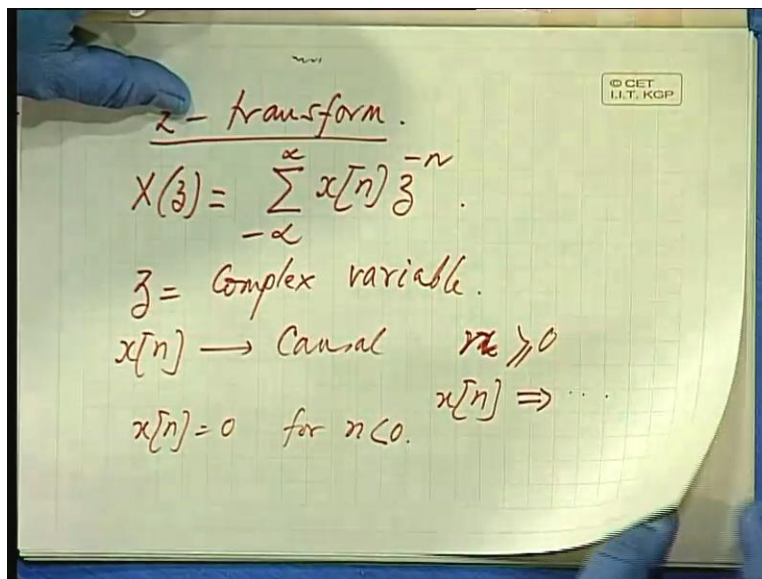


Digital Signal Processing
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Lecture - 5
Z-Transform

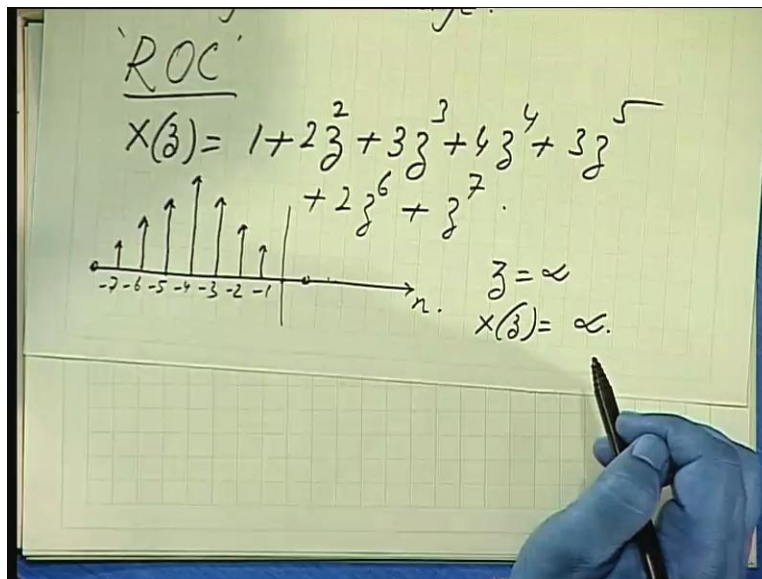
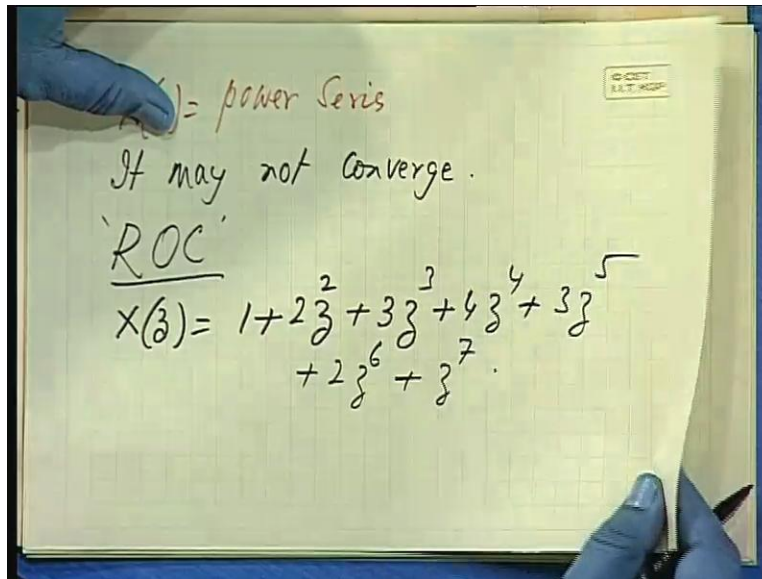
We just give you the definitions of Z transform, we shall continue with that.

