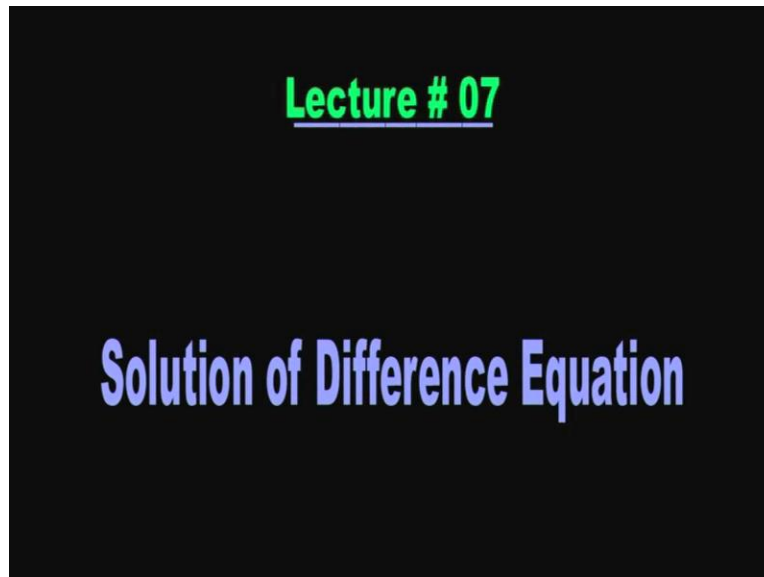


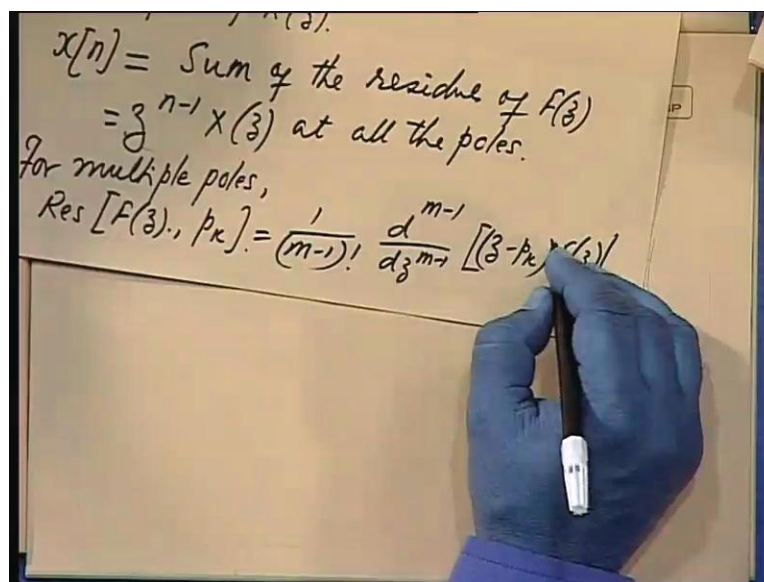
Digital Signal Processing
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Lecture - 7
Solution of difference Equation

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Before we start, before we start I just like to point out a small slip, in the last class there should have been the derivative that you take for multiple poles; this should have been multiplied by m, please make the corrections, z minus p K to the power m. In the example, you have used this, it was a small slip. We are discussing in the last class; the inverse DTFT from the given frequency domain representation, as given by this expression.

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A photograph of a hand holding a pen over a piece of paper with handwritten mathematical equations. The equations are:

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) \cdot e^{j\omega n} d\omega$$

$$= \frac{1}{2\pi} \int_{-\pi}^{\pi} \sum_{k=-\infty}^{\infty} x[k] \cdot e^{-j\omega k} \cdot e^{j\omega n} d\omega$$

The paper has a small logo in the top right corner that reads "© CET I.I.T. KGP".

A photograph of a hand holding a pen over a piece of paper with handwritten mathematical derivations. The equations are:

$$= \sum_{k=-\infty}^{\infty} x[k] \int_{-\pi}^{\pi} e^{j\omega(n-k)} d\omega$$

$$= \sum_{k=-\infty}^{\infty} x[k] \left\{ \frac{\sin(n-k)\pi}{(n-k)\pi} \right\}$$

Below the last equation, there is a downward arrow pointing to the text:

$$= \begin{cases} 1 & n = k \\ 0 & n \neq k \end{cases} \delta[n-k]$$

So, $x[n]$ will be this discrete line. Now, how do you know, that this is going to give you the same sequence, you get my point? Given a sequence $x[n]$, if I take its Fourier transform by this integral, do you get back the same sequence? So, let us see let us verify this relation 1 by 2π i minus π i to plus π i ; this is nothing but $\sum_k x[k] e^{-j\omega k}$ into $e^{j\omega n}$, $d\omega$, okay.

You may write this as, $\sum_k x[k] \int_{-\pi}^{\pi} e^{-j\omega k} e^{j\omega n} d\omega$, okay. So, that gives me if we integrate this, we will get after substitution of these values $\sin(n-k)\pi$ by $n-k$ into π . And this is; this particular function will be equal to 1 when n is equal to k and equal to 0 otherwise.

