

Analog Electronic Circuits
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Lecture - 26
Common Emitter Amplifier (contd.)
(Part A)

Dear students, welcome back to our NPTEL course on Analog Electronic Circuits; myself Pradip Mandal from E and EC department of IIT Kharagpur. So, we are going to continue our previous topic namely the Common Emitter Amplifier, we have started this topic in the previous class and we are going to continue on the same thing.

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So, what are the plan we do have it is in the previous class we have discussed about the CE amplifier with fixed bias. And, today we will be going little detail of another kind of bias

called self bias and in the previous discussion we already have cleared that fixed bias it is having some stability issue, particularly the operating point stability issue which is resolved by this self biasing that is what we will be discussing in detail.

Then subsequently we will be discussing about the self biased CE amplifier and then its corresponding analysis having two parts. One is the DC operating point analysis and then small signal analysis which is eventually giving us small signal equivalent circuit of CE amplifier having self bias. And, then subsequently we will be talking about the mapping of the small signal equivalent circuit of CE amplifier on a voltage amplifier.

And, then in typical voltage amplifier it is having three important parameters; how those parameters can be obtained in the small signal equivalent circuit as a methodology we will be discussing as well as this example. And, then subsequently we will be discussing about two numerical examples. So, under the numerical examples we will be having analysis for a given design, we will do the analysis to find the numerical value of gain and operating point of course.

And, then we will be giving some design guidelines for achieving some performance of an amplifier. So, this is what we will be covering today. So, let us talk about the biasing scheme and then let me compare the two biasing scheme.

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Biasing of CE amplifier: Fixed-bias vs. Self-bias

- In Fixed-bias, I_B is defined by R_B (and $V_{CC} - V_{BE(on)}$)
- In Self-bias, I_E and hence I_C are set by R_E (and $V_{BB} - V_{BE(on)}$)

$I_C = \beta_F I_B$ $I_C \approx I_E = \frac{(V_{BB} - V_{BE(on)})}{R_E}$

$R_{BB} \ll R_B$

So, as we have discussed that this is the; this is the fixed bias kind of circuit and here what we have done it is the base current, base terminal current particularly the DC current I_B it is decided by the V_{CC} and then $V_{BE(on)}$ and this R_B . So, in fixed bias circuit the base current it is well defined by the base resistor called R_B and then supply voltage minus base to emitter diode on voltage.

So, once say this base current is defined and that is fixed then the corresponding collector current of the transistor it can be obtained by simply multiplying this I_B with the β_F of the transistor. So, what we have discussed that since I_C it is a direct function of the β_F there may be a situation, in case if the β_F of the transistor it is changing then the collector current directly getting affected.

And, if the collector current is getting affected the drop across this resistance as a result the collector to emitter voltage of the transistor that may vary. So, we can say that operating point of the transistor it was getting heavily affected by variation of the beta of the transistor. In contrast to that we are going to discuss about this circuit which is referred as self bias.

And, in this self bias what we have it is this emitter resistor we are connecting in series with emitter to the ground. And, on the other hand at the base we prefer to give a DC voltage rather of course, we do not want from this voltage should be ideally a DC voltage because we like to feed the signal here.

So, but then the voltage here DC voltage here we want predominantly define in other words we want this Thevenin equivalent resistance of this bias circuit should be as small as possible in terms of the bias stability. So, even if we consider practical value of V_{BB} this R_{BB} it is much smaller than whatever R_B we do have.

So, we can say that the R_{BB} in the self biased circuit it is much smaller than the R_B of the fixed bias circuit. So, we can say that even if say the base current is flowing through this circuit I_B it is flowing through this circuit the drop across this R_{BB} it is very small. As a result we may say that this voltage it is approximately V_{BB} , now once this voltage DC voltage it is remaining almost constant then V_{BB} minus the V_{CE} the base to emitter on voltage.

So, V_{BB} minus V_{BE} on divided by the whatever this R_E that defines the current flow here. So, in other words we can say that this emitter current I_E in the self biased circuit it is defined by this voltage difference and then divided by R_E . So, if R_{BB} it is independent quote and unquote independent of this the beta of the transistor.