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We will see some other important properties of Z transform. We defined, $X(z)$ as $\sum_{n=-\infty}^{\infty} x[n] z^{-n}$ to the power minus n , n varying from minus infinity to plus infinity, z is a complex variable. Now, if $x[n]$ is causal, if $x[n]$ is causal; causal means it exists only for n greater than or equal to 0, all right for n greater than or equal to 0, $x[n]$ exists, okay. This is non-zero, may be zero but $x[n]$ is 0 for all negative n , this a causal sequence.

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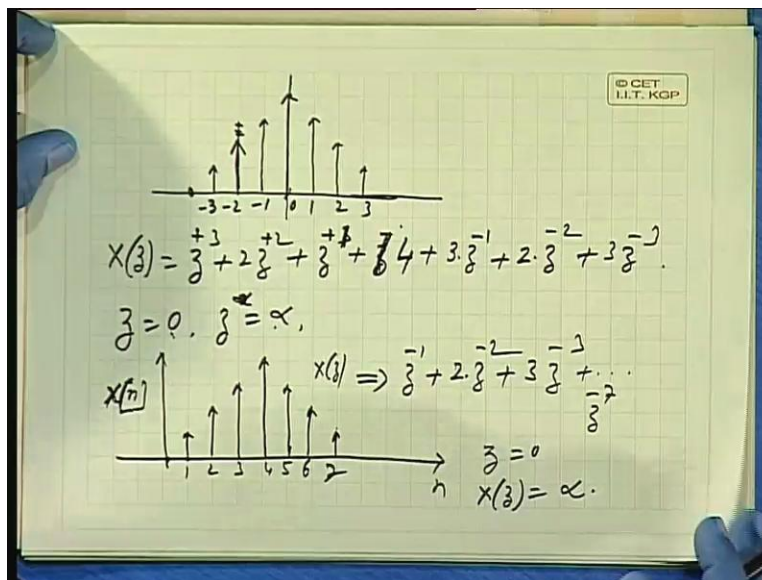
Now, $x(z)$ is a power series. Let you can see, $x(z)$ it may not converge at all, it may or may not converge. So, let us investigate under what situation it may not converge, the region in which it converges we call region of convergence or in short R O C, okay. So will put the boundaries where it will be converging or it will not be converging.

Let us take the following series, $x(z)$ equal to 1 plus 2 z^2 plus 3 z^3 plus 4 z^4 plus 3 z^5 plus 2 z^6 plus z^7 . What is this sequence; in terms of z to the power minus one, if we try to

see. It is a sequence which exists in the negative domain of time okay, zero, one, two, three, four, three, two, one and then zero. So minus 1, minus 2, minus 3, minus 4, minus 5, minus 6, minus 7 this is n. Now, for such a sequence where for negative values of n only x n exists and this is all zero. The value of x z will be this or does it exist for all values of z, at z is equal to infinity, x z becomes infinity; we would like to see when the function becomes infinity, that means it is not converging.

When z is very very large as it is tending to infinity, the function exist also tends to infinity. So, the function exists except at z is equal to infinity. Let us take, the similar sequence but with a shift.

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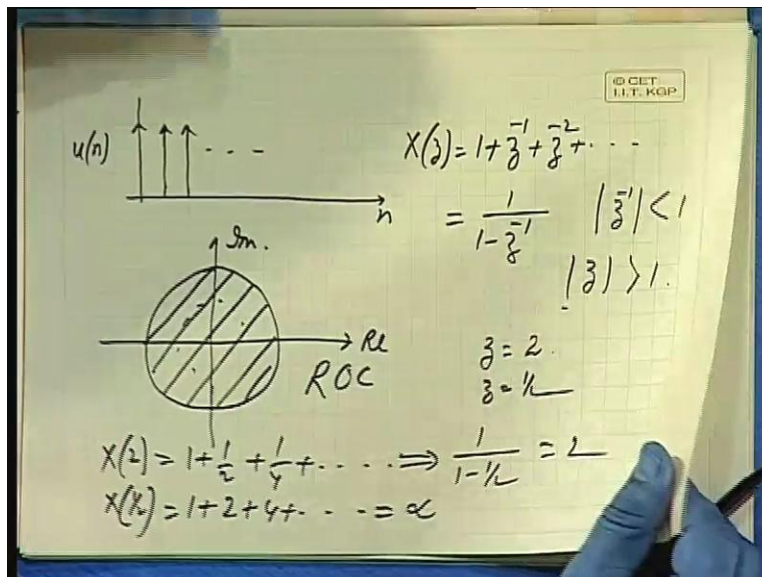
Suppose, we have the same sequence with a shift four, three, two, one, three, sorry sorry this is two, one, zero; so minus 1, minus 2, minus 3 and then this is 1, 2, 3, 0. What will be the value of x z? z to the power minus 3 starts from here plus 2 into z to the power minus 2 plus z to the power minus 3 plus z to the power minus, is it z to the power minus 4? It should be minus 1, then 4 into z to the power 0 plus 3 into z to the power it is sorry, plus 2 plus 1, is it all right, now 2 into z to the power minus 2 plus 3 into z to the power minus 3, okay.

