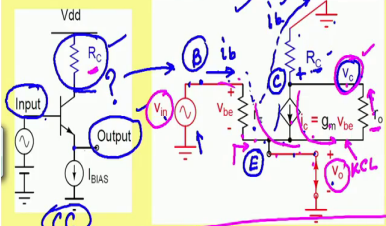


Analog Electronic Circuits
Prof. Pradip Mandal
Department of Electronics and Electrical Communication Engineering
Indian Institute of Technology, Kharagpur

Lecture – 46
Common Collector and Common Drain Amplifiers (Contd.): Analysis (Part B)

(Refer Slide Time: 00:32)

Small signal analysis of more realistic CC amplifier (including R_c)



• Voltage gain:

$$v_{be} = (v_{in} - v_o) \quad v_c = R_c \frac{(v_{in} - v_o)}{r_\pi}$$

$$\frac{(v_o - v_c)}{r_o} = \frac{(v_{in} - v_o)}{r_\pi} + g_m(v_{in} - v_o)$$

$$v_o \left\{ 1 + g_m r_o + \frac{R_c + r_o}{r_\pi} \right\} = v_{in} \left\{ g_m r_o + \frac{R_c + r_o}{r_\pi} \right\}$$

$$v_o = v_{in} \frac{(R_c + r_o + g_m r_\pi r_o)}{r_\pi + R_c + r_o + g_m r_\pi r_o} \approx 1$$

Yeah, welcome back after the short break and we are discussing about the Common Collector Amplifier, considering the; considering the resistance R_c connected in the collector terminal in between collector and supply voltage V_{dd} . So, let us see the circuit, which is the small signal equivalent circuit given here. The input voltage V_{in} , we are applying at the base and then, we do have the collector terminal which is not a C ground rather it may be having a signal called say V_c . So, this is very important change compared to our previous analysis. So, we do have

non-zero value of V_c and then, a rest of the things are same namely at the emitter we are observing the collector sorry, the output current output voltage V_o .

So, the V_{be} ; V_{be} it is V_{in} minus V_o as you have discussed before and but then V_c if you see once we have this output terminal is open, whatever the current it is flowing here, base current that is flowing through this circuit and finally, it is going to the ground here. Because this current need to come back to the source through the ground.

So, we can say that the current flow after reaching to the emitter whether it is branching to the active device or through this r_{naught} ; finally, they are converging to the ground and we can say that this is also same as the base current i_b . So, if I call this is i_b , this is also base current. So, if i_b is flowing through this R_c , the voltage getting developed here it is we call V_c . So, we can say that V_c at this point with respect to ground is R_c multiplied by i_b . On the other hand, i_b is V_{in} minus V_{naught} divided by r_{pi} . So, what we can say that this V_c equals to R_c multiplied by i_b and i_b , it is V_{in} minus V_{naught} divided by r_{pi} .

So, this is what the important thing and that directly affects the current flow through this r_{naught} . So, now, if I apply KCL at the emitter node, what we are getting? Here, it is current flowing through this r_o which is v_o minus v_c divided by r_{naught} . So, that is equal to the summation of the two currents; one is the base current and other is the through the active device. So, the base current is v_{in} minus v_{naught} divided by r_{pi} and the current flowing through this voltage dependent current source, it is g_m into the V_{be} rather minus V_{be} going from going from yeah, the it is basically coming here yeah V_{be} . So, we do have v_{in} minus v_{naught} .

So, summation of this current and this current, it is equal to this current so that is what we have written here. Now, this expression of this v_c , it is in terms of v_{in} and v_o . So, this equation, it can be utilized to replace this V_c as a result we can get an expression which involves only v_o and v_{in} . Now, if we rearrange that equation, what we will be getting is v_o multiplied by 1 plus g_m into r_{naught} plus R_c divided by r_{pi} plus r_{naught} divided by r_{pi}

equals to v_{in} multiplied by g_m into r_{π} plus R_c divided by r_{π} plus r_{π} divided by r_{π} .