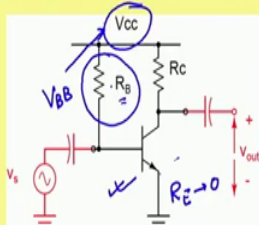
So, naturally the collector current it is also quote and unquote independent of the beta β . So, in this circuit in the self bias circuit in contrast to the fixed bias the emitter current V_{BB} minus V_{BE} on divided by R_E . So, that is this current and we can approximate if the beta is very high we can approximate that the collector current is very close to that.

So, in this expression since beta is not there; so, we can say that the collector current is quote and unquote independent of beta of the transistor so, that is the main purpose here. In other words so, based on our requirement if we fix the value of this R E and then V BB assuming that R BB it is very small, then the collector current it is almost decided. As a result the DC operating point of the transistor it is almost fixed. So, that is the main advantage here. So, let us see what may be the analysis of this circuit to compare these two circuits performance.

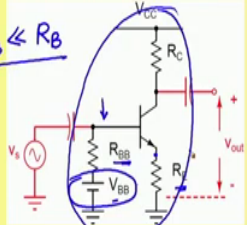
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Biasing of CE amplifier: Fixed-bias vs. Self-bias

- In Fixed-bias, I_B is defined by R_B (and $V_{CC} - V_{BE(on)}$)
- In Self-bias, I_E and hence I_C are set by R_E (and $V_{BB} - V_{BE(on)}$)



$I_C = \beta_F I_B$



$I_C \approx I_E = \frac{(V_{BB} - V_{BE(on)})}{R_E}$

In fact, if you see it carefully what can be seen here that V BB you may consider as a special case it is V CC and R BB it is similar to R B. However, as I said that in this circuit R BB it is much smaller than R B and that can be obtained because we are adding this emitter resistor R E.

So, in contrast to over this circuit where emitter is connected to ground. So, we do have the flexibility to change this voltage and predominantly we can say that this circuit is more like a voltage bias rather than current bias what we are seeing in the fixed circuit.

But, nevertheless we may consider that this circuit is more generalized and the self bias circuit it may be treated as in general it may be treated as a special case where you may say that R_E it is going to 0. And, then V_{BB} is going to be equal to V_{CC} and whatever the R_B is there. So, to compare the expression of the collector current in the two circuits probably we can analyze the circuit and we can find the expression of I_C .

And, then we can probably the through equation we can compare the collector current expression in there in the two circuits. Namely, in the fixed bias circuit the collector current it will be having lot of dependency on the beta of the transistor whereas, for self bias circuit the dependency it will be less. So, let us see the analysis of this circuit here.

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Biasing of CE amplifier (cond..)

• **Analysis to find I_C :**

$$V_{BB} = V_{BE(on)} + R_{BB} \cdot I_B + R_E(1 + \beta_F)I_B$$

$$I_B = \frac{(V_{BB} - V_{BE(on)})}{\{R_{BB} + (1 + \beta_F)R_E\}}$$

$$I_C = \beta_F I_B = \frac{\beta_F (V_{BB} - V_{BE(on)})}{\{R_{BB} + (1 + \beta_F)R_E\}}$$

$$\frac{\partial I_C}{\partial \beta_F} = \frac{I_C}{\beta_F} \cdot \frac{V_{BB} - V_{BE(on)}}{\{R_{BB} + (1 + \beta_F)R_E\}}$$

$$\frac{\Delta I_C}{I_C} = \frac{\beta_F}{\beta_F} \cdot \frac{(V_{BB} - V_{BE(on)})}{\{R_{BB} + (1 + \beta_F)R_E\}}$$

$\frac{\Delta I_C}{I_C} = \frac{[V_{BB} - V_{BE(on)}]}{(1 + \beta_F) R_E}$ for small β_F

So, in the next slide we are going to analyze this circuit in detail. So, what we have here it is we do have the same circuit we have discussed we do have V_{BB} . Now, if I consider say this loop, this loop then we can say V_{BB} equals to V_{BE} on plus drop across this resistance which is R_{BB} multiplied by I_B plus the drop across this emitter register R_E . And the drop here it is of course, the resistance multiplied by the emitter current I_E .