Now, what are the values for which $x(z)$ becomes infinity? z is equal to 0 it becomes infinity, z is equal to infinity also it becomes infinity. So, its region of convergence will exclude these two points; at these two points the function becomes infinity, is that all right? If we shifted further by four more steps then we will have a function like this, zero, one, two, three, four, three, two, one okay.

What will be the sequence, corresponding sequence z to the power minus 1? It is $x(z)$, z to the power minus 1 plus z to the power minus 2 into 2 plus 3 into z to the power minus 3 and so on, at the end z to the power minus 7, 1, 2, 3, 4, 5, 6, 7 all right. Now, when this function becomes infinity that you have to identify, that you can see from here; when z is equal to 0, anybody? z is equal to zero, okay when z is equal to zero $x(z)$ is infinity.

So, a sequence that we have considered in the three cases is having the same nature of variation, but it is shifted all right from left to right and the region of convergence also changes; earlier it was not converging for z is equal to minus z is equal to infinity then z is equal to 0 and infinity and now z is equal to 0. So, region of convergence will be depending on the nature of the polynomial, nature of the function $x(z)$ that you get.

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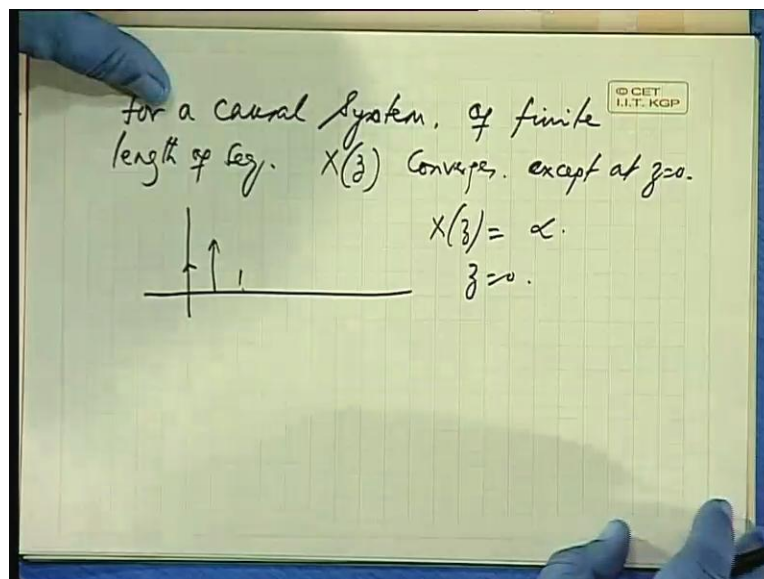


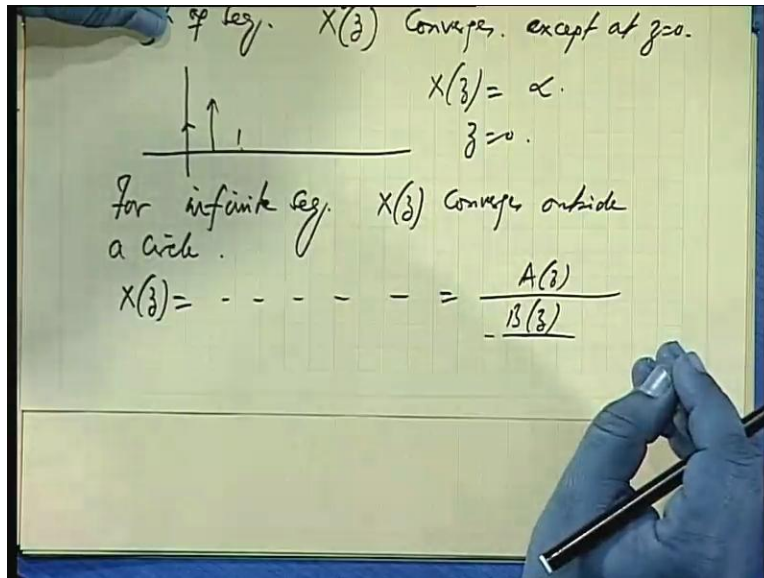
Let us consider this sequence, well known step function which is one, one, one and continues. So, what will be corresponding Z transform? $x(z)$, $1 + z^{-1} + z^{-2} + \dots$ and so on; it will be $1/(1 - z^{-1})$, it is g.p series provided, $|z^{-1}| < 1$ or $|z| > 1$, okay.

So, the region of convergence, z is a complex quantity. So, in the complex z plane it will define a unit circle, okay. So, this is a forbidden range and this is R O C, the space outside is a region of convergence; region of convergence is the space outside, this is real and imaginary part, all right. Now, for this sequence you can see for yourself, if I put z is equal to 2 then it is $1 + \frac{1}{2} + \frac{1}{4} + \dots$ and so on.

