

So, we do have a DC supply voltage V_c . So, we can make a drop of this voltage and that gives us the corresponding collector voltage. In fact, if you if you go step by step to that for a given V_{INC} so, what is the emitter voltage. So, emitter voltage at this node or in this where you say merged circuit the emitter voltage equals to V_{INC} minus V_{BE} on. So, it may be around 0.6 or 0.3 depending on the transistors material.

So, that gives us the voltage here and hence the current flowing through this R_T equals to V_{INC} minus V_{BE} on divided by R_T . In fact, this current it is flowing through both Q_1 and Q_2 . So, here also we can say this is I_{RT} which is summation of the two emitter currents and if I

say that both of the transistors are identical. So, we can say that half of this current is emitter current and practically they are defining the corresponding collector current.

So, the voltage at the two output nodes, if I say that $V_{O\ DC}$ equals to V_{CC} minus R_C ; $R_C\ 1$ or $R_C\ 2$ both are same this multiplied by I_{RT} by 2. So, that gives us the corresponding collector voltage and we do have the emitter voltage. So, now, we have to apply a meaningful voltage here. So, that both the transistors should be in active region of operation, not only they should be in active region of operation it should be having sufficient the voltage here should be having sufficient room for the signal.

So, that we can have good amount of signal amplitude it should be able to accommodate. So, the range over which this resists this voltage it is allowed, it is referred as common mode range. And in this case of course, it depends on the corresponding value of R_T and R_C . So, whenever we will be talking about numerical examples, where we will be having value of this R_T and R_C . And the supply voltage there we shall see the upper limit and lower limit of this input common mode voltage.

Only thing is that qualitatively. So, these are the expressions, but qualitatively I must say that it is having a good range. So, it is not necessary that we need to have very precise DC voltage here for proper operation, it will be having a good range over which the circuit it will be working fine. Only thing is that depending on this value of this V_{INC} the corresponding collector current it may vary.

So, I should say that collector current is strong function of V_{INC} and hence all the small signal parameters namely g_M and r_{not} and so and so, they are strong function of this voltage. But as long as the devices are in active region of operation typically they do not have any problem for proper functionality of the differential amplifier. Now once you get the DC operating point next thing is that what may be the possible signal swing.

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Large signal analysis of differential amplifier realized by BJTs (contd.)

•D.C. operating point
•Input common mode range
•Output swing

So, let us see in the next slide yeah. So, input in fact, we already have discuss this point; the input common mode range. So, the range over which this common DC voltage it is allow to vary. Now next thing is that once we have this V_{INC} and then what may be the range of this voltage over which the transistor both the transistors they are remaining in active region of operation.

So, if you see the range, pictorially suppose we do have the total voltage range. So, this is V_c and this is ground and suppose this input common mode voltage it should be higher than 0.6 or V_{be} . So, that at least we will be getting a meaningful voltage here.

So, V_{INC} it should be higher than say 0.6 and above. So, that is the lower limit of this and then for a given V_{INC} , we do have some current flow here and then of course, there will be some higher drop. So, the corresponding DC voltage here it may be somewhere here. Now

this is $V_{O\ DC}$ now this DC voltage it can go as high as or it may go close to the V_c . On the other hand it can go as low as suppose we do have some V_{INC} given to us. So, V_{INC} we are expecting this should be higher than this lower limit.

So, suppose this V_{INC} it is given to us then the V_O individual output V_{O1} and V_{O2} with respect to this DC it can go as high as towards the V_c or it can go as low as towards the V_{INC} ; maybe with a margin of 0.2 or 0.3. So, whatever the range we are talking here, this side and this side. So, that is referred as the possible signal swing. Now for a given input DC voltage, we want this DC voltage should be towards the middle of it. So, that the positive swing of the output signal and negative swing of the output signal they should be equal.

So, that gives us the maximum possible not only maximum possible peak to peak voltage, but also that will be helping us to get the maximum amplitude of the sinusoidal signal right. So, again as I said that whenever will be going through some numerical examples, we shall explain little detail of how to pick the right value of the resistor and this resistor. So, similar to BJT for MOSFET circuit also we do have similar kind of situation.