And, I_E it is $1 + \beta$ of the transistor times I_B so, this is the drop across this R_E . So, if we rearrange this equation to get the expression of I_B , now by taking this I_B left side and what we are getting here it is I_B equals to V_{BB} minus V_{BE} on divided by R_{BB} plus $1 + \beta$ into R_E . Note that here there is no approximation.

So, here as I said that if we consider two special cases, the same equation will be helping us to find the expression of the base current. Now, if I am having this base current we can get the

collector current by multiplying with beta of the transistor. So, the expression of the collector current is given here, it is same as this part along with multiplying with this beta.

Now, if we like to see what will be the variation of this collector current with variation of beta, we can take partial derivative of this equation with respect to beta. And, if we assume that this part, this V_{BB} on it is quote and unquote independent quote and unquote independent of beta F though it is not theoretically exactly correct, but practically that is quite consistent. Particularly, since we are not looking at V_{BE} independently rather we are observing $V_{BB} - V_{BE}$ on.

So, even if it is having some variation, but if I consider $V_{BB} - V_{BE}$ on variation with respect to beta that can be ignored. So, if I assume that this part is remaining constant and then if I take partial derivative with this I_C with respect to beta F what we are getting here it is $V_{BB} - V_{BE}$ on divided by this denominator.

Note that here this part it is I_B ; so, the expression of this I_B it has been used here to make a to get a nice form. But, in this form what you can see if we further analyze if I take the change in collector current divided by the collector current; that means, fractional change in collector current divided by change in beta divided by beta.

So, we can say that fractional change in collector current divided away divided by fractional change in beta that is having this expression which is $V_{BB} - V_{BE}$ on divided by $R_{BB} + 1 + \beta$ into R_E . So, if we assume that this part, this part it is much higher than this part.

So, if I consider this R_{BB} it is much smaller then we can say that this is $V_{BB} - V_{BE}$ on; so, divided by $1 + \beta$ into R_E . Now, due to this $1 + \beta$ coming on the denominator, this is since it is inversely proportional with beta F under this condition of course, then we can say that sensitivity of the collector current change with respect to beta F it will be quite small.

On the other hand the same equation; same equation if I consider R_E equals to 0; so, if I consider this is on the other hand if I consider for the fixed bias where this R_E equals to 0 for fixed bias. And, for that the dependency of the collector current with respect to β_F or fractional change in collector current divided by fractional change in β_F it is only this one.

So, as a result you can see that this part it will be quite significant compared to this one which is the fixed bias case. So, or directly maybe you can use this equation and then you can compare the two circuits namely. So, we do have we can replace this part by I_B and then we can say that it is having strong dependency on β_F . So, anyway so, what we like to say here it is we like to bias this circuit in terms of voltage.

And, we like to place this emitter resistor to get the better stability of the operating point; particularly if we are changing the transistor by another one having different β or may be due to the thermal runaway if the β is changing, this circuit it will be having a better stability.

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Self-biased CE amplifier

• **Practical circuit**

$$V_{BB} = V_{CC} \frac{R_2}{R_1 + R_2}$$
$$R_{BB} = \frac{R_1 \cdot R_2}{R_1 + R_2}$$

$$V_{out} \approx \frac{V_{BB} - V_{BE}}{R_E} R_C$$

Now, what is the practical circuit let us see. So, this is what the practical circuit you can see that instead of having independent voltage at the bias voltage at the base what we have here it is potential divider from V_{CC} by this R_1 and R_2 . So, the voltage coming here if I consider Thevenin equivalent voltage source of this one along with the V_{CC} what we can get is V_{BB} it becomes V_{CC} multiplied by R_2 divided by R_1 plus R_2 .

And, then Thevenin equivalent resistance R_{BB} what we have drawn in the previous circuit R_{BB} it is parallel connection of this R_1 and R_2 or you can say R_1 multiplied by R_2 divided by R_1 plus R_2 ; so, that is the expression of R_{BB} . So, as I said that this is the practical implementation of this self bias and of course, we do have the R_E part emitter resistor that connected here.