So, this gives me $\sum_k \delta[n-k] x[k]$. So, basically that gives you the series as $x[n]$, basically; this is a delta function n is equal to k and n not equal to k , this function is basically $\delta[n-k]$. So, $x[k]$ multiplied by $\delta[n-k]$ is $x[n]$ itself, okay. Now, we take some difference equations and see what will be the solution in the normal time domain. How to compute by the normal classical method?

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Handwritten notes on a whiteboard showing a difference equation and its solution. The equations are:

$$y[n] = a_0 x[n] + a_1 x[n-1] + \dots + a_N x[n-N] + \underbrace{a_{-1} x[n+1]}$$

Causal Sys.

$$\sum_{k=0}^N a_k y[n-k] = \sum_{k=0}^M b_k x[n-k]$$

$$y[n] = \sum_{k=0}^M b_k x[n-k] - \sum_{k=1}^N a_k y[n-k]$$

$$\sum_{k=0}^{\infty} a_k y[n-k] = \sum_{k=0}^{\infty} b_k x[n-k]$$

$$y[n] = \sum_{k=0}^M b_k x[n-k] - \sum_{k=1}^N a_k y[n-k]$$

$y[n]$ = Auto. reg. + Moving Av.
ARMA (N, M)

You have say a $0 \times n$ plus $y[n]$ is equal to a $0 \times n$ plus a $1 \times n$ minus 1 plus and so on, $A \times n$ minus k , all right. Now, the output of the see for a sequence like this; suppose you have a term like a minus $1 \times n$ plus 1 and so on, suppose the $x[n]$ sequence is considered in the left hand side of the time domain, can you have an output? Can you have an output which will be a function of the previous inputs, as well as inputs which are yet to come?

It is not possible, inputs which have not yet come how can the system response, identify that, all right? So, that will be a non-causal system, okay. So, causal system means, this kind of a term will not be present; for $y[n]$ I cannot have on the right hand side, the dependence of $y[n]$ on $x[n+1]$ plus 2 , okay.

Sometimes we try find out, some kind of a relationship between the present value and the future values; that is possible only when you are having offline data that is, the history is already available, so what is going to come say after one hour that is already known to you, that record is there.

So, can you correlate with the present value, if it is possible? There are now causal systems also; you take an image all right, you focus a light here, now it is not time domain function, it is phase domain function, means phase domain this is plus all right. So, a light here will be influencing the illumination here, here, here, it will also be influencing the illumination on this side, all right.

So, if I consider in-space these as positive side then this as negative side, then in that case at any instant of time; we are not considering time as such, the illumination of a particular of a particular point will be depending also on its previous values, I mean the location can be on both sides, that what I mean.

So, there we can have functions with z to the power plus 1, if you consider just one dimensional; say one line of a particular object, of an image, then if this is z to the power minus 1, the normal backward shift operator then this will be forward shift operator z to the power of plus 1, so both are possible. So, this is that way something like a non-causal system, otherwise all natural processes are causal systems.

Now, suppose we have a difference equation; $k y$ on this side which is the summation of weighted values like this, equal to $b_k x_{n-k}$, k varying from say 1 to M , k varying from say, 0 to M and here it is 0 to N , okay. You can write y solution y_n , as $b_k x_{n-k}$ summation minus k varying from 0 to M minus $a_k y_{n-k}$, k varying from 1 to N , okay.

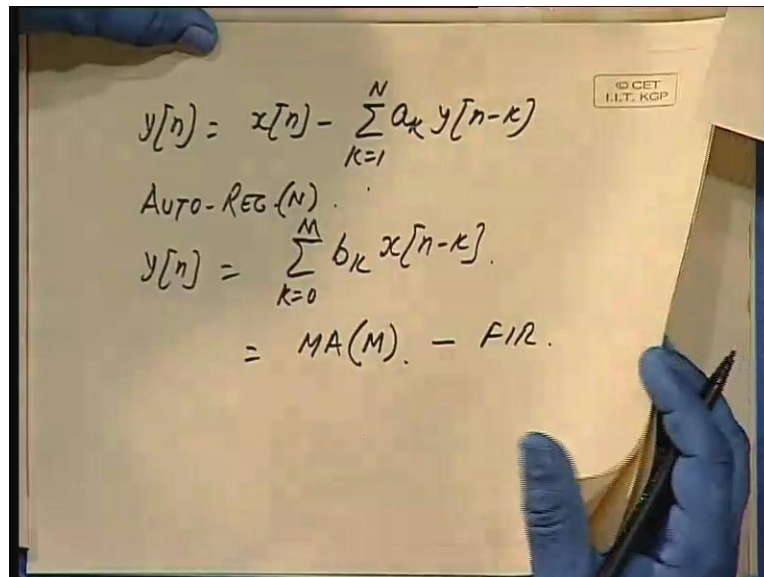
So, when you have a general expression like this; that is y_n depends on the previous values of y , y_{n-1} , y_{n-2} because k starts from 1, $n-1$, $n-2$ and so on and some values of the inputs then that is known as auto-regressive process, okay. Auto-regressive and moving average, moving average means; it is a weighted mean of the inputs average of the input, average in some sense.