And this a g.p series with reduced magnitude of these terms, so this will be tending to $1/(1 - \frac{1}{2}) = 2$. If I put z is equal to half then it becomes $1 + 2 + 4 + \dots$ and so on, tends to infinity, okay. So, any point inside if I choose, I have just taken half you can say choose any good, any value, any complex value it will be diverging.

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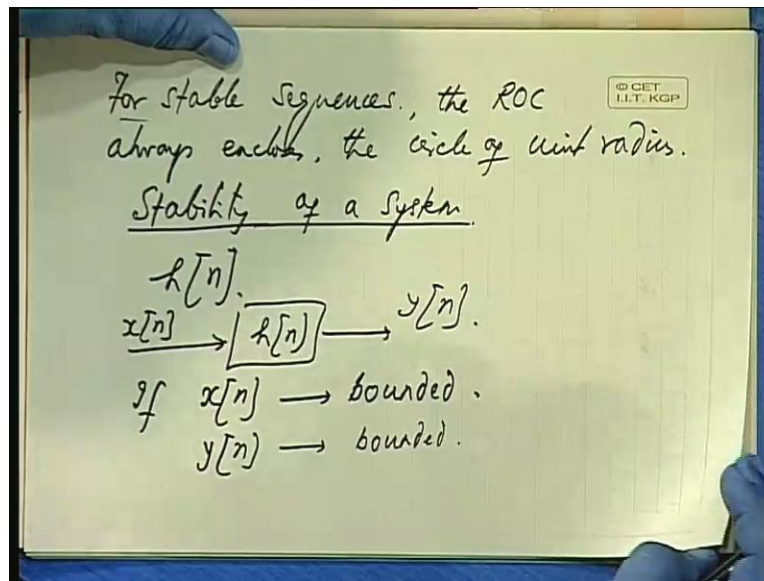


So for a causal system, for a causal system of finite sequence length of finite length of sequence; $x(z)$ converges because you have got finite number of terms, okay except at zero, except at zero your causal sequence means, the terms will be $x(0) + x(1)z^{-1} + x(2)z^{-2} + \dots$ and so on, all right.

It will be causal sequence means; it will have some value here, some value here and so on. If I put z is equal to zero, z to the power minus one will tend to infinity. So all the terms will be giving you finally, $x(z)$ will tend to infinity, at z is equal to zero. For infinite sequence, that is sequence of infinite duration $x(z)$ converges outside a circle. The circle is bounded by the poles okay; poles means if it is an infinite sequence, all right if you can resolve it in terms of some $A(z)$ by $B(z)$, then the roots of $B(z)$ will be called poles.

So, at those points at those values the function will try to blow up. This becomes zero means, the function will be having an infinite values. So at poles, the value of the function will be infinity all right so that will be excluded, the region of convergence will have to exclude those poles, all right. You must have studied in complex algebra, analytic functions; so wherever there are poles, there will be the function will try to go to infinity, so for an analytic function it excludes those poles, is it not?

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For stable sequences, we have not yet defined stability, what stability? So, next we will take up stability, for stable sequences the R O C, the region of convergence always encloses the circle of unit radius; this point we will discuss after we define what stability is, okay. Stability of a sequence of a system; when we define the impulse response of a system, impulse response of a system h_n , this dictates the barrier of the system all right.

Now, what are the conditions on h_n , what are the conditions on h_n for stability? What do you mean by a stable sequence or a stable system? If this is driven by an input which is bounded then output must be bounded; that means if x_n is bounded for a bounded input sequence, if x_n is bounded then y_n also should be bounded for a stable sequence h_n , all right. If that is violated, then the system is unstable all right.

What is it actually telling us, it says that; if I give a input sequence whose magnitude does not grow beyond a certain value, whose maximum value is limited then the output will also have a maximum value which will be limited, which will be less than infinity, output cannot tend to infinity if it does, then it is an obstacle system, all right.

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$$\text{Let } |x[n]| \leq M_x.$$
$$y[n] = \sum x[n-k] \cdot h[k].$$
$$|y[n]| \leq \sum |x[n-k]| |h[k]|$$