So, the typically this circuit is quite popular and then we do use this emitter resistor to avoid the thermal runaway problem or the dependency of the operating point on the beta. Now, in this circuit similar to fixed bias the voltage at this point it is having a DC voltage defined by this R_1 and R_2 and then V_{CC} . Practically, this V_{BB} it is defining this DC voltage and then on top of that we are feeding the AC signal through this coupling capacitor. So, we do have the AC signal riding over there. So, we do have the AC signal riding over this one. So, the voltage here it is having a DC voltage and then on top of that we do have the signal.

Now, at the output similar to the previous case the it is also having a DC voltage level. So, this DC voltage level at this point particularly at this point it is having the drop across this R_E plus whatever the V_C voltage you do have or we may say that V_{CC} minus the drop across this resistance that gives the this DC voltage. So, we can say this DC voltage it is V_{CC} minus R_C into I_C . And, then to get the I_C we have discussed that once this voltage it is given here then that minus V_{BE} on.

So, that divided by R_E that gives the emitter current and that actually it gives the approximately the collector current. So, I_C you may say that this is approximately equal to V_{BB} minus V_{BE} on and divided by this R_E so, that is the emitter current; so, approximately you can consider. So, that is how we are getting DC voltage and once we have the AC signal at the base coming from the signal source naturally this is also having the small signal current I_C along with the DC current.

And, as a result it is also producing a drop across this resistance and hence we do have the signal at the output. Only thing is that since we are subtracting from the DC voltage this drop across the R_C we are subtracting from the V_{CC} . So, the signal here it is having 180 degree phase difference with respect to whatever the signal we do have. So, in this circuit let me go a little detail of analyzing the input part and the output part namely the input port.

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Self-biased CE amplifier

• Practical circuit

$$V_{BB} = V_{CC} \frac{R_2}{\{R_1 + R_2\}}$$
$$R_{BB} = \frac{R_1 \cdot R_2}{\{R_1 + R_2\}}$$

↓ input port

So, if I consider input port is this one DC wise, in addition to that we do have the signal. And, the output port we do have on the other hand consist of on the collector resistor, transistor and also this emitter resistor. So, you need to see that this emitter resistor it is part of the input, input port as well as the output port as well as the output port.

So, in the input port while we will be analyzing we need to be careful when whenever we will be talking about the I_B current; we need to see how much the current actually it is flowing. In fact, the current here it is not only I_B it is flowing through this one, but also the I_C . So, whenever we are writing say I_B here the current here it will it is 1 plus beta times I_B ; so, that we need to be careful.

So, and then of course, then we will be talking about the small signal thing. So, let us see the input port circuit what we are discussing here along with the signal part.

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CE amplifier – Self-bias (contd.)

• **Circuit at Input port- Large signal**

$$V_{BB} = V_{CC} \frac{R_2}{R_1 + R_2}$$

$$R_{BB} = \frac{R_1 \cdot R_2}{R_1 + R_2}$$

So, at the input port what we have here it is the so, this is the; this is the circuit we are talking about, let me mark it here. So, at the input port we are talking about this one. So, we do have the R 1 and R 2 coming in parallel to give us the R BB. So, R BB which is parallel connection of the R 1 and R 2 and then V BB it is the voltage Thevenin equivalent voltage coming from V CC and R R 1 and R 2 together.

So, we do have V BB it is having the expression earlier we have discussed about that V CC multiplied by R 2 divided by R 1 plus R 2. So, then out of the transistor from base to emitter

we do have the base emitter diode and then at the emitter node. So, this is the base node and this is the emitter node and at the emitter node we are having this emitter resistor.

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CE amplifier – Self-bias (contd.)