You give more weightage to the present value of the input, a little less weightage to the previous input and so on, but the previous inputs also have an influence in the present on the present

output. So, this is a moving average part and it is regressing on itself, its previous values, so it is autoregressive.

So, it is auto-regressive and moving average more commonly known as, ARMA process, all right. ARMA, auto-regressive part what is the order? N, ARMA N, moving average, what is the order? M. So, we write ARMA N M, it is an ARMA, M N, N M process or somebody may choose P Q, ARMA, P Q process.

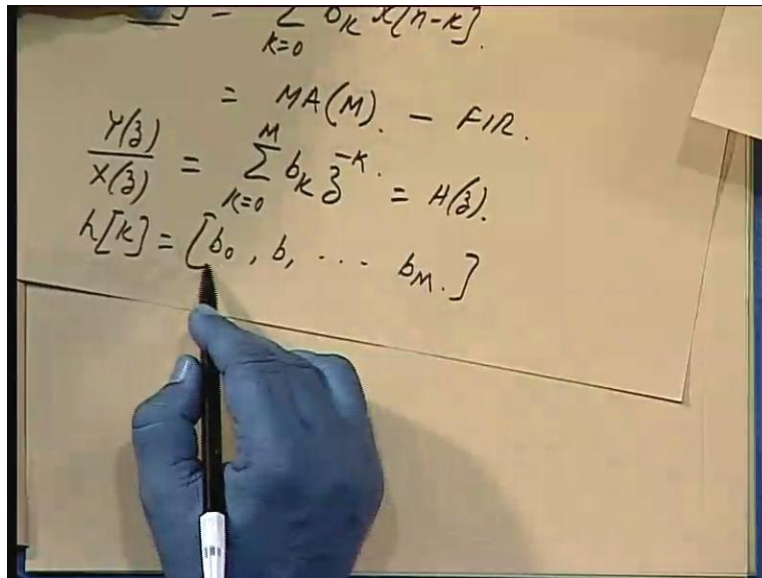
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$$y[n] = x[n] - \sum_{k=1}^N a_k y[n-k]$$

Auto-REG(N)

$$y[n] = \sum_{k=0}^M b_k x[n-k]$$

= MA(M) - FIR.



When you have y_n , as just one term of x_n and other terms of y ; we call it simple autoregressive process, there no moving average, just one input is that okay. When you have y_n dependent on only inputs, it is called moving average process, all right; moving average of order M , this is autoregressive of $A R$ process of order N and this also called FIR okay, Finite impulse response.

Suppose we have, suppose we have a system like this; will come to this very soon, a system like this, what will be say in the Z domain, what will be the transfer function? $Y z$ by $X z$, it will be summation $b_k z$ to the power minus k , k varying from 0 to M , okay. So, this is $H z$, is it not? So, what will be h_K impulse response? It will be b_0, b_1 up to b_K, b_M okay.

So, if I excite the system if excite the system, with an impulse then this system; if I give an impulse as input then it will give you output up to b_M and after that it will be 0 . So, it is an infinite length, so that is that is why it is called finite impulse response system, FIR system. It is same as, in mathematics we called it MA , moving average module, it one and the same thing.

If on the other hand, if you have a system like this that is $Y z$ by $X z$ comes out as 1 by some $A z$, that is basically 1 by 1 plus a $1 z$ to the power minus 1 plus a $2 z$ to the power minus 2 and so on, the polynomial $a_N z$ to the power minus N , okay.

(Refer Slide Time: 14:17)

$$\frac{X(z)}{A(z)} = \frac{1}{1 + a_1 z^{-1} + a_2 z^{-2} + \dots + a_N z^{-N}}$$

IIR.
Ex.
$$H(z) = \frac{1}{1 - 0.8z^{-1}}$$

$$Y(z) = \frac{X(z)}{1 - 0.8z^{-1}}$$

Ex.
$$Y(z) = \frac{X(z)}{1 - 0.8z^{-1}}$$

$$Y(z) = \frac{X(z)}{1 - 0.8z^{-1}}$$

$$y[k] - 0.8y[k-1] = x[k].$$

$$y[k] = 0.8y[k-1] + x[k].$$

If you get something of this type then it is called an IIR function, infinite impulse response. Let us see, what it means infinite impulse response. Let us take just one term, suppose $H(z)$ equal to $1 / (1 - 0.8z^{-1})$, it can be plus or minus, it really does not matter. Then $Y(z)$ is equal to $X(z) / (1 - 0.8z^{-1})$ okay or multiplied by this $Y[k] - 0.8Y[k-1] = x[k]$, if I multiply by this and then take the inverse, this will be the difference equation.

So, Y_k is a solution, I can write $0.8 Y_{k-1} + X_k$, this is a difference equation. Now, I want to see its impulse response; that is X_k is δ_k , what will be the response, okay.