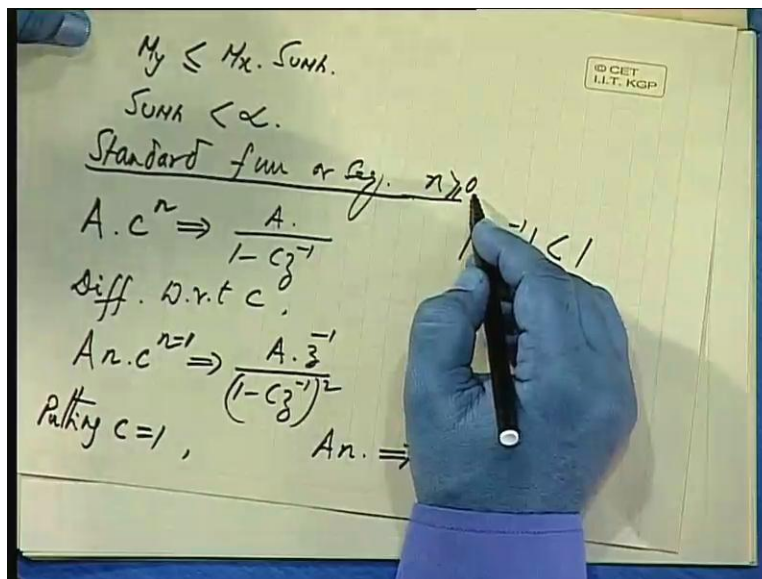
$$|y[n]| \leq \sum |x[n-k]| |h[k]|$$
$$\leq \sum M_x |h[k]|$$
$$\leq M_x \sum |h[k]|$$
$$\leq M_x \cdot \text{Sum } h$$
$$|y[n]| \rightarrow \text{is bounded, i.e. } |y[n]| \leq M_y.$$

So, let us see the proof of this. Let the magnitude of x at the maximum value, be restricted to M_x , so x can be at the most equal to less than M_x . Now $y[n]$ is $\sum x[n-k] \cdot h[k]$, is it not? So, $y[n]$ magnitude will be less than or equal to magnitude of x at $n-k$ into magnitude of $h[k]$, do you all agree? It has to be less than the product of the two magnitudes and then submitted.

So that means, it must be less than if I take the maximum value of this which is M_x into $\sum_{n=0}^{\infty} K$; it is definitely less than this product, I taken the maximum value of this summation. I can put the summation, since this is a constant. Summation sign here, so y_n if this summation is denoted as $\sum_{n=0}^{\infty} y_n$, I call it $\sum_{n=0}^{\infty} y_n$; it must be less than this product all right.

Therefore, $\sum_{n=0}^{\infty} y_n$ must be finite for magnitude of y_n to be finite, to be less than a quantity finite quantity. This to be bounded, that is if this is to be less than or equal to $\sum_{n=0}^{\infty} M_y$ then M_y must be less than M_x into $\sum_{n=0}^{\infty} h$ therefore $\sum_{n=0}^{\infty} h$ must be less than infinity, it must be a bounded quantity, it must be a finite quantity, okay.

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So this is the condition for stability of a system. You add up the magnitude of all the values of h , it must be converging to some finite value. Now, let us take some standard functions and evaluate the Z transform, standard functions for sequences. We have seen A into C to the power n , if you take the summation; it will be A by 1 minus $C z$ inverse, provided this magnitude is less than 1 , is it not? This we saw last time, C to the power n if you take the Z transform the series that you get, this is a common ratio, so that magnitude must be less than 1 .

If I differentiate with respect to the common quantity C, as you have done in case of Laplace transform, if you differentiate with respect to C differentiate with respect to C then I get, A n C to the power n, mind you; I am considering sequence for which exist for n greater than equal to 0, that is we are taking only causal sequences, okay. And I list out the corresponding region of convergence.

So, A n into C n will give me n minus 1, thank you. If I differentiate, I will get A by 1 minus C z inverse whole squared into z to the power minus 1 okay, minus 1 and minus 1 make final 2 minus 1 will give me plus, what will be the region of convergence? The same C z inverse must be less than 1. If I put C is equal to 1, what do you get; A into n will give me A into z inverse by 1 minus z inverse whole square, what is A into n, if you remember what is A into n?

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$$A_n \rightarrow \frac{A z^{-1}}{(1-z^{-1})^2} \quad \text{ROC } |z| > 1$$

$$A n^2 u[n] \rightarrow \frac{A z (z+1)}{(z-1)^3} \quad |z| > 1$$

$$C = e^{-\alpha + j\beta}$$

$$C^n \rightarrow e^{-\alpha n} e^{j\beta n}$$

A into n is basically ram function, moment I have defined n greater than equal to 0, so all these functions should have their implied u n as a multiplier, all right so, it is basically a ram function, it is implied though it is not mentioned because I am taking only n greater than equal to 0. So, A z inverse by 1 minus z inverse squared, now what are the regions of, what is the region of convergence, I leave it to you to find out the condition where it becomes infinity, okay.

Similarly, $A n^2 u[n]$ try to check whether you get this; I can always multiply by z to the power minus 2, z to the power plus 2. So, this will become $A z$ divided by z minus 1 whole square, okay. In many books we will find, they write in terms of z all right, Z transform and this becomes z minus 1 whole cube; derivative it yourself and again z greater than 1, z inverse less than 1 means z greater than 1, okay.

