

Analog Electronic Circuits
Prof. Pradip Mandal
Department of Electronics and Electrical Communication Engineering
Indian Institute of Technology, Kharagpur

Lecture – 45
Common Collector and Common Drain Amplifiers (Contd): Analysis (Part A)

Dear students, welcome back to our online NPTEL certification course. The topic of this course is Analog Electronic Circuit. Myself Pradeep Mandal from E and ECE department of IIT, Kharagpur. Today's topic of discussion it is Common Collector and Common Drain Amplifiers, rather I should say it is continuation of this topic. Previous day we have discussed about relatively idealistic bias situation and today we are going to a little detail considering some practical circuit components also.

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CONCEPTS COVERED

Concepts Covered:

- Motivation of using CC and CD amplifiers
- Basic operation and Biasing of CC and CD amplifiers
- Analysis for voltage gain, impedance and input capacitance considering realistic components
 - R_L
 - R_s
 - R_c / R_D
- Design guidelines
- Numerical examples

The slide features a dark blue background on the left with the text 'CONCEPTS COVERED' in yellow. The right side has a yellow background with a list of topics. A small video inset in the bottom right shows a man in a white shirt. At the bottom, there are logos for IIT Kharagpur and a navigation bar.

So, what we have to cover today, as I said we have discussed the motivation part of the common source and sorry common collector and common drain amplifier, basic operation and biasing also it is done. And we are going to go a little detail of analysis of voltage gain and impedance, input capacitance, considering realistic biasing and their associated components.

In fact, in the previous class we have discussed about the analysis of the circuit for voltage gain, impedance and input capacitance ignoring these components, and today we are going to see that what will be there you know consequences if we consider on a practical components. So, let us start with the common collector amplifier, considering the R_L , and then we will be moving to the next one is considering the source resistance R_S , and then we will go for the collector terminal resistance.

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Small signal analysis of more realistic CC amplifier (including R_L)

• **Input capacitance:**

$$C_{in} = C_{\pi} (1 - A_v) + C_{\mu}$$

$$C_{in} = \frac{C_{\pi} r_{\pi}}{(g_m r_{\pi} + 1) \cdot (r_o \parallel R_L) + r_{\pi}} + C_{\mu}$$

• **Input resistance:**

$$R_{in} = (g_m r_{\pi} + 1) \cdot (r_o \parallel R_L) + r_{\pi}$$

• **Voltage gain:**

$$A_v = \frac{v_o}{v_{in}} = \frac{(g_m r_{\pi} + 1) \cdot (r_o \parallel R_L)}{(g_m r_{\pi} + 1) \cdot (r_o \parallel R_L) + r_{\pi}}$$

So, to start with let us let you consider the common collector amplifier and also in the common collector amplifier we are including this R_L . So, this R_L may be coming from the bias circuit, representing the finite conductance of the bias circuit or maybe and or maybe additional load resistance we are connecting at the output node with respect to ground.

So, whatever it is let us consider this R_L in our analysis and here we do have small signal equivalent circuit of the common collector amplifier having this R_L included. So, we do have r_{π} here and then g_m into V_{be} coming from the BJT transistor, and then we do have this R_L bias circuit it is then the DC part it is 0 in this small signal equivalent circuit. So, likewise the DC part also it is 0, we do have only the signal coming to the base terminal. So, we do have base terminal here, and then we do have the emitter terminal here and then we do have the collector terminal. And this is the output port, so the output voltage it is V_o and the emitter with respect to ground.

Now, in our previous analysis where we have excluded this R_L there we have seen the expression of the input capacitance. Basically, the input capacitance at the base with respect to the AC ground. And there what we have seen if you consider the two parasitic components capacitive components one is c_{μ} and then we do have the c_{π} here.