(Refer Slide Time: 16:30)

$x[k] = \delta[k]$
 $x[0] = 1, x[1], x[2], \dots = 0.$
 $y[0] = 0.8y[-1] + x[0]$
 $= 0 + 1 = 1$
 $y[1] = 0.8 \times 1 + 0 = 0.8.$
 $y[2] = 0.8 \times 0.8 + 0 = 0.64$

IIR

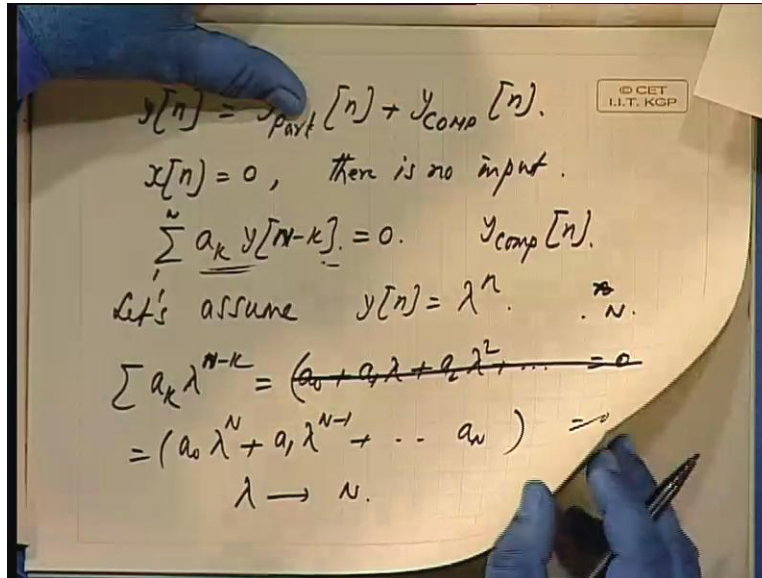
So, X_k is δ_k , that is x_0 is 1 then x_1, x_2 etcetera are all 0's, is it not? You are giving just an impulse at k equal to 0 and after that it is all zero, so what will be y_0 ? If the system is initially unexcited then there is, if it is a causal system; before the input is given there was no output, so y at minus 1 is 0, okay. So, y_0 will be 0.8 into y at minus 1 which is 0 plus x at 0, which means 0 plus x at 0 is 1. So, y_0 will be 1.

What will be y_1 ? 0.8 , it will be now y_0 which is 1 plus x at 1 is 0, so that will be 0.8 . What will be y at 2? It will be 0.8 into again 0.8 plus 0. So, 0.64 , so it will continue to give you output which will be gradually diminishing in magnitude but it continues theoretically, it continues upto infinite. It does not stop at a definite point.

You can approximate, you can approximate truncate it somewhere, but otherwise it is an infinite sequence; so that is why it is called IIR, this kind of response is infinite impulse response. Now, what will be the response of any system, in any difference equation that we have written of this

kind? A difference equation of this type, causal system; what will be the response, due to any general input? So, the input whatever we give will decide the output but apart from that like you have done in mathematics for a differential equation, there are two parts two components or solution, a particular integral and a general, okay.

(Refer Slide Time: 19:16)



So, y_n will have y particular plus y complimentary, okay. So, y complimentary when there is no input, when x_n is 0 there is no input; there is no input you calculate a $K y_n$ minus k to the right hand, it is all 0. So, this solution is the complimentary solution and why particular there is particular solution will be due to x_n itself.

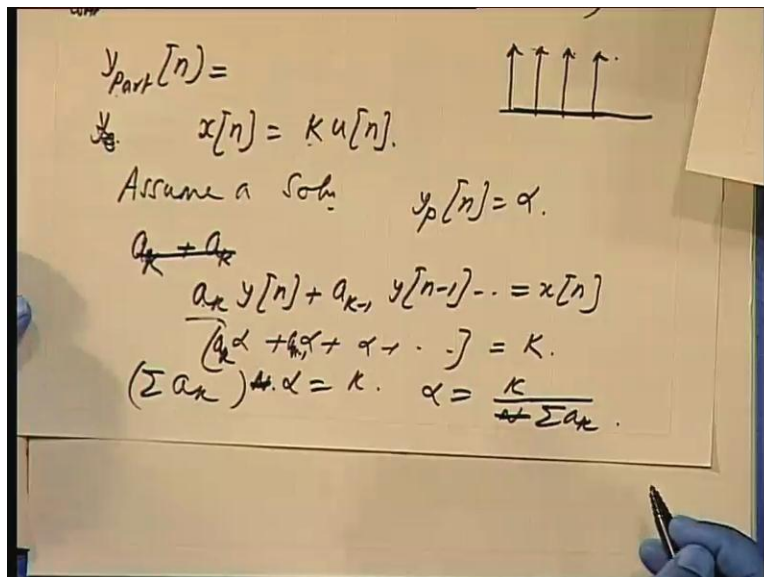
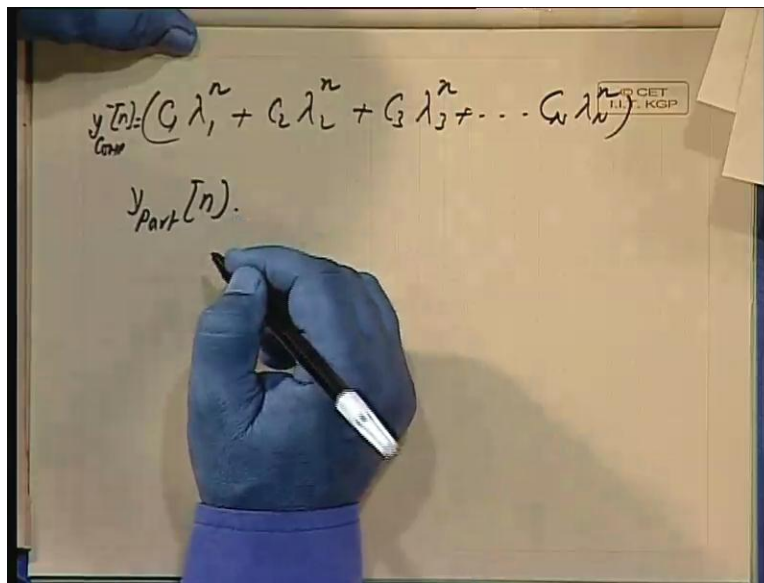
So we again start with an assume solution here. So, let assume a solution y_n is equal to lambda to the power n , okay. So, substitute it here. So, a K lambda n minus k summation, all right which will be a 0 plus a 1 lambda plus a 2 lambda squared and so on, all right. And that is equal to 0, you get the solution for lambda, is this all right, or other way?