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The image shows a hand holding a pen pointing to a whiteboard with the following handwritten text:

$$A n^2 u[n] \rightarrow \frac{A z(z+1)}{(z-1)^3} \quad |z| > 1$$

$$C = e^{-\alpha + j\beta}$$

$$C^n \rightarrow e^{-\alpha n} \cdot e^{j\beta n}$$

$$A \cdot e^{-\alpha n} \cdot e^{j\beta n} \Rightarrow \frac{1}{1 - e^{-\alpha} \cdot e^{j\beta} \cdot z^{-1}}$$

If C is, once again we come back to the first relation involving C . Suppose, C is equal to e to the power minus alpha plus j beta then C to the power n means e to the power minus alpha n e to the power j beta n , do you all agree? Now, what will be the Z transform of this sequence? So, A into e to the power minus alpha n into e to the power j beta n , you just put in place of C this quantity.

So, it will be 1 by 1 minus e to the power minus alpha, e to the power j beta into z to the power minus 1, okay. I just substituted for C e to the power minus alpha plus j beta, if you segregate the real and imaginary part; what do you get? Let us see.

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Separating the real and Im parts

$$\mathcal{Z} e^{-\alpha n} \cos \beta n \Rightarrow \frac{z^2 - z \cos \beta}{z^2 - 2z e^{-\alpha} \cos \beta + e^{-2\alpha}}$$

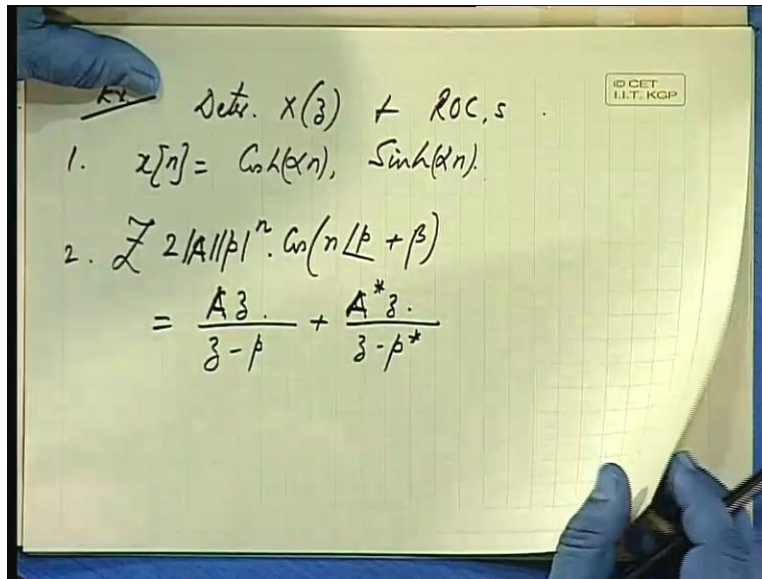
$$\Rightarrow \frac{1 - z^{-1} \cos \beta}{1 - 2z^{-1} e^{-\alpha} \cos \beta + e^{-2\alpha} z^{-2}}$$

$$\mathcal{Z} e^{-\alpha n} \sin \beta n = \frac{z^{-1} \sin \beta}{1 - 2z^{-1} \cos \beta e^{-\alpha} + e^{-2\alpha} z^{-2}}$$

Separating the as you are done in the case of Laplace transform, separating the real and imaginary parts and equating them on both sides; we get Z transform of $e^{-\alpha n} \cos \beta n$, as $z^2 - z \cos \beta$ by $z^2 - 2z e^{-\alpha} \cos \beta + e^{-2\alpha}$, okay. Some books write $e^{-\alpha}$ to the power minus alpha is just r naught, okay.

So you can do that, because $e^{-\alpha}$ appears as it is. One may write also this as, $1 - z^{-1} \cos \beta$, this is also written in many books in this form; $e^{-\alpha}$ to the power minus alpha $\cos \beta$ plus $e^{-2\alpha}$ to the power minus twice alpha z^{-2} . Similarly, Z transform of $e^{-\alpha n} \sin \beta n$, will be $z^{-1} \sin \beta$ divided by the same denominator, okay. We shall be using it later on in partial fractions in certain situations that we will see very soon, I will keep it aside.