So, what are the contributions c_{μ} the right side of c_{μ} it is connected to ground. So, it is directly contributing to the input capacitance as is, as you have discussed before. On the other hand, this c_{π} it is making a connection from input to output of this amplifier. And the gain of this circuit we have discussed if you see here in fact, if we recall c_{π} it is having two parts, one is c_{π} into $1 - \text{voltage gain of this amplifier}$ plus c_{μ} , and this voltage gain to get the expression of this one we need to get the voltage gain.

So, in our previous analysis we obtain the voltage gain v_o divided by v_{in} where R_L it was very high. Now, if we consider this R_L what you are getting this R_L coming in parallel with r_{naught} . So, whatever the previous expression we are having namely $g_m r_{\pi} + 1$ into r_{naught} , and in the denominator it was g_m into $r_{\pi} + 1$ into $r_{\text{naught}} + r_{\pi}$.

Now, if we have this R_L which is essentially coming in parallel with r_{π} then whatever the derivation we have done before that can be as well utilized just by replacing this r_{π} by R_L in parallel. So, wherever we do have r_{π} in our previous derivation if you replace that by r_{π} in parallel with R_L , then we do get the corresponding voltage gain.

So, if you put this expression of this voltage gain here, so this is A_v . So, if you put that voltage gain expression here, what we will be getting here it is c_{μ} in it is having c_{μ} and then c_{π} part it is having a factor which is a having in the numerator we do have r_{π} and in the denominator we do have g_m into $r_{\pi} + 1$, then r_{π} and R_L in parallel plus this r_{π} . So, that is the expression of c_{in} .

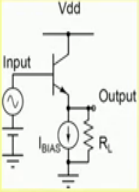
Likewise, when you consider the input capacitance if you see the expression or if you recall the previous expression of the input resistance without considering this R_L , what we had it is R_{in} it was g_m into $r_{\pi} + 1$ into r_{π} plus r_{π} .

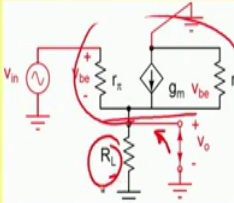
Now, in presence of this R_L what you have to do? Again, instead of r_{π} here we need to replace this by r_{π} in parallel with R_L . So, the expression of the input resistance we do have it is g_m into $r_{\pi} + 1$ into r_{π} in parallel with R_L plus r_{π} . So, that is how when you consider this R_L in our circuit. What we are getting it is this additional you know modification and that is that is good enough to get its effect.

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Small signal analysis of more realistic CC amplifier (including R_L)





• **Input capacitance:**

$$c_{in} = \frac{c_{\pi} \cdot r_{\pi}}{(g_m \cdot r_{\pi} + 1) \cdot (R_L \parallel r_o) + r_{\pi}} + c_{\mu}$$


$\approx c_{\mu}$ (since $(g_m \cdot r_{\pi} + 1) \cdot (R_L \parallel r_o) + r_{\pi} \rightarrow \infty$)

• **Input resistance:**

$$R_{in} = (g_m \cdot r_{\pi} + 1) \cdot (r_o \parallel R_L) + r_{\pi} = (\beta + 1) \cdot r_o \parallel R_L + r_{\pi}$$

• **Voltage gain:**

$$\frac{v_o}{v_{in}} = \frac{(g_m \cdot r_{\pi} + 1) \cdot (r_o \parallel R_L)}{(g_m \cdot r_{\pi} + 1) \cdot (r_o \parallel R_L) + r_{\pi}} = \frac{(\beta + 1) \cdot r_o \parallel R_L}{(\beta + 1) \cdot r_o \parallel R_L + r_{\pi}}$$



In fact, you can further simplify this one wherever we do have the g_m into r_{π} you may replace this by β of the transistor. So, here also we can write this is $\beta + 1$ into r_o naught in parallel with R_L plus r_{π} . So, likewise here also you can write this part you can write $\beta + 1$ into this one. So, similarly for the voltage gain here you can write in terms of β instead of g_m and r_{π} , so we can say that this is $\beta + 1$ into r_o in parallel with R_L divided by $\beta + 1$ into r_o in parallel with R_L plus r_{π} .