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Large signal analysis of differential amplifier realized by MOSFETs

Key Equations:

- $V_{0-DC} = V_{DD} - R_D I_{RT}$
- $R_{D1} || R_{D2} = \frac{R_D}{2}$
- $V_{INC} = V_{GS_{1,2}} + I_{RT} R_T$
- $I_{RT} = \left(\frac{K W}{L} \right)_{M_1+M_2} \times \frac{1}{2} (V_{GS_{1,2}} - V_{th})^2$

Key Points:

- D.C. operating point
- Input common mode range
- Output swing

Conditions:

- $V_{INC} > V_{th}$

Namely the DC operating point, it is a strong function of the input common mode voltage.

So, suppose we do have input common mode voltage coming to the gate or transistor 1 and transistor 2 and that essentially so, this is V_{INC} . So, that defines the gate voltage here now again here also we are assuming these two transistors they are identical and these two resistors they are identical. And since both the gate nodes they are getting the same voltage probably we can fold it and we can equivalently say that we do have a single chain. Where we do have M_1 and M_2 they are connected in parallel way and then here we do have V_{INC} .

And then we do have we do have R_T here and then we do have 2 R_D s are in parallel. In fact, if they are equal you may say that simply R_D by 2. And then here we do have the V_{DD} . Now for this circuit, if you may recall if we analyze this loop; then if the parameter of the transistors

are given namely threshold voltage and transconductance factor. So, from that we can find what will be the corresponding current flow here.

For a given value of this voltage, so, if you analyze this circuit what we can say that suppose this circuit the combined resistors they do have transconductance factor of say $K_n W/L$ it is given to us. Then it is having V_{GS} and then also if its threshold voltage is given to us. Then we can say that V_{INC} equals to V_{GS} of transistor 1 or 2 plus the total current. So, if I_{RT} this is the current is say I_{RT} .

So, I_{RT} multiplied by R_T . So, this is one equation another equation we can get is that I_{RT} should be equal to this $K_n W/L$ this is combined transistors M_1 plus M_2 together multiplied by half and then the corresponding v_{gs} of transistor 1 2 minus V_{th} square. So, we are dropping $1 + \lambda v_{ds}$ part. So, if you solve say if you consider this equation and this equation and if you solve, then you can find the value of this current flow through this R_T .

And then you can say half of this I_{RT} it is flowing through this resistor and this resistor from that you can find what will be the corresponding drop. And that gives us the DC voltage common DC voltage $V_{O,DC}$ equals to V_{DD} minus R_D multiplied by I_{RT} by 2 right. So, that is how we can get the output DC voltage. In fact, either you consider this circuit where the drop here it is R_D by 2 multiplied by I_{RT} or if you consider half of this circuit where current is half and the resistance is R_D whatever it is. So, we got the expression of output voltage DC output voltage for a given input voltage.

Now, again this voltage it may be having a range and this voltage definitely it should be higher than threshold voltage of the transistor. So, that we can get a positive voltage and hence rather it will be having some current flow here. And that current flow it will be flowing through this resistor keeping these two devices and active condition and preferably the drain voltage and gate voltage should be such that both the transistors should be in saturation region. And not only they should be in saturation region if we have a DC voltage here and DC voltage here it should be such that signal should be having some headroom.

So, the again this is also important aspect, but unless we do have numerical values of the supply voltage and resistors, probably it will be little too hypothetical to analyse. So, once will be going through numerical examples, then we will be talking about the calculation of input common mode range and DC operating point and then the corresponding output signal swing. Now at the input at the input.

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Large signal analysis of differential amplifier realized by MOSFETs

• D.C. operating point
 • Input common mode range
 • Output swing

Now, if we apply voltage. So, far we are talking about small signal; now in case if we apply say large signal keeping this voltage constant.