• Circuit at Input port- Large signal

$$V_{BB} = V_{CC} \frac{R_2}{R_1 + R_2}$$

$$R_{BB} = \frac{R_1 \cdot R_2}{R_1 + R_2}$$

$$V_{BB} - V_{BE(on)} = I_B (R_{BB} + r_{\pi}) + I_B (\frac{1}{\beta}) R_E$$

$$I_B = \frac{V_{BB} - V_{BE(on)}}{(1 + \beta) R_E}$$

$$I_C = \beta I_B \approx \frac{\beta (V_{BB} - V_{BE(on)})}{(1 + \beta) R_E}$$

Now, at the emitter of course, we do have the; we do have the additional current flowing into this. So, whenever we will be talking about I_B it is flowing here, we need to consider that I_C also which is beta times I_B as a result the current flowing through this part it is 1 plus beta times I_B . So, if it is DC current will be talking about 1 plus beta F times I_B .

Now, if I consider if I replace this diode base emitter diode by its equivalent circuit shown here. So, which is parallel connection of the V_{BE} on voltage and then small r_{π} and again this node we need to consider that collector current is coming here which is beta times I_B .

So, now, if we analyze; if we analyze this circuit if we analyze this circuit what we can get here it is the I_B . So, rather $V_{BB} - V_{BE}$ equals to I_B times R_{BB} plus r_{π} and then plus I_B multiplied by $1 + \beta$ times R_E . So, here we may or may not be able to sorry I need to consider this is R_{BB} .

So, in this case we may or may not be able to ignore this R_{π} in this case. So, we may have to consider this entire portion, but if we consider say drop across these two resistances; since R_{BB} it is much smaller than the R_E part then typically compared to this part we may ignore this part. So, we may say that I_B equals to $V_{BB} - V_{BE}$ divided by $1 + \beta$ into R_E .

In fact, in our previous analysis we have ignored this part compared to this one, but even if you consider this whole thing these two together it can be ignored compared to this part. So, as a result what is the expression of the current I_B we are getting? It is $V_{BB} - V_{BE}$ divided by $1 + \beta$ into R_E . So, large signal analysis if we do then we can find the corresponding base current, whenever we will be talking about some numerical example we will see that this current it will be very small.

And, once we get this βI_B then the collector current it is I_C equals to β into I_B . So, that we can approximate that $V_{BB} - V_{BE}$ divided by $1 + \beta$ into R_E and then whole thing multiplied by β . So, we may ignore we may ignore this one part and then we may cancel this part and this part.

So, that gives us the collector current it is independent of β ; so, that is what again we are converging to the same point. So, that is what we do, the input port we do this large signal analysis. Only thing as I said that we need to be careful that this R_E it is also taking the current of the output port. Now, from this circuit after getting the large signal current probably we can do the small signal analysis.

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The slide, titled "CE amplifier - Self-bias (contd.)", illustrates the small signal analysis of a common-emitter amplifier. It features three main diagrams: 1) A DC biasing circuit with a base resistor R_{BB} connected to V_{BB} , a base-emitter junction, and an emitter resistor R_E connected to ground. A signal source V_s is shown at the input. 2) A small signal equivalent circuit where the base-emitter junction is replaced by its dynamic resistance r_{π} . The base node is connected to the signal source V_s and the base resistor R_{BB} . The emitter node is connected to the emitter resistor R_E and is AC-grounded. Handwritten notes include "Small signal eqm. Ckt" and $(1+\beta)i_b$ flowing through R_E . 3) A complete AC equivalent circuit showing the base node connected to V_{BB} (AC ground), R_{BB} , r_{π} , and the signal source V_s . The emitter node is connected to R_E and AC ground. The collector node is connected to R_C and AC ground. A graph of V_{in} vs t shows a sinusoidal input signal. A small inset diagram shows the full amplifier circuit with V_{in} at the base and V_{out} at the collector. A video feed of a presenter is visible in the bottom right corner.

So, I do have a different slide for that. In fact, the same circuit we are drawing here, but what we are talking about; we are going to talk about the small signal analysis. So, once we have this entire circuit and we like to see what will be the corresponding small signal equivalent circuit; equivalent circuit.

What you have to do here we have to make it AC ground, we have to drop this part and then we have to short it. So, what we have it is we do have a signal source shorted to the base node. So, this is the base node this is of course, the emitter node. And, then we do have the AC ground and then we can keep this R_{π} and then we do have the R_E , but again we need to be careful that we also have the collector current to be considered.