So, whatever it is all practical purposes, for all practical purposes even if you consider the this R_L you can well approximate that this part after multiplying with $\beta + 1$, it will be very high and you may say that this part it is almost 0 compared to c_{μ} . So, as a result you may say that c_{in} it is practical it is c_{μ} .

Likewise, when you consider the input resistance it is we do have r_{π} coming in series with β into $r_{\text{naught}} + R_L$. So, though r_o or r_{naught} coming in parallel with R_L , but even though if I consider the effect of R_L since we do have $\beta + 1$ it is coming in the multiplication again this may be you may consider this is very high. So, we may say that this is going to be a very high compared to other resistances in the circuit. So, likewise when you consider the voltage gain even if I consider the effect of R_L this is this can be well approximated by close to 1.

So, the bottom line is at even if I consider R_L the main property of the input resistance to be high, voltage gain it is close to 1, input capacitance is very small defined by c_{μ} , those things are getting retained. Also the other parameter we have to think of is that the output resistance. So, if you see this circuit and at this point if you see what is the output resistance, it is basically the output resistance coming from the rest of the circuit coming in parallel with R_L .

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Small signal analysis of more realistic CC amplifier (including R_L)

$$c_{in} = \frac{c_{\pi} \cdot r_{\pi}}{(g_m \cdot r_{\pi} + 1) \cdot (R_L || r_o) + r_{\pi}} + c_{\mu}$$

$$R_{in} = (g_m \cdot r_{\pi} + 1) \cdot (r_o || R_L) + r_{\pi}$$

$$\frac{v_o}{v_{in}} = \frac{(g_m \cdot r_{\pi} + 1) \cdot (r_o || R_L)}{(g_m \cdot r_{\pi} + 1) \cdot (r_o || R_L) + r_{\pi}}$$

• Output resistance:

$$R_o = \frac{v_x}{i_x} = \frac{1}{\left(g_m + \frac{1}{r_o} + \frac{1}{r_{\pi}} + \frac{1}{R_L} \right)} \approx \frac{1}{g_m} \rightarrow \text{"low"}$$

So, that is I guess it is very straight forward to converge. So, I do have, yes. We do have this is the expression of r_{naught} . The derivation of the c_{in} and voltage gain we already have done, so these things are already done. Now, in addition to that if you want to know what will be the output resistance is basically this resistance coming in parallel with whatever the resistance we do have in this circuit.

And again, if we refer back to our previous analysis without considering this R_L , where to get the output resistance what you have done is that we made this is AC ground and then we stimulate this circuit by say v_x , and then we observed the corresponding current say i_x , and then if you say what is the; so, this is i_x then if I take the ratio of v_x by i_x that is supposed to be giving us the output resistance.

And earlier whatever the discussion we had you may recall, that if I apply v_x here the current flow here it is v_x divided by r_{naught} , current flow here it is g_m into v_x because the V be equals to minus v_x , and then current flow here it is since this is a c ground, so this is 0, v in is 0. So, this current it is again v_x divided by r_{pi} . So, likewise the current flowing through this circuit it is v_x divided by R_L .

So, the total current i_x it is actually summation of all these four currents. And so if you if you simplify that what will be getting is that output resistance is reciprocal of the total conductance; total conductance coming from all these four elements, one is g_m part, another one is 1 by r_{naught} , then another one is 1 by r_{pi} here and then 1 by R_L . So, the total resistance again it is its expression is given here, but all practical purposes, even if I consider practical value of R_L this g_m it dominates over rest of the things, so you may approximate this is equals to 1 by g_m . So, this is as I said that basic property we are looking out of this common collector which is buffer voltage buffer it should be low, so that is what here also we are obtaining the same thing.