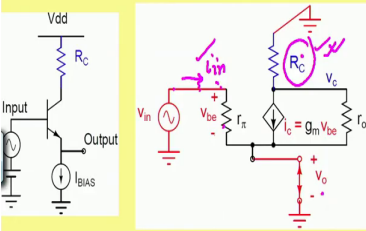
So, from this relationship, between v_{out} and v_{in} that gives us the voltage gain. In fact, we can say that v_{out} equals to v_{in} multiplied by R_c plus r_o plus g_m into r_{π} r_o divided by R_c plus R_c plus r_{π} plus r_o plus g_m into r_{π} into r_o . In fact, if you see here this part it is common here and here. So, effectively you may say that the relationship between v_{in} and v_{out} if I say, it is the impedance here which is r_{π} and the impedance of rest of the circuit there.

Before we connect anything here whatever the impedance, we are having it is R_c plus r_o plus g_m into r_{π} into r_o . So, this part is essentially the impedance at this point, before we connect anything of course, here. So, whatever it is in this case again this part, it is typically dominates and so, we can see compared to r_{π} , definitely this is higher because we do have g_m into r_{π} multiplied by this r_{π} . So, we can say that this part you can ignore and this can be well approximated by 1.

So, the voltage gain even if we have say R_c getting connected here and if we have say the V_c voltage, it is allowed to develop then also the voltage gain it is close to 1. Now, let us see what will happen for the other parameters namely the output resistance and then, input capacitance and so and so and input resistance right. So, we do have let us move to the next slide to do that.

(Refer Slide Time: 07:16)

Small signal analysis of more realistic CC amplifier (including R_C)



• Input resistance:

$$R_{in} = \frac{v_{in}}{i_{in}} = \frac{v_{in}}{(v_{in} - v_o)/r_{\pi}} = \frac{r_{\pi}}{1 - v_o/v_{in}}$$

$$v_o = v_{in} \cdot \frac{(R_C + r_o + g_m \cdot r_{\pi} \cdot r_o)}{(r_{\pi} + R_C + r_o + g_m \cdot r_{\pi} \cdot r_o)}$$

$$R_{in} = (r_{\pi} + R_C) + r_o + (g_m \cdot r_{\pi} \cdot r_o) \rightarrow \infty$$

So, let us concentrate on the input resistance and here, we do have the same small signal equivalent circuit and for input resistance, what we have it is if we are applying V in here whatever the i in it is flowing. If I get the expression of i in in terms of V in that gives us the corresponding input resistance. So, r in it is v in divided by i in; whereas, i in, it is the base terminal current and base terminal current is v in minus v o divided by r pi. So, the expression of i in it is v in minus v o divided r pi. Now, probably you can rearrange this equation namely, we can write in this form r pi divided by 1 minus v naught divided by v in.

So, expression of this v naught by v in which is the voltage gain, we already have seen and that v o divided by v in is essentially this part. So, if I make 1 minus this part, what will be getting here in the denominator it is r pi. It will be getting here r pi and then, below of that will be having this whole thing. So, I should say this whole factor 1 minus vo divided v in, it is

becoming r_{π} divided by the whole thing here and this r_{π} and this r_{π} is getting cancel, we can take this is in the numerator.

So, as a result input resistance what we can get it is this only. So, here we are just summarizing that. Input resistance is r_{π} in series with R_c in series with r_{naught} and then, the active device contribution. Again, because we do have this term coming in series, even though we do have this R_c this is also going to be very high. So, we can consider this is very high. It is quite obvious that since we are considering this resistance in series, it is increasing the overall resistance, only by this part and this part earlier, we obtained that whatever the input impedance we are having.

So, the effect of non-zero value of this R_c , it is increasing this resistance even higher. So, anyway that is in our favour considering the required input port characteristic. So, we have seen the voltage gain, we have seen the input resistance. Now, let us see the other parameter in the next slide.