It is lambda to the power n , so okay a 0 lambda to the power n , so it will be a 0 lambda to the power N ; it should be capital N because it is going upto a certain, see this n is a variable n and

this N is a finite length, since we are considering a finite length of this sequence, 1 to capital N , so I should have taken N minus k , okay.

So, it will be λ^N minus 1 and so on up to a N , that is equal to 0; so you can calculate the roots of λ , there are N roots all right. So, once you know the roots, linear combination of the roots will be the solution.

(Refer Slide Time: 22:37)



So, $C_1 \lambda_1^n + C_2 \lambda_2^n + C_3 \lambda_3^n$ and so on; $C_N \lambda_N^n$ will be solution Y complimentary, λ_1 to the power n λ_2 to the power n λ_3 to the power n, because our assumed solution was λ to the power n as y^n , okay. Y particular integral, particular solution will be depending on the type of input, if you give a sinusoidal input correspondingly output will also be sinusoidal, if you give a constant input like an impulse, like a step then output will also be a constant, okay.

So, you once again you assume a solution corresponding to the type of input that you give; if it is not a regular input then of course it becomes a little difficult. So, Y will be depending on, say we give an input of step x^n is a some k times u^n then correspondingly; we can assume a solution α because it is a constant, so output corresponding to this part will be a constant.

So from the equation, that you have already got say; a 0 we wrote a k, all right, a k a k plus a sorry a k y^n plus a k minus 1 y^{n-1} and so on. This will be all α plus α plus α , so on because they are all constant, n terms should be equal to the given value k into u^n k, okay. So, N times α is k, so α you have got k by N.

The input has been given as an constant magnitude, step function, all right; so all the time it will be equal to k. So in the difference equation, this is given as x^n which is k times u^n , so you get with of course with sorry sorry a k minus 1 etcetera, I am sorry. So, it will be summation a k into α is equal to k, so α is summation a k, all right. Let us workout a small example, it will be clear. Once you have got this, try to use it with the initial conditions, try to match with the initial conditions, you can calculate all the constants.

(Refer Slide Time: 26:49)

$y[n] + 5y[n-1] + 4y[n-2] = x[n]$
 $y[-1] = 2, y[-2] = 1$
Determine the response $x[n] = 4u[n]$.
I $x[n] = 0$.
 $y_{\text{comp}}[n] \Rightarrow \lambda^2 + 5\lambda + 4 = 0$
 $\lambda = -4, -1$
 $y_{\text{comp}}[n] = C_1(-4)^n + C_2(-1)^n$

We have a system, $y[n] + 5y[n-1] + 4y[n-2] = x[n]$, all right. y at minus 1 is equal to 2; y at minus 2 is equal to 1, determine the response corresponding to $x[n]$ is equal to 4 times $u[n]$, okay. So, let us first consider when $x[n]$ is 0 that is a complimentary solution. y complimentary will be, we assume λ to the power n , so the solution for that will be $\lambda^2 + 5\lambda + 4 = 0$, okay equal to 0.

This is also known as characteristic equation; λ is equal to minus 4 or minus 1, okay. So, the solution will be some C_1 times minus 4 to the power n plus C_2 times minus 1 to the power n into of course $u[n]$, okay.

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The whiteboard shows the following handwritten work:

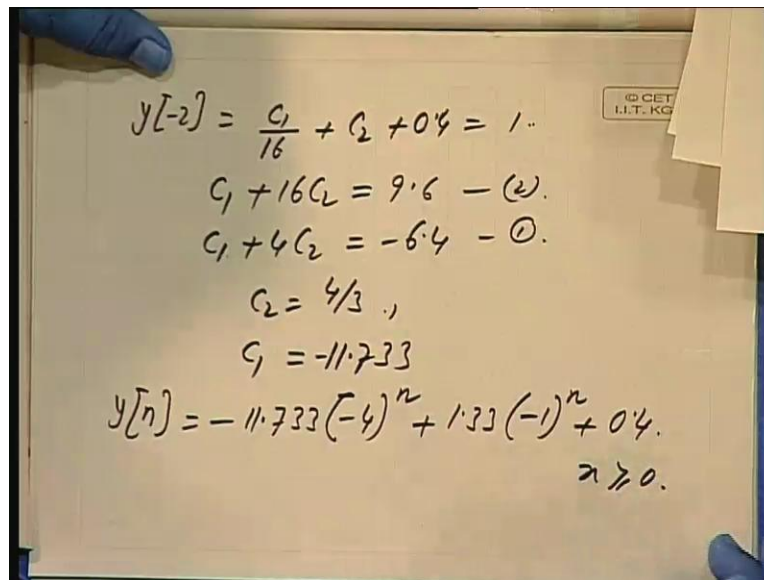
$$\begin{aligned} \text{ii)} \quad y_p[n] &= \alpha \\ \alpha + 5\alpha + 4\alpha &= 4 \\ \alpha &= \frac{4}{10} \\ y[n] &= C_1[-4]^n + C_2[-1]^n + 0.4 \\ y[-1] &= C_1(-4)^{-1} + C_2(-1)^{-1} + 0.4 = 2 \\ -\frac{C_1}{4} - C_2 &= 1.6 \\ C_1 + 4C_2 &= -6.4 \quad \text{--- (1)} \end{aligned}$$

Next, we take y particular solution as some constant α because you are exciting it by, a step function then α plus 5 into α plus 4 into α will be equal to 4, is it not? So, α plus 5 into α plus 4 into α is equal to 4, α is equal to 4 by 10, all right. Now, $y[n]$ is equal to C_1 into minus 4 to the power n plus C_2 minus 1 to the power n plus 0.4, because it is a complimentary solution plus particular solution, all right.

Substitute the initial values, initial condition are initial conditions are y at minus 1, what be the value at minus 1? C_1 minus 4 to the power minus 1, N is equal to minus 1 plus C_2 minus 1 to the power minus 1 plus 0.4 and that is given as 2; this is minus of 4, so minus C_1 by 4 minus C_2 is equal to 2 minus 0.4, that is 1.6, okay.

If you simplify C_1 plus 4, C_2 is equal to minus 6.4, okay this is one equation. We substitute in the second equation, second condition.

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The whiteboard shows the following handwritten work:

$$y[-2] = \frac{C_1}{16} + C_2 + 0.4 = 1$$
$$C_1 + 16C_2 = 9.6 \quad \text{--- (2)}$$
$$C_1 + 4C_2 = -6.4 \quad \text{--- (1)}$$
$$C_2 = \frac{4}{3}$$
$$C_1 = -11.733$$
$$y[n] = -11.733(-4)^n + 1.33(-1)^n + 0.4$$

$n \geq 0$

So, y at minus 2 is similarly minus 4 to the power minus 2; so it will be 1 by 16 plus 1 by 16 minus 1 to the power minus 2, is again plus 1. So, C_1 by 16 I am just writing the final values 0.4 and that is equal to 1 or from this C_1 plus 16 C_2 is equal to 9.6, 0.4 if I bring it to this side, it becomes 0.6 0.6 into 16.9.6, this a second equation.

First equation was C_1 plus 4 C_2 equal to minus 6.4 okay; this is was the first equation. So, if you subtract one from the other, C_1 will get cancelled. So, 12 C_2 will give me 9.6 plus 6.4. So, C_2 becomes 4 by 3, checks whether, is it all right, okay. And then this is approximately 11.733, okay with a negative sign, C_1 is negative.

So the solution will be $y[n]$ equal minus 11.733 minus 4 to the power n plus 1.33 minus 1 to the power n plus 0.4, this is a general solution for n greater than equal to 0, is it all right. Whether same thing you can solve by Z transform method also, that we have already seen.

Sometimes when you study signals, sometimes you want to compare one signal with other; especially in communication. Some signals are coming, you want to check whether it is matching with some other signal or not. So, you have got the sample values, you try correlate them, so it

define a correlation function between two signals. If it is between two different signals say, $x[n]$ and $Y[n]$ we call it cross correlation. If it is a signal with it itself then it is auto-correlation

So, cross correlation we define r_{xy} of lag l as, $\sum_{n=-\infty}^{\infty} x[n] y[n-l]$; l equal to 0 plus minus 1 plus minus 2 and so on, l can have lag positive or negative.

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Correlation

$$r_{xy}[l] = \sum_{n=-\infty}^{\infty} x[n] \cdot y[n-l] \quad l = 0, \pm 1, \pm 2, \dots$$

lag 'l'

$$r_{yx}[l] = \sum_{n=-\infty}^{\infty} y[n] \cdot x[n-l] = r_{xy}[-l]$$

lag l

$$r_{yx}[l] = \sum_{n=-\infty}^{\infty} y[n] \cdot x[n-l] = r_{xy}[-l]$$

Auto-correlation

$$r_{xx}[l] = \sum_{n=-\infty}^{\infty} x[n] \cdot x[n-l]$$

$$r_{xx}[0] = \sum_{n=-\infty}^{\infty} x^2[n] = \text{Energy of } x[n]$$

$r_{xx}[l] \rightarrow$ even fun.