Now, that is about the common collector amplifier if you consider its counterpart mass counterpart namely the common drain stage and then if you consider this R_L for that, what you will be getting it is similar kind of things we can get only difference is that this r_{pi} , it will be it will not be there. So, you may say that because this r_{pi} is theoretical, it is in finite for common drain amplifier. So, this input resistance for common drain stage any way it is high. So, we may ignore this part. Rest of the parameters namely c in voltage gain and then output resistance we can think of. And everywhere again we can converge to the same conclusion.

So, let us see what I do have for you, yeah.

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Small signal analysis of more realistic CD amplifier (including R_L)

• **Voltage gain:**
$$A_v = \frac{v_o}{v_{in}} = \frac{g_m \cdot (r_{ds} || R_L)}{\{g_m \cdot (r_{ds} || R_L) + 1\}} \approx 1$$

• **Output resistance:**
$$R_o = \frac{1}{\left(g_m + \frac{1}{r_{ds}} + \frac{1}{R_L}\right)}$$

• **Input capacitance:**
$$C_{in} = c_{gs}(1 - A_v) + c_{gd} = \frac{c_{gs}}{\{g_m \cdot (r_{ds} || R_L) + 1\}} + c_{gd} \approx c_{gd}$$

So, the common drain stage considering R_L and its voltage gain, voltage gain it is given here. So, again you may recall the previous analysis where we have not considered R_L , for that the voltage gain it was g_m into r_{ds} divided by g_m into r_{ds} plus 1.

Now, in presence of this R_L this R_L it is coming in parallel with r_{ds} . So, you may say that wherever we do have the r_{ds} , we can we can replace this r_{ds} , this r_{ds} , and this r_{ds} by r_{ds} in parallel with R_L and that gives us the corresponding voltage gain. So, likewise if I use this expression to find the input capacitance and its expression it is c_{gs} multiplied by 1 minus voltage gain plus c_{gd} , where c_{gs} it is the capacitance from gate to source.

So, we do have c_{gs} here, and then likewise we do have the c_{gd} gate to drain and so this is c_{gd} . So, the expression of the input capacitance is c_{gs} multiplied by 1 minus 1 minus A_v and then plus c_{gd} , so that gives us if I use the expression of the A_v here, what we are getting is

that c_{gs} it is getting divided by a big factor here and as a result you may approximate again this by c_{gd} .

So, note that even if I consider the effect of R_L since this part it is much higher than 1, its effect it is very negligible and as a result the effect of c_{gs} it is very small. So, in fact, it is also very clear that if the voltage gain it is approximately one then this part it becomes approximately 0 or very small. So, we do have only c_{gd} left behind for the c_{in} . And again for the output resistance if I consider this R_L in addition to the other components what we have it is; so, let me use different color, yeah.

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Small signal analysis of more realistic CD amplifier (including R_L)

• Voltage gain: $R_{in} \rightarrow \infty$

$$\frac{v_o}{v_{in}} = \frac{g_m \cdot (r_{ds} || R_L)}{\{g_m \cdot (r_{ds} || R_L) + 1\}} \approx 1$$

• Output resistance: $R_o = \frac{1}{g_m + \frac{1}{r_{ds}} + \frac{1}{R_L}}$

• Input capacitance: $c_{in} = c_{gs}(1 - A_v) + c_{gd} = \frac{c_{gs}}{\{g_m \cdot (r_{ds} || R_L) + 1\}} + c_{gd}$

Handwritten notes in red:

- $v_{gs} = -v_x$
- $i_x = \frac{v_x}{R_L} + \frac{v_x}{r_{ds}} + g_m v_x$
- $\approx \frac{1}{g_m}$ "low"
- "low"

So, if I want to know what will be the output resistance, let me use red color, yeah. So, we have to make it this is ground and then we need to stimulate this circuit by say v_x and then current flowing through this is i_x and the moment we are applying say v_x here the current

flowing through this r_{ds} it is v_x divided by r_{ds} and then v_x incidentally it is same as minus v_{gs} .

So, as a result the current flow through this active device it is g_m into v_x because v_{gs} is equal to minus v_x . So, this minus sign it is suggesting that the direction of the current it will be from source to drain and g_m multiplied by that v_x . So, here of course, we do not have any conducting path and then the current flowing through this R_L it is v_x divided by R_L .