(Refer Slide Time: 10:22)

Small signal analysis of more realistic CC amplifier (including R_c)

• Output resistance:

$$i_x = g_m v_x + \frac{(v_x - R_c i_x)}{r_o}$$

$$\frac{v_x}{i_x} = \frac{(R_c + r_o)}{(1 + g_m r_o)} \approx \frac{R_c + r_o}{g_m r_o} \approx \frac{r_c}{r_o} \approx \frac{1}{g_m}$$

$R_{out} = r_{\pi} \parallel \left\{ \frac{(R_c + r_o)}{(1 + g_m r_o)} \right\} \approx r_{\pi} \parallel \frac{R_c + r_o}{g_m r_o} \approx r_{\pi} \parallel \frac{1}{g_m}$

So, if you see the output resistance. So, we do have the same small signal model and to know the output resistance, we have to make the signal is equal to 0 namely base terminal we are making it AC ground. We are stimulating the circuit with V_x and we are observing the corresponding current flowing through the circuit. So, what we have here it is this i_x , it is summation of all these currents. So, the current flowing through the device here the whatever voltage dependent current source, it is g_m into V_{be} and incidentally V_{be} equals to minus v_x . So, we can say that this current is g_m into v_x and then, we also have the current flowing through this device and that device, it is we do have say V_x minus whatever the current flow we do have.

Now, for simplicity what we have done is that since we are stimulating this terminal by V_x directly, instead of considering this entire circuit and try to find what will be the impedance. We may split this part; one is this r_{π} part and then, rest of the things. Now, if I get the

impedance of this rest of these things, then the total output impedance it will be simply this output resistance coming from the insertive one in parallel with r_{pi} .

So, let us do this exercise considering only the insertive part. So, if we say that the current flowing through this V_x without considering this one it is i_x , then this current flow actually it is same through this R_c . So, the current flow through this R_c it is i_x . So, the voltage here V_c , it is R_c multiplied by i_x . So, the current flow through this r_{naught} it is V_x here minus V_c divided by r_{naught} and V_c it is R_c into i_x .

So, that it is a small you know small trick by which we can we can simplify the analysis. If you want you can consider the entire circuit and you can find the output resistance. But if you do this one, it will be the analysis it will be simpler. Now, what we obtain here, it is the relationship between i_x and V_x and from that we can get the ratio of V_x divided i_x that gives us V_c plus sorry R_c plus r_{naught} divided by 1 plus g_m into r_{naught} .

So, interestingly, the resistance coming from this circuit, it is again it will be quite small. This part it may be dominating. So, we may approximate this by R_c plus r_{naught} divided by 1 plus g_m into r_{naught} . In later of our discussion, in other circuits, we may come across the similar kind of circuit and try to remember this sorry this party you may ignore; try to remember this outcome of this analysis.

So, there based on the value of this R_c and then, r_{naught} it may be in the order of 1 by g_m that may be the conclusion. Now, since we do have the r_{pi} here, the total resistance, output resistance; it is the resistance coming from this encircle part and then, r_{pi} part. So, this is again you may consider this is r_{pi} in parallel with R_c plus r_{naught} divided by g_m into r_{naught} and if I say that this is this may be dominating over this R_c . Then, you may approximate this by r_{pi} in parallel with 1 by g_m which is again it can be said that 1 by g_m it is dominating over r_{pi} .

So, the conclusion is that this output resistance in the order of $1/g_m$ which is quite and not quite load ok. So, let us see the effect of this R_c on the input capacitance that is very tricky. So, in the next slide, we will be analysing the circuit to get the input capacitance.

(Refer Slide Time: 15:48)

Small signal analysis of more realistic CC amplifier (including R_c)

• Input Capacitance:

$$v_o = v_{in} \cdot \frac{(R_c + r_o + g_m \cdot r_{\pi} \cdot r_o)}{(r_{\pi} + R_c + r_o + g_m \cdot r_{\pi} \cdot r_o)}$$

$$v_c = R_c \cdot \frac{(v_{in} - v_o)}{r_{\pi}} = \frac{R_c v_{in}}{r_{\pi}} \left\{ 1 - \frac{v_o}{v_{in}} \right\}$$

$$= \frac{v_{in}}{r_{\pi} + (R_c + r_o + g_m \cdot r_{\pi} \cdot r_o)}$$

$$C_{in} = c_{\pi}(1 - A_{vE}) + c_{\mu}(1 - A_{vC})$$

$$= c_{\pi} \cdot \frac{r_{\pi}}{r_{\pi} + R_c + r_o + g_m \cdot r_{\pi} \cdot r_o} + c_{\mu} \cdot \frac{(R_c + r_o + g_m \cdot r_{\pi} \cdot r_o)}{r_{\pi} + R_c + r_o + g_m \cdot r_{\pi} \cdot r_o}$$

Yes. So, things are getting really getting complicated, let us see what are the complications are getting here. So, we do have the small signal equivalent circuit, we do have the C_{μ} and then, C_{π} and note that compared to our previous discussion, the voltage here it is not AC ground rather voltage it is V_c and V_c , it is of course, it is function of input voltage.