$r_{xy}(l)$ will be $\sum_{n=-\infty}^{\infty} Y_n x_{n-l}$, okay. This n is over this range minus infinite plus infinite. So, this will be if you change the variables; you will get $r_{xy}(l)$, this is something very similar to which term, very similar to convolution, is it not? You try to find out the certain difference between convolution of a sequence up to sequences and cross correlation, there is a difference, all right.

So, here it is n , $n-l$ the other one is $l-n$. Now, auto-correlation of a sequence x , it $r_{xx}(l)$ is equal to $\sum_{n=-\infty}^{\infty} x_n x_{n-l}$, okay. $r_{xx}(0)$ will be $\sum_{n=-\infty}^{\infty} x_n^2$, it is also called the energy of the signal. Now this is an even function, $r_{xx}(l)$ is an even function, can I tell me why, this is say x_n this is also x_n all right; now if I give it a shift by l all right or if I give shift to this one, it is one at the same thing.

Means, if I shifted forward or backward, it is symmetrical, it is a same sequence, all right. So, it is an even function, is it not? For example, if you have hundred points; $x_0, x_1, x_2, x_3, \dots, x_{100}$ then if I take a lag of 1, so it will be x_0 into x_1, x_1 into x_2, x_2 into x_3 and so on, x_{99} to x_{100} .

If I take a lag on the other side; it will be x_1 into x_0, x_2 into x_1 , it is just x_1 and x_2 ; I mean the the two sequences are reversed all right, but they give you the same product, so it is an even function....Now, let us before we go to any other discussions; let us define a positive definite matrix, definite matrix or positive definite function.

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Positive Definite Matrix

$$\underline{\alpha} = \begin{bmatrix} \alpha_1 \\ \alpha_2 \\ \alpha_3 \end{bmatrix} \quad \underline{\beta} = \begin{bmatrix} \beta_1 \\ \beta_2 \\ \beta_3 \end{bmatrix}$$
$$\underline{\alpha}^T \underline{\beta} = \begin{bmatrix} \alpha_1 & \alpha_2 & \alpha_3 \end{bmatrix} \begin{bmatrix} \beta_1 \\ \beta_2 \\ \beta_3 \end{bmatrix} = \alpha_1 \beta_1 + \alpha_2 \beta_2 + \alpha_3 \beta_3$$
$$= \underline{\beta}^T \underline{\alpha}$$

$$= \underline{\beta}^T \underline{\alpha}$$
$$\begin{bmatrix} y_1 \\ y_2 \\ y_3 \end{bmatrix} = \underline{\alpha}^T \underline{x} = \alpha_1 x_1 + \alpha_2 x_2 + \alpha_3 x_3$$
$$\underline{\alpha} = \begin{bmatrix} \alpha_1 \\ \alpha_2 \\ \alpha_3 \end{bmatrix}$$

What is a positive definite function? Okay, let us take a vector alpha 1, alpha 2, alpha 3 and a vector vector beta as beta 1, beta 2 beta 3, okay. So, what is alpha transposed beta; it is an inner product which will be alpha 1, alpha 2, alpha 3 into beta 1, beta 2, beta 3, is alpha 1 beta 1 plus alpha 2 beta 2 plus alpha 3 beta 3, it is same as beta transposed alpha, is it not?

If I would have transposed these and then multiplied by this, I would have got the same product, is that so. Now, let us take y as α transposed x okay. A vector x is converted to a α transformed to a new vector y , so y_1, y_2, y_3 is a vector which is $\alpha_1 \times 1$ plus $\alpha_2 \times 2$ plus $\alpha_3 \times 3$, okay where α is α_1, α_2 , sorry sorry sorry I am sorry, this not the inner product.

(Refer Slide Time: 41:02)

The image shows a whiteboard with handwritten mathematical derivations. At the top, it states $y = Ax$. Below this, a vector y is expressed as a matrix A multiplied by a vector x with components x_1, x_2, x_3 . The next line shows the components of y as y_1, y_2, y_3 . Then, the inner product $y^T y$ is calculated as $(Ax)^T (Ax)$, which simplifies to $x^T (A^T A) x$. The matrix $A^T A$ is labeled as a positive definite matrix. The vector x is written as $(x_1 \ x_2 \ x_3)$ and the matrix $A^T A$ is written as (P) .

y is equal to, okay y is equal to Ax , all right, any matrix A . So, x_1, x_2, x_3 multiplied by A will give me vector Y , all right. The elements of x may be positive or negative, I don't know all right. What is α transposed α ? Any vector pre multiplied by its transpose will give the will give me α_1^2 plus α_2^2 plus α_3^2 which is always positive, okay.

So, what is y transposed Y ? It will be $Ax^T Ax$, all right which is $x^T A^T A x$ all right. This $A^T A$ is this entire thing is always positive, because $Y^T y$ is always positive. I have shown for any vector α , $\alpha^T \alpha$ is always positive; so any vector Y will be always giving you, $Y^T y$ as a positive quantity, okay. So, this is a positive quantity irrespective of the vector x .