So, we do have these 3 components. And what gives us this i_x is equal to v_x divided by R_L plus v_x divided by r_{ds} plus g_m into v_x . Now, from this one we can directly say that the v_x divided by i_x equals to reciprocal of these total conductance. Again, even if I consider the effect of R_L for all practical purposes we can say that g_m it dominates, as a result here also we can say that the output resistance is approximately $1/g_m$, which is you may say that low cut and uncut low, input capacitance is low and the voltage gain it is approximately 1, right. And of course, the R in it remains infinite. So, this summary is that even if I consider R_L the basic property of the buffer it is getting maintained.

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Small signal analysis of more realistic CD amplifier (including R_L)

- **Voltage gain:**

$$\frac{v_o}{v_{in}} = \frac{g_m \cdot (r_{ds} || R_L)}{\{g_m \cdot (r_{ds} || R_L) + 1\}}$$
- **Output resistance:**

$$R_o = \frac{1}{\left(g_m + \frac{1}{r_{ds}} + \frac{1}{R_L}\right)}$$
- **Input capacitance :**

$$c_{in} = c_{gs}(1 - A_v) + c_{gd} = \frac{c_{gs}}{\{g_m \cdot (r_{ds} || R_L) + 1\}} + c_{gd}$$

Now, so far we have discussed about the effect of R_L and so likewise we may think of the resistance at this terminal resistance at this terminal considering some practical purposes.

Now, if I consider say source resistance and still if I consider this is 0 then even if I connect the source resistance R_s , since from this terminal to this terminal, there is no DC path. So, you may say that even if I consider R_s the behavior of the circuit remains unchanged. So, for common drain stage we need not to consider this case because we know that whether we consider R_s or not or it is impact it is not there.

However, if I consider say common collector stage; obviously, there will be the corresponding element here called r_{π} . So, it is better to in case if we consider R_s , then we need to; need to analyze its effect particularly for common collector stage.

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Small signal analysis of more realistic CC amplifier (with R_s)

• Output resistance:

$$v_{be} = -v_x \cdot \frac{r_\pi}{(R_s + r_\pi)}$$

$$A_v = \frac{R_{in}}{R_s + R_{in}} \cdot \frac{g_m r_\pi}{g_m r_\pi + 1} \approx 1 \approx 1$$

$$i_x = g_m \cdot v_x \cdot \frac{r_\pi}{(R_s + r_\pi)} + \frac{v_x}{r_o} + \frac{v_x}{(R_s + r_\pi)}$$

$$R_o = \frac{v_x}{i_x} = \frac{1}{\left\{ \frac{g_m \cdot r_\pi}{(R_s + r_\pi)} + \frac{1}{r_o} + \frac{1}{(R_s + r_\pi)} \right\}}$$

So, let us see the common collector amplifier considering this source resistance, signal source resistance. So, we do have the circuit given here. So, we are considering this source resistance R_s . Again, R_L we are ignoring, so we are considering one effect at a time. And then the small signal model of this circuit is given here and of course, we should have considered the signal here. Now, this supposed to be the output v_o . Now, what are the parameter we need to consider, the voltage gain? Input resistance, then output resistance, and then C_{in} .

Now, if you see that if I consider this R_s the behavior of the circuit from here to the output from the base terminal to the emitter terminal, the behavior remains the same. In fact, if you see the input resistance of this part and if I call this is the input resistance that remains the same as what we have discussed before. So, we are not going to consider the analysis for again repeating this R_{in} . Same thing for C_{in} also.

So, whatever the r_{in} we do have, so r_{in} also remains unchanged from base terminal to the emitter terminal. Even if I consider R_s and the voltage gain if you see of course, if I consider this is the primary input and this is a corresponding primary output, and if I say the voltage gain from again from base to collector, the emitter terminal gain remains the same namely close to 1, only difference is that if I consider R_s I do have v_{in} in here and then we are applying the signal through this source resistance R_s to the effective input resistance, whatever the effective input resistance we do have.