So, if I say collector current is I_C that is also flowing to this R_E . So, if we have say this current we do have say small i_b . So, this collector current it is beta times beta naught times i_b .

b. So, as a result the current flowing through this circuit it is $1 + \beta$ times i_b into R_E ; so, that is the drop.

So, probably you may drop this current source, you may drop this part and then instead you may simply say that this registered it is getting multiplied by $1 + \beta$ of the transistor. And so, then we may forget about this part, you may say that the resistance is getting increased; so, that what normally it is followed.

But, whatever the way you feel we need to be careful that we need to consider this collector current or probably the emitter register we need to consider its amplified version of the emitter register. Now, this V_s it is coming here and it is going to the base terminal. Note, that this is emitter terminal so; obviously, V_s it is not same as V_{BE} as it was for fixed bias.

It is rather having a voltage drop here which is called base to emitter voltage different from this V_s . Now, let us look into the output side. So, whenever we will be talking about the small signal equivalent circuit, we must consider this analysis. So, while will be going to the small signal equivalent circuit for the entire emitter amplifier common emitter amplifier then we have to consider this circuit. So, will be coming back, but for the time being let we consider the output port. And, again for this output port what we have it is emitter register it is common.

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CE amplifier – Self-bias (contd..)

- **Circuit at Output port: Large signal and small signal**

The diagram illustrates the self-biasing of a common-emitter (CE) amplifier. It shows a BJT with a base bias network consisting of resistors R_1 and R_2 connected to a supply voltage V_{CC} . The collector is connected to V_{CC} through a resistor R_C , and the emitter is connected to ground through a resistor R_E . A decoupling capacitor C is connected between the collector and emitter nodes. The output voltage V_{out} is taken from the collector. The diagram highlights the DC biasing and the AC signal components. The collector current is labeled as $I_c = I_{CQ} + i_c$, where I_{CQ} is the DC collector current and i_c is the AC signal component. The output voltage is shown as a superposition of a large signal and a small signal. The voltage across the collector resistor is labeled as $R_C \times I_d$.

So, let me go to the; so, here again we do have the large signal and a small signal notion. And, here we do have the; here we do have the actual circuit and at this point we do have the voltage shown here and we are going to talk about at this circuit along with this DC decoupling capacitor.

So, what we have at the collector? We do have the R_C , collector resistor connected to the V_{CC} and then collector to emitter we do have the current. And if it is of course, it is having a DC current I_{CQ} plus in case if it is having small signal small i_c also.

But, whatever it is this entire current it is flowing through this emitter resistor and in addition to that at this node the base current is also coming. However, this base current it is small

compared to this collector current. So, we may or may not ignore this part that we can see, but that this is what the output port.

And, once to find the DC operating point one say I_C it is known by analyzing the input port then you can say that current is flowing here. And, the DC voltage coming here which is V_{CC} minus R_C multiplied by this I_C ; so, that is the DC voltage coming here.

So, the voltage at the output node or the collector node rather it is V_{CC} minus this drop; so, this is V_{CC} minus R_C into I_C . Now, once you obtain the DC voltage here let us look into the small signal part.

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CE amplifier – Self-bias (contd.)

- **Circuit at Output port: Large signal and small signal**

The slide displays three main diagrams:

- Full Circuit Diagram:** Shows a common-emitter amplifier with a base biasing network consisting of resistors R_1 and R_2 connected to V_{CC} and ground. The collector is connected to V_{CC} through resistor R_C , and the emitter is connected to ground through resistor R_E . An input signal V_s is applied to the base, and the output V_{out} is taken from the collector. A graph shows a sinusoidal V_{out} waveform over time t .
- DC Operating Point Analysis:** Shows the collector current I_C flowing through R_C , resulting in a voltage drop $I_C R_C$. The DC output voltage is $V_{CC} - I_C R_C$.
- Small-Signal Equivalent Circuit:** Shows the AC equivalent circuit where V_{CC} and ground are AC grounds. The collector current is represented by a dependent current source $i_c' = g_m v_{be}$ in parallel with R_C and R_E . Handwritten notes indicate that the output voltage is $V_{out} = -i_c'' R_C = -g_m v_{be} R_C$.