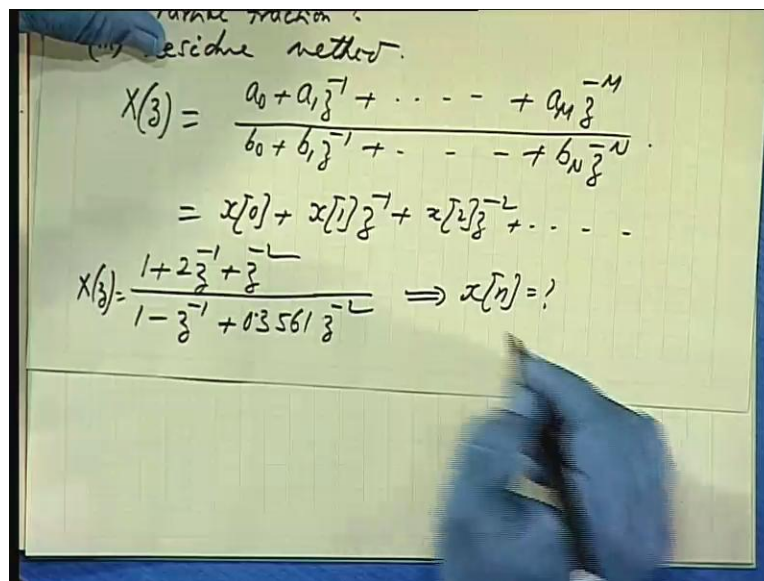
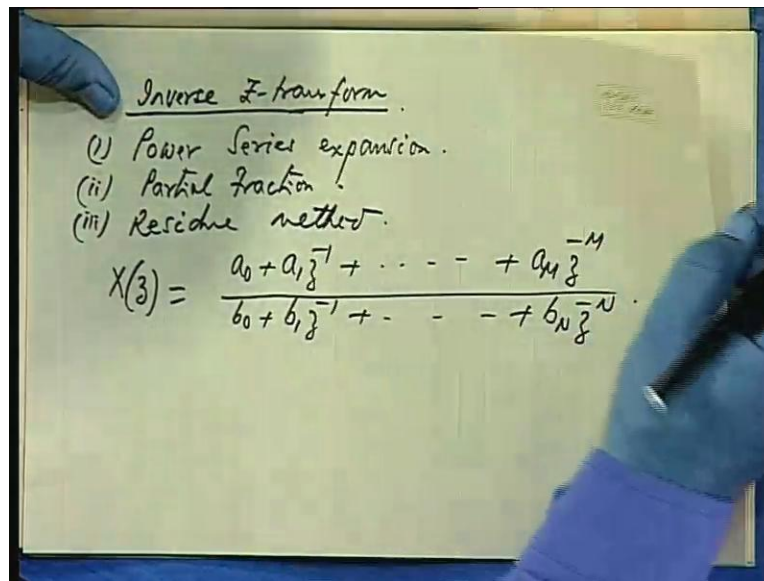
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I would like you to try, try to find out the R O C's determined $X z$ and R O C' s for the two functions; $x n$ equal to $\cos n$ and $\sin n$, okay, αn , $\cos \alpha n$, $\sin \alpha n$. And second, prove that Z transform you try this yourself e to the power n , I am writing this C is basically A, okay; A $\cos n$ angle of p plus β is $C z$ by z minus p plus C star z by z minus p star, A sorry I have change it to A, okay to avoid confusion with a earlier C, okay.

So, A is a complex quantity all right. Now let us see, how to in you can evaluate the Z transform of many such functions with the same simple principle of addition.

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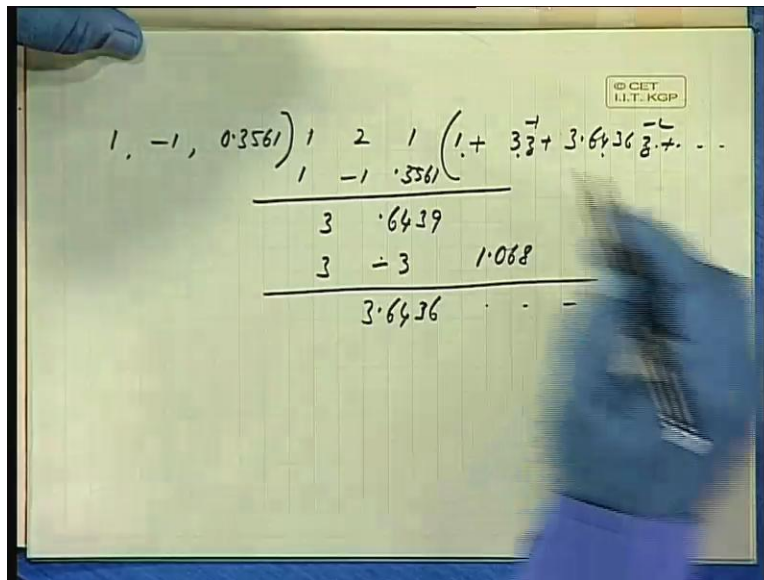