Now, to consider the effect of this R_s in the in the voltage gain we need to consider what is the potential division it is happening before it before the signal arrives to it is a base terminal. So, what is the additional factor we will be getting in the voltage gain? It is R_{in} divided by R_s plus R_{in} . So, this is the additional factor. So, if I multiply this with v_{in} that gives us the base voltage, and then from the base point to the corresponding collector point the gain it is approximately 1.

Now, if you see here most of the time this R_{in} as I said R_{in} is very high. So, this factor whole factor it is approximately 1. So, I should say that the voltage gain A_v which is in this case R_{in} divided by R_s plus R_{in} multiplied by whatever the previous voltage gain we are having. Namely, if you recall it was $g_m r_{\pi}$ divided by $g_m r_{\pi} + 1$. So, it was it was I should say rather this portion it was approximately 1, and this is also getting approximately 1, as a result overall it is also becoming approximately 1.

So, again we are not going to discuss about the output similar to output input resistance, we are not going to discuss about the voltage gain. Only thing we are going to discuss here it is the output resistance.

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Small signal analysis of more realistic CC amplifier (with R_s)

• Output resistance:

$$v_{be} = -v_x \cdot \frac{r_{\pi}}{(R_s + r_{\pi})}$$

$$i_x = g_m \cdot v_x \cdot \frac{r_{\pi}}{(R_s + r_{\pi})} + \frac{v_x}{r_o} + \frac{v_x}{(R_s + r_{\pi})}$$

$$R_o = \frac{v_x}{i_x} = \frac{1}{\left(\frac{g_m \cdot r_{\pi}}{(R_s + r_{\pi})} + \frac{1}{r_o} + \frac{1}{(R_s + r_{\pi})} \right)} \approx \frac{(R_s + r_{\pi})}{g_m} \approx \frac{R_s}{g_m}$$

So, let us see what is the output resistance of the common collector amplifier will be getting in presence of this source resistance R_s . Now, to get the source resistance sorry the output resistance of this circuit, what you have to do? We need to stimulate the circuit by say v_x and we need to observe the corresponding current here i_x . So, before we go into the i_x expression, we need to make the voltage, the signal here it is 0, so that is why you are making this a c ground and if we apply v_x here, the voltage coming at the base to emitter terminal, so between base to emitter terminal what we have it is V_{be} that is it is essentially the potential division of this v_x appearing across this r_{π} .

So, we can say that the voltage appearing across r_{π} it is v_x multiplied by r_{π} divided by R_s plus r_{π} , but then this side we for V_{be} we call this is positive and this is negative and v_x we are applying here it is positive and this is negative So, as a result the expression of the V_{be} in this case it is minus v_x multiplied by r_{π} divided by R_s plus r_{π} . So, this is one relationship

we obtained between V_{be} and v_x . Now, the current flow through this active device is g_m into V_{be} , so that becomes g_m into V_{be} instead of V_{be} we can write minus v_x into so and so on.

So, we may say that the current actually it is; so, this starts flowing in this direction and its expression is this one. So, this current expression is given here. On the other hand, the current flowing through this i_{nought} r_{nought} , it is v_x whatever the voltage you are applying here divided by r_{nought} and the current flowing through the base terminal towards the base, namely this current it is v_x divided by R_s plus r_{pi} . So, this is what we are getting. So, the total i_x current we can write in terms of v_x . So, from this one we can get the ratio of v_x divided by i_x and that gives us the output resistance. So, which is reciprocal of the conductance of the 3 parts and the first part conductance it is g_m into r_{pi} divided by R_s plus r_{pi} , second one is 1 by r_{nought} , third one it is 1 by R_s plus r_{pi} .