Now, we may recall that the voltage at the output node V_o , we already have this expression of this V_o in terms of V_{in} having this factor. So, numerator it is smaller than the denominator by

this r_{π} and then, the expression of the V_c , it is the R_c multiplied by whatever the current is flowing and that is $V_{in} - V_o$ divided r_{π} that we have discussed before.

This can be rearranged in this form. We can take v_{in} outside. We are getting a factor $1 - \frac{v_o}{v_{in}}$. So, if I use the expression of v_o in terms of v_{in} and if I plug in that expression here for v_o divided by v_{in} , what I will be getting here? It is this factor in the numerator, I will be having this part minus this part and will be having r_{π} in the numerator and in the denominator whatever the things we do have here, it will be coming there. And the numerator this r_{π} it is getting cancelled here. So, what we have it is in the numerator, we do have R_c and then V_{in} and then, rest of the things in the denominator.

So, the expression of the v_c , it is v_{in} in terms of v_{in} it is v_{in} multiplied by R_c divided by r_{π} $R_c + r_{naught} + g_m$ into r_{π} into r_{naught} . So, anyway this is as expected; this is smaller than the input voltage. Note that we are not connecting any resistance here, the moment we connect the resistance here, the voltage here at the collector terminal it is expected to be having much higher value. But right now, we are not considering. If I consider R_L is infinite here, earlier we consider R_L .

So, if I consider that this gives an indication that V_{vc} , it is a fraction of the v_{in} . In fact, intuitively we can see that if this side is open and if I start from the base terminal and then, if you go to the emitter terminal, then if you go to the collector and then if you go to the ground, what we can see here it is we are seeing one resistance r_{π} ; here, we do have one resistance R_c and also we do have this resistance r_{naught} multiplied by $1 + g_m$ into r_{π} , that can be expanded in this form.

So, from here to here that is the impedance and then the drop across this R_c is the V_c . So, you may say that v_{in} it is getting potentially divided by this factor. So, that makes sense this expression of this one makes sense. So, whatever it is over the both r_{π} sorry C_{π} and C_{μ} are getting miller affected. So, both C_{μ} other terminal of the C_{μ} , it is a non AC ground. And of course, C_{π} , it is also the other end, it is connected to V_o that is also non ac ground right they are having signals.

So, the input capacitance at the base terminal, what will be getting it is C_{π} multiplied by $1 - A_{vE}$ and let you call this is gain it is A_{vE} and this A_{vE} , it is V_o divided by V_{in} . On the other hand, contribution of the C_{μ} , it is C_{μ} multiplied by $1 - A_{vC}$ and what is A_{vC} ? We are defining this A_{vC} by A_C divided by V_{in} and we already have the expression of these two ratios; one it is from here; another it is from here. So, if you plug in those expressions and if we subtract from one, the first part if you see, what will be getting in here? It is the denominator and if I compare denominator and numerator, we do have only r_{π} is the difference. So, in the first part, what we will be getting is that C_{π} getting multiplied by r_{π} divided by the entire thing.

So, likewise when you consider the contribution of the C_{μ} and if I take this ratio getting subtracted from 1, what will be getting? Here, it is except R_c which is here rest of the things it will be there. So, that is why this C_{μ} , it is getting multiplied by this part which is having R_c , then $r_{\pi} \parallel g_m$ into r_{π} into $r_{\pi} \parallel g_m$ and then, denominator it is same as this one. So, you may say that this part this factor, this factor it is approximately 1; whereas, this part it is very small. So, approximately 0. So, we can say that the C_{in} , it is approximately equals to C_{μ} for all practical purposes.