So when you get such a function of X , that is a vector x_1, x_2, x_3 etcetera multiplied by a matrix P which is transpose A is a square matrix okay; post multiplied by the vector again, x_1, x_2, x_3 are the elements. If it is always positive then this is called a positive definite matrix, okay and this function is a positive definite function of X , whatever be the value of x the elements may be positive or negative; it is always positive, okay.

It is something like an energy function, say if move my hand in this direction there is lots of frictional energy all right, if I move it backward then also it will be positive, heat will be dissipated whether I move it this way or that way, all right. So, it is something like energy which is always positive so positive definite matrix and positive definite functions is defined.

Positive definite function is when it is always positive, if it is sometimes positive, sometimes zero, it is called positive semi-definite. Similarly, if some function is always negative just put a negative sign on this, this becomes a negative definite or negative semi-definite.

If you have a finite energy sequence, X_n and a combination of a x_n plus another sequence with a with a lag, what will be the energy content of this signal?

(Refer Slide Time: 44:34)

The image shows a whiteboard with handwritten mathematical equations. The equations are as follows:

$$E = \sum_{n=-\infty}^{\infty} \{ax[n] + y[n-1]\}^2$$

$$= a^2 \sum_{n=-\infty}^{\infty} x[n]^2 + 2a \sum_{n=-\infty}^{\infty} x[n] \cdot y[n-1] + \sum_{n=-\infty}^{\infty} y[n-1]^2$$

$$= a^2 \cdot r_{xx}(0) + 2a \cdot r_{xy}(1) + r_{yy}(0) \geq 0$$

$$\begin{aligned}
&= a^2 \left\{ \sum_{n=-\infty}^{\infty} x[n]^2 + 2a \sum_{n=-\infty}^{\infty} x[n]y[n-1] + \sum_{n=-\infty}^{\infty} y[n-1]^2 \right\} \\
&= a^2 r_{xx}(0) + 2a r_{xy}(1) + r_{yy}(0) \geq 0. \\
&(a \ 1) \begin{bmatrix} r_{xx}(0) & r_{xy}(1) \\ r_{xy}(1) & r_{yy}(0) \end{bmatrix} \begin{bmatrix} a \\ 1 \end{bmatrix} \geq 0. \\
&\quad \downarrow \\
&\quad +ve \text{ semi def. matrix.}
\end{aligned}$$

That means, you have a linear combination of two signals, all right you are measuring the energy content of that signal; if x_n is a finite energy, y_n is a finite energy then it will also be a finite energy sequence, all right. And let us see, what will be the energy of this sequence. Energy of this sequence is summation of this, which will be a squared x_n squared plus twice a sigma $x_n y_{n-1}$ plus sigma y_{n-1} square, equal to a squared this is nothing but $r_{xx}(0)$ plus $2a$, this is nothing but $r_{xy}(1)$ and this is $r_{yy}(0)$, okay because there all varying over minus infinitive to plus infinitive.

We are assuming the sequence to be, extended in both the directions all right. And that must be always greater than equal to 0, because this is finite energy, this is finite energy. So, we can write this as a 1×1 $r_{xx}(0)$, $r_{xy}(1)$ 1×1 , $r_{yy}(0)$, this matrix into a 1 that is greater than equal to 0; so this is a positive semi-definite form matrix, positive semi-definite matrix.

If it is positive semi-definite, basically for a two by two matrix it is very simple, the determinant must be always greater than equal to 0, determinant must be always greater than equal to 0.

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The whiteboard contains the following handwritten mathematical expressions:

$$r_{xx}(0) \cdot r_{yy}(0) - r_{xy}^2(l) \geq 0$$
$$r_{xy}(l) \leq \sqrt{r_{xx}(0) \cdot r_{yy}(0)}$$
$$\leq \sqrt{E_x \cdot E_y}$$
$$r_{xx}(l) \leq r_{xx}(0)$$
$$\leq E_x$$

A hand holding a pen is visible at the bottom left of the whiteboard.

That means $r_{xx}(0) \cdot r_{yy}(0) - r_{xy}^2(l)$ should be greater than equal to 0 or r_{xy} should be less than equal to $\sqrt{r_{xx}(0) \cdot r_{yy}(0)}$; that is under root of E_x energy of signal x , energy of signal y , okay. So by the same logic when we are considering the same signal x , therefore r_{xx} at any lag should be less than r_{xx} , this is l ; into r_{xx} under root will be $r_{xx}(0)$ or that is equal to E_x .

So, if you take a signal; if you take its autocorrelation with a lag say, 1 or 2 or 3 that will be always less than or equal to at the most; the value of the autocorrelation when you are taking at n equal to 0, when you are multiplying a quantity with itself and then take the product sum, that is giving the maximum value.

So, we will stop discussion on this, at this point. Next time, we would like to take up discrete Fourier transform, in most of the books they do not distinguish between discrete Fourier transform, discrete Fourier series and discrete time Fourier transform. So, we would like to bring out the certain difference between these terms first and then we will go over to the competition of discrete Fourier transform.

Discrete Fourier transform is the most widely used transform in the signal processing and its numerical; the algorithmic version is fast Fourier transform which is a basically competition of DFT, very fast algorithm for competition purpose. So, we will take up DFT and FFT in the next class. Thank you very much.