Then what happens? So, if I say that it is perfectly differential namely this is v_{in-d} by 2 and this is minus v_{in-d} by 2. So, if I make these two signals 0; obviously, voltage here and voltage here they are same; namely V_{O-DC} and this is of course, V_{INC} . Assuming that v_{in-c} it is having a value which is within its acceptable range and then output voltage it is also having

meaningful value. Now if I increase this voltage subsequently, if I am also increasing this voltage with this minus sign. So, what we are expecting here it is this voltage it is it will increase with respect to its DC.

And this voltage will decrease with respect to the DC. So, if we plot the difference of these two voltages called v_{od} with increase of this $v_{in,d}$. Then what we can get here it is, input to output transfer characteristic, but then input it is in differential form so, is the corresponding output. So, if the input is very small the behaviour here to here it was quite linear. So, this may be linear, but as you are increasing this input voltage beyond some limit then what will happen is, this may enter into non-linear characteristic. Likewise this side also it will be entering into non-linear characteristic.

And this V_{od} it is defined as capital VO_1 minus capital VO_2 ; note that this VO_1 and VO_2 they are representing large signal voltage both of them are having the same DC voltage of VO_{DC} . But then once we subtract it that got removed. So, range of this V_{od} it is in fact, positive and negative this is 0 level and this is positive and this is negative.

So, same thing for v_{in} if I am making these two voltages together, but if we are making this voltage it is higher and higher. Then we means define this, $V_{in,d}$ as capital $V_{in,1}$ minus capital $V_{in,2}$.

And both of them they are having this $V_{in,C,DC}$. So, voltage here it is $V_{in,1}$ capital $V_{in,1}$ and voltage here it is $V_{in,2}$ and both of them are having the common DC and then differential part. Now whenever you are talking about small signal, what we have essentially considered it is that range of this input voltage it is small enough. So, that we are talking in the small range and slope of slope of this characteristic curve it is the gain small signal gain as we increase this voltage beyond certain range.

Then it may enters into the saturated condition. So, whenever we are talking about differential amplifier and if we are restricting the operation within this linear range, then it is it can be treated as a good amplifier having good linearity between input to output. But once you go for and some other application, where circuit is entering into the saturated situation then also it is

having some application. But the circuit will not be called it is a linear circuit. In fact, then circuit it becomes like a comparator.

So, the basic differential amplifier structure can be used for both for amplification purpose for analog application as well as it can be used for comparator. So, later I whenever from the situation permits, we may elaborate on that. But just I like to say that differential amplifier whatever the amplifier we have discussed it is having a specific application not only for analog, but it is also having some application called mixed signal where the output it may be more like a logic signal high or low.

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Conclusion:

- ❑ Small signal equivalent circuits of differential amp.
- ❑ Small signal analysis of diff. amp. for
 - Differential mode of stimulus (operation)
 - Common mode of stimulus (operation)
 - Generalized and pseudo-differential stimulus (operation)
- ❑ Large signal analysis for
 - D.C. operating point → $V_{INC} \rightarrow V_{O-DC} \Rightarrow$ Output Swing
 - Input common mode range and
 - Output signal swing
- ❑ Numerical examples to be covered in next lecture

I think yeah that is all we do have. So, to conclude to summarize what we have discussed today. We started with small signal equivalent circuit for differential amplifier realized in either in BJT or MOSFET version. And then we have talked about small signal analysis for

differential amplifier specifically for three different modes of operation extensively for differential mode and common mode.

And then also we have talked about generalized stimulus and pseudo differential is a special operation. Then we have talked about the large signal analysis, where we mention about the importance of DC operating point and the DC operating point it is a strong function of the input common mode voltage. V_{INC} which also defines the output difference DC voltage and both this input common mode voltage and output DC voltage.

And the range of operation of the transistor they are defining the output swing. So, we did not get a chance to elaborate on these two topics particularly input common mode range and output swing; because it may be difficult to appreciate without any numerical value. So, whenever we will be talking about numerical examples, we shall further elaborate on these parameters. And so, our next discussion it will be numerical examples on differential amplifier will to be covered in the next lecture. I think that is all.

Thank you for listening.