So, again, we are converging to the similar kind of conclusion, namely the input capacitance is low, input resistance of this circuit it is high; output resistance, it is remaining low, voltage gain it is approximately remaining 1. So, we have done this analysis for a common collector circuit considering this collector resistance.

Similar kind of things we can do for the MOS counterpart and there instead of R_c , we may say that we may be connecting R_D at the drain terminal and this may be the corresponding MOS transistor and we can do the similar analysis and in the small signal equivalent circuit, it will be very similar except this corresponding r_{π} will not be there or whatever the analysis so far we have done, that can be utilized judiciously by considering this r_{π} approaching to very high value ok.

So, this is what you can share the analysis across different types of the circuit configuration rather different types of circuit namely analysis from common collector, you can propagate to common drain. Note that it is not possible the other way because in common drain circuit. We do not consider this gate to source resistance. So, as a result deriving the expression, it will be for the common collector from the common drain, it is not possible.

I should say rather analysis for common collector, it is relatively generic that can be extended for the common drain circuit. So, as I said that let us look into the common drain amplifier considering the resistance connected to the drain terminal.

(Refer Slide Time: 25:45)

Page 23/23

Small signal analysis of more realistic CD amplifier (including R_D)

- **Voltage gain:**

$$v_o = v_{in} \cdot \frac{(R_c + r_o + g_m \cdot r_\pi \cdot r_o)}{(r_\pi + R_c + r_o + g_m \cdot r_\pi \cdot r_o)}$$

CC

CD $v_o = v_{in} \cdot \frac{g_m \cdot r_o}{(1 + g_m \cdot r_o)} \rightarrow A_v = \frac{g_m \cdot r_o}{(1 + g_m \cdot r_o)} \approx 1$
- **Output resistance:**

$$R_{out} = r_\pi \parallel \left\{ \frac{(R_c + r_o)}{(1 + g_m \cdot r_o)} \right\}$$
- **Input Capacitance:**

$$c_{in} = c_\pi(1 - A_{vE}) + c_\mu(1 - A_{vC})$$

$$= c_\pi \cdot \frac{r_\pi}{(r_\pi + R_c + r_o + g_m \cdot r_\pi \cdot r_o)} + c_\mu \cdot \frac{(R_c + r_o + g_m \cdot r_\pi \cdot r_o)}{(r_\pi + R_c + r_o + g_m \cdot r_\pi \cdot r_o)}$$

So, we do have this resistance connected to the drain terminal and then, we do have the RD coming in the small signal equivalent circuit and then, we do not have any r pi here. So, rather if I say that whatever the previous analysis we have done, if I consider the r pi in those

equation approximately going to be very high. Then, we can get the equation for the common drain. So, let us start with the voltage gain and let us follow that approach for common. So, this is the equation, we obtain for common collector circuit and let us try to do for the common drain circuit, namely this circuit.

So, if I say that expression of the V_o can be obtained by considering this equation which you already have derived and that can be done by making this is going to be very high and this is going to be very high and this is also going to be very high. Namely, in the numerator if I keep this term and in the denominator if I keep this term and this term and then, if I pull out this r_{pi} and if I cancel it what will be getting here it is V in multiplied g_m into r_{naught} in the numerator and in the denominator, we will be having 1 plus g_m into r_{naught} .

So, from here directly we can say that the voltage gain, it is g_m into r_{naught} divided by 1 plus g_m into r_{naught} ok; yeah. So, here this R_D , I should say on the in the voltage gain, it is hardly having any effect and again finally, we approximate this by 1 . Now, similar kind of things we can do for the output resistance. This part let me clear; yeah.

(Refer Slide Time: 27:05)

Page 28/29

Small signal analysis of more realistic CD amplifier (including R_D)

• Voltage gain:

$$v_o = v_{in} \cdot \frac{(R_c + r_o + g_m \cdot r_{\pi} \cdot r_o)}{(r_{\pi} + R_c + r_o + g_m \cdot r_{\pi} \cdot r_o)}$$

• Output resistance:

$$R_{out} = r_{\pi} \parallel \left(\frac{(R_D + r_o)}{(1 + g_m \cdot r_o)} \right)$$