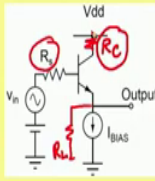
So, in this case again depending on the value of this R_s we may consider say this R_s with respect to r_{pi} , but most important thing is that we do have g_m into r_{pi} getting multiplied. So, now, we can directly see that compared to this term this term it is dominating, and same the similar kind of conclusion we can make because this r_{nought} it is reasonably high, as a result you can say that this term it is dominating. So, the resistance you may say that R_s plus r_{pi} divided by g_m into r_{pi} .

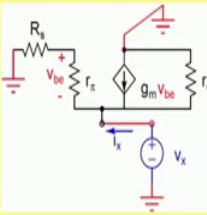
Now, if I say that R_s is varying, R_s is maybe comparable with r_{pi} then it remains low, but if say R_s is very high if it is very high then of course, we may not be able to approximate this by 1 by g_m . So, depending on the relative value of this R_s and r_{pi} , we can get the corresponding output resistance. But again even if the R_s it is in the order of same as r_{pi} then also you can say that this is in the order of 1 by g_m . So, in case say R_s equals to r_{pi} , then this may be 2 divided by g_m that that gives you some idea that what may be the effect of this R_s on the output resistance.

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Small signal analysis of more realistic CC amplifier (with R_s)






• Output resistance:

$$v_{be} = -v_x \cdot \frac{r_\pi}{(R_s + r_\pi)}$$

$$i_x = g_m \cdot v_x \cdot \frac{r_\pi}{(R_s + r_\pi)} + \frac{v_x}{r_o} + \frac{v_x}{(R_s + r_\pi)}$$

$$R_o = \frac{v_x}{i_x} = \frac{1}{\left\{ \frac{g_m \cdot r_\pi}{(R_s + r_\pi)} + \frac{1}{r_o} + \frac{1}{(R_s + r_\pi)} \right\}}$$



So, this is the effect coming from the R_s . So, likewise if I consider, yeah. So, so far we have considered the R_s , earlier we have considered the effect of R_L and then other part is also possible depending on the circuit connection. It may not be intentional, but there may be some practical situation due to which there may be resistance coming in the collector terminal, in between the collector terminal and the V_{dd} .

So, let us consider the effect of this resistance call R_c on different parameters of the common collector amplifier.

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Small signal analysis of more realistic CC amplifier (including R_c)

CC

• Voltage gain:

$$v_{be} = (v_{in} - v_o) \quad v_c = R_c \frac{(v_{in} - v_o)}{r_\pi}$$

$$\frac{(v_o - v_c)}{r_o} = \frac{(v_{in} - v_o)}{r_\pi} + g_m(v_{in} - v_o)$$

$$v_o \left\{ 1 + g_m r_o + \frac{R_c}{r_\pi} + \frac{r_o}{r_\pi} \right\} = v_{in} \left\{ g_m r_o + \frac{R_c}{r_\pi} + \frac{r_o}{r_\pi} \right\}$$

$$v_o = v_{in} \frac{(R_c + r_o + g_m r_\pi r_o)}{(r_\pi + R_c + r_o + g_m r_\pi r_o)}$$

So, what we have it is, yeah. So, here we do have the circuit the corresponding circuit, we do have the common collector circuit having this R_c connected to V_{dd} . Note that, please do not get confused that output still remains, the output terminal is basically the emitter of the transistor even though we do have a meaning maybe some signal coming to the collector terminal, but we may not be really concerned about this signal unless otherwise it is stated.

So, our input and output terminal remains base and emitter respectively. And even though we do have R_c connected here still we call this is common collector amplifier. We may say that collector is having some resistance, but still we call it is common collector without any you know confusion.

And here we do have the small signal equivalent circuit and if you see the small signal equivalent circuit we are adding this R_c s. And then, if you see this circuit we can go step by

step and we can you can find what will be the impact of this R_c on different parameters namely voltage gain, then output resistance, and then input capacitance and so and so.

So, let me take a small break and then we will come back to discuss all those things, namely we will start with voltage gain.

Thank you.