And similar to the input port here again for small signal what are the things we will be doing, it is the this terminal will be will be considering AC ground. So, this is AC ground and here

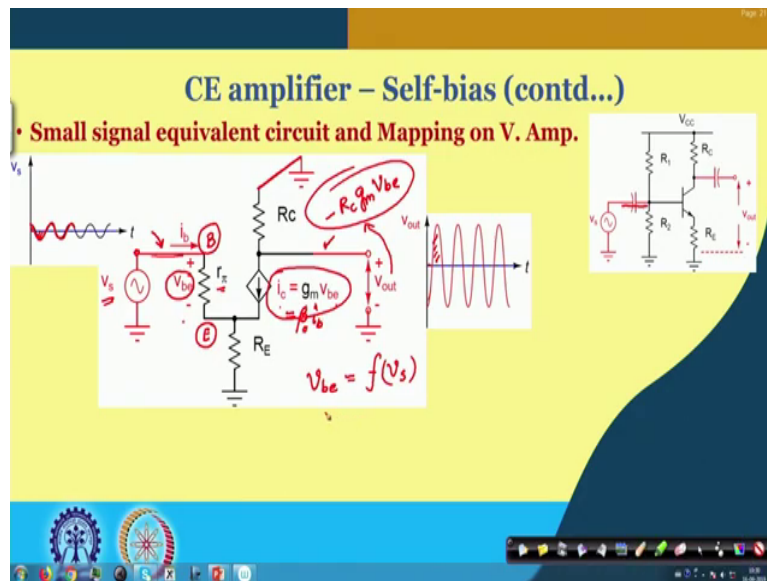
this DC part it will be removed. So, though it is having total current is I_C plus small i_c , but this will be dropped and whatever the current it will be flowing through this one it is only small i_c ; so, this is also small i_c . And, the corresponding capacitor here of course, it will be getting shorted.

So, whatever the voltage small signal voltage you are getting that will be called small signal output voltage, while we are dropping this DC part. So, if I draw the small signal equivalent circuit, we do have the R_C connected to AC ground and then we do have the small signal current, small i_c . And, then at the emitter we do have the R_E maybe we can think of the base current coming here i_b and then at the output we can say this is the open circuit output signal.

So, this i_c of course, we may be having a different expression. One of them it is i_c equals to transconductance of the transistor multiplied by base to emitter voltage and since this is ideal current source; so, we may to find this output voltage we may prefer to see how much the drop across this resistor is appearing while small signal current i_c it is flowing.

So, the output voltage at this point output rather signal it is minus i_c into R_C which is of course, this is equal to minus g_m into V_{BE} into R_C . So, while we will be combining this small signal equivalent circuit at the output port and small signal equivalent circuit at the input port, then we can get the combined small signal equivalent circuit.

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So, let us see what you are going to get here yes. So, we do have this small signal equivalent circuit shown here, where now if you see at the collector side we are making AC ground first thing. The capacitor here we are shorting; so, that is what we are doing and then the collector current it is having only the small signal collector current, small i_c which is g_m times V_{be} .

And, the V_{be} it is the voltage drop between the base terminal and the emitter terminal ok. So, this is the voltage drop across you can say that this is drop across R_{π} ; so, of course, we are giving the signal here. So, the voltage coming at this point in this circuit since we are eliminating the DC we do have plus and minus kind of signal and the output here we are removing the DC.

So, we are getting the only the signal part and in this illustration as I said that the signal here it is in opposite phase of the input signal that is because of the voltage coming here it is minus

R_C into g_m into V_{be} . Now, if you see here this V_{be} of course, it is not same as V_s ; so, this is not V_s .

So, since the V_{be} it is not same as V_s , it is rather only a part of it we need to do the detailed analysis of the circuit. Now, we are going to take a small break, but just to after we come back we need to find what will be the relationship between this V_{be} and the V_s . So, V_{be} as function of V_s and while we will be doing this we may use this important relation that this i_c we may say that this is g_m into V_{be} or, we may say that this is β times i_b .

Anyway we will come back and we will be having little more discussion to find the output voltage.