$R_{out} = \frac{R_D + r_o}{1 + g_m \cdot r_o} \approx \frac{1}{g_m}$

• Input Capacitance:

$$c_{in} = c_{\pi}(1 - A_{vE}) + c_{\mu}(1 - A_{vC})$$

$$= c_{\pi} \cdot \frac{(R_c + r_o + g_m \cdot r_{\pi} \cdot r_o)}{(r_{\pi} + R_c + r_o + g_m \cdot r_{\pi} \cdot r_o)} + c_{\mu} \cdot \frac{(R_c + r_o + g_m \cdot r_{\pi} \cdot r_o)}{(r_{\pi} + R_c + r_o + g_m \cdot r_{\pi} \cdot r_o)}$$

So, if I consider say this part, earlier we have seen that if I want to know the resistance or impedance at the output terminal and if I stimulate this circuit by a voltage source called V_x and then, if I consider the corresponding i_x and then, if I take the ratio of this two that is supposed to be given as the resistance coming from this circuit. And the previous analysis shows that the corresponding resistance is R_c plus r_o divided by $1 + g_m r_o$.

Now, this R_c instead of R_c , we can write R_D for this case and rest of the things it is same. Now, if I consider this part also while you are considering this output resistance of course, you have to make it ground and since, this r_{π} it is very high. So, this is very high. So, we can say that you can neglect this part and you can consider only this one.

So, what we have is r_{out} or r_{out} , it is R_D . This is D ; R_D plus r_{out} divided by $1 + g_m r_{out}$. So, that is how we do get the expression of output resistance. Again, this is in the order of $1/g_m$. Now similarly, we can find what will be the input capacitance. Again, for this circuit let me clear it up; yeah.

(Refer Slide Time: 28:50)

Small signal analysis of more realistic CD amplifier (including R_D)

• Voltage gain:

$$v_o = v_{in} \frac{(R_c + r_o + g_m \cdot r_{\pi} \cdot r_o)}{(r_{\pi} + R_c + r_o + g_m \cdot r_{\pi} \cdot r_o)}$$

$\frac{v_o}{v_{in}} = \frac{g_m r_o}{1 + g_m r_o} \approx 1$ $R_{in} \rightarrow \infty$

• Output resistance:

$$R_{out} = r_{\pi} \parallel \left\{ \frac{(R_c + r_o)}{(1 + g_m \cdot r_o)} \right\} \approx \frac{1}{g_m} \text{ "low"}$$

• Input Capacitance:

$$C_{in} = c_{\pi} (1 - A_{v1}) + c_{gs} (1 - A_{v2})$$

$$= c_{\pi} \frac{(R_c + r_o + g_m (r_{\pi} r_o))}{(r_{\pi} + R_c + r_o + g_m (r_{\pi} r_o))} + c_{gs} \frac{(R_c + r_o + g_m (r_{\pi} r_o))}{(r_{\pi} + R_c + r_o + g_m (r_{\pi} r_o))}$$

$C_{in} \approx C_{gd}$ (low)

So, to get the input capacitance C_{in} , we do have the contribution coming from C_{gs} earlier it was for common collector it was C_{π} . So, we need to replace this by C_{gs} . So, likewise here also we are having now it is C_{gd} ; earlier in common collector circuit, it was C_{μ} . So, that need to be replaced by C_{gd} and then, we do have the voltage gain from here to here. So, it is not emitter this rather from gate to source and gate to drain and this, this gate to source gain we already obtained here which is given here.

And we say that the corresponding voltage gain v_o divided by v_i in it was $g_m r_{naught}$ divided by $1 + g_m r_{naught}$ and so, that that gives us the corresponding ratio likewise you can find what will be the this the A_{vD} . So, that is v_o/v_i ; v_o is the voltage here divided by v_i . And if you plug in that expression, what you will be getting is that C_{in} , it is C_{gs} multiplied by a factor like this and then, C_{gd} multiplied by this factor. Now, whatever the factor we are writing here, here and here of course, therefore, common collector. To convert the equivalent factor for common drain, what you have to do you have to you know make this term, this term and this term dominating.

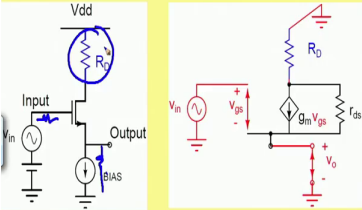
So, likewise here also we can see that these terms are dominating. So, in this case what we will be getting is that C_{in} it will be C_{gs} multiplied by 1 divided by $1 + g_m r_{naught}$. So, let me write here C_{in} equals to C_{gs} multiplied by 1 divided by $1 + g_m r_{naught}$ and then, C_{gd} . Here C_{gd} multiplied by we do have here $g_m r_{naught}$ divided by $1 + g_m r_{naught}$. So, this part it is approaching to 0 ; this part it is approaching to 1 ok.

So, then we can say that C_{in} ; C_{in} it is approximately equal to C_{gd} right again this is quote and unquote low. This is approximately 1 ; output resistance is in the order of $1/g_m$ which is quote and unquote low. Input resistance of course, remains high. So, then you can say that for common drain even if I consider R_D , then the basic required property of the buffer, it remains the same.

(Refer Slide Time: 32:40)

Page 25/28

Small signal analysis of more realistic CD amplifier (including R_D)



• **Voltage gain:**

$$v_o = v_{in} \cdot \frac{(R_C + r_o + g_m \cdot r_\pi \cdot r_o)}{(r_\pi + R_C + r_o + g_m \cdot r_\pi \cdot r_o)}$$


• **Input Capacitance:**

$$c_{in} = c_\pi(1 - A_{vE}) + c_\mu(1 - A_{vC})$$

$$= c_\pi \cdot \frac{(R_C + r_o + g_m \cdot r_\pi \cdot r_o)}{(r_\pi + R_C + r_o + g_m \cdot r_\pi \cdot r_o)} + c_\mu \cdot \frac{(R_C + r_o + g_m \cdot r_\pi \cdot r_o)}{(r_\pi + R_C + r_o + g_m \cdot r_\pi \cdot r_o)}$$

• **Output resistance:**

$$R_{out} = r_\pi \parallel \left\{ \frac{(R_C + r_o)}{(1 + g_m \cdot r_o)} \right\}$$



Now, so far, we have considered either R D or R L here or maybe R s here. Now, if I consider maybe multiple elements together, things of course, it will be getting more and more complicated.

(Refer Slide Time: 33:00)

Exercise: Small signal analysis of more realistic CC / CD amplifier (including R_L and R_c/R_D)

- Output resistance: ?
- Input Capacitance: ?
- Voltage gain: ?

The slide also features a video feed of a man in a light blue shirt in the bottom right corner and a Windows taskbar at the bottom.

You may try out; you may try out to solve say this kind of circuits, where we do consider we do consider say the R_c part and then, R_L part. So, likewise we can consider R_D and R_L together and then, you can try to see what is the corresponding input resistance voltage gain and then, input capacitance ok. So, I rather suggest you, you can work it out in our numerical examples. So, we will be covering. So, we may not be repeating the analysis again here ah. But in our numerical examples, we will be discussing that. I think what we have, it is I think today let us let us conclude whatever the things we have covered.

(Refer Slide Time: 33:58)

Conclusion:

- ☐ CC and CD amplifiers works as voltage mode buffer
- ☐ Basic operation and biasing has been discussed
- ☐ Analysis for gain, impedance and input capacitance has been discussed
- ☑ Analysis for voltage gain, impedance and input capacitance considering realistic components
 - ☐ RL
 - ☐ RS
 - ☐ RC / RD
- ☑ Numerical examples and Design of CC and CD to be covered

So, far as I say that we are talking about common collector and common drain amplifier. Previous lecture, we have discussed about this part namely the c in common collector and common drain as voltage mode buffer. Basic operation and biasing, we just touched upon and then, we did the analysis for voltage gain and then, impedance and then, input capacitance for idealistic bias. And today, what you have done it is, we have done the analysis considering realistic components to get the voltage gain, impedance and input capacitances.

Realistic components are the load resistance may be connected at the emitter or common or the source of the transistor respectively for BJT and MOS transistor and then, source resistance, then also we have discussed about the effect of resistance at the collector terminal or drain terminal. What we are planning to go for the next class, it is numerical examples and some design guidelines. I think that is all I do have to cover.

Thank you for listening.