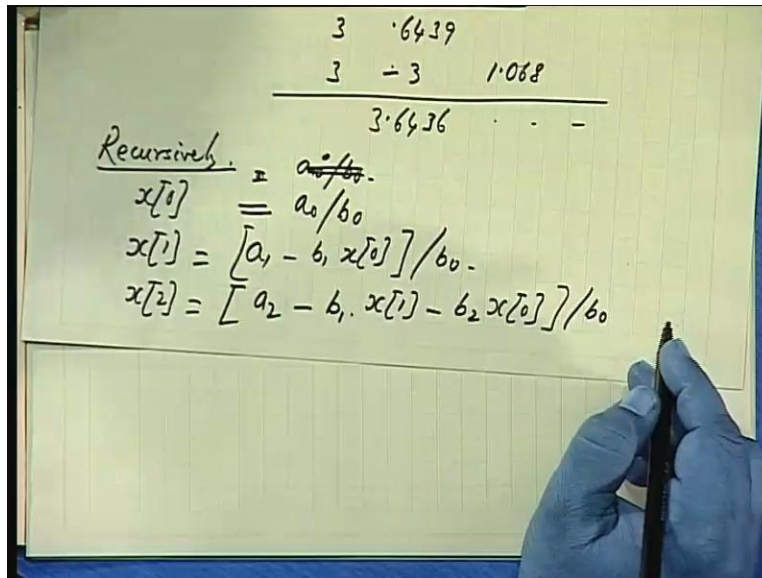
Now, let us see how you get the inverse of these functions? Given a function $X(z)$, how do you find out the sequence $x[n]$ that will be our next task, inverse Z transform okay. So, there are three standard methods very simple, some of you must have studied in control systems also. Power series expansion, second one is partial fraction and third one is residue method, residue calculation, okay.

So the first one, $X(z)$ suppose; we give you $X(z)$ as $a_0 + a_1 z^{-1}$ like this a polynomial $a_m z^{-m}$ to the power minus m divided by $b_0 + b_1 z^{-1} + \dots + b_N z^{-N}$. By long division, you generate this as $x_0 + x_1 z^{-1} + x_2 z^{-2} + \dots$ and so on; all right you can divide it and then get these.

For example, if I have $1 + 2z^{-1} + z^{-2}$ divided by $1 - z^{-1}$; say I have taken a simple quantity, z^{-2} , $0.3561 z^{-2}$, what will be the sequence if $X(z)$ is given by this ratio of two polynomials, what will be x_n ? It will be an infinity sequence.

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So, you divide it, take the denominator of coefficients 1, minus 1m, 0.3561; I am not writing z to the power zero, z to the power minus one, z to the power minus two, just put little gap in between the coefficients; similarly one two one, 1, 2, 1, okay. So, the division gives me 1, 1, minus 1, 0.3561. So, after subtraction I get 3.6439 then next division gives me, 3, 3 then minus 3, 1.0681; three into this much so that gives me, 3.6436 and so on, okay.

So, next value will be 3.6436. Now this is x zero, this is x one, this is x two and so on because this is coming as a constant. What you are getting is 1 plus 3 z inverse plus into z to the power minus 2 and so on, so this will be endless, so will be the sequence. So, the coefficients of z inverse, z to the power minus 2 and so on, they will represent x zero, x one, x two, okay it is as simple as that.

Recursively, we can write this as the same result we can write x 0, x zero is basically a 0 by b 0, okay sorry it should be here, okay. x 1 will be a 1, sorry b 1 into x 0 whole thing, divided by b 0. x 2, I write one or two more terms and then a 2 minus b 1 into x 1 minus b 2 into x 0, divided by b 0.

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$$x[n] = \left[a_n - \sum_{i=1}^n x[n-i] \cdot b_i \right] / b_0$$

ii Partial fraction.

$$X(z) = \frac{a_0 + a_1 z^{-1} + \dots + a_M z^{-M}}{b_0 + \dots + b_N z^{-N}}$$

$\frac{M}{N}$ $M=N$

$$= C_0 + \frac{C_1}{1-p_1 z^{-1}} + \frac{C_2}{1-p_2 z^{-1}} + \dots + \frac{C_N}{1-p_N z^{-1}}$$

$$= C_0 + \sum_{k=1}^N \frac{C_k}{1-p_k z^{-1}}$$

So generalizing this, it will be $X[n] = [a_n - \sum_{i=1}^n x[n-i] \cdot b_i] / b_0$. So, you can use, I would request you to write as short program. Take any polynomial a and b , a and b and then get the values of $x[n]$ from this recursive relation, okay may be up to n is equal to ten or twelve. Write a short program and then see for yourself, give the coefficients and then find out the values, okay.

The next method is partial fraction, $X(z) = a_0 + a_1 z^{-1} + \dots + a_M z^{-M}$ to the power minus m divided by $b_0 + \dots + b_N z^{-N}$. If M is equal to N then we can write; if M is equal to N then we can write this as, some $C_0 + C_1 / (1 - p_1 z^{-1}) + C_2 / (1 - p_2 z^{-1}) + \dots + C_N / (1 - p_N z^{-1})$ and so on.

$C_N / (1 - p_N z^{-1})$ to the power minus 1 equal to, I can write this as a very general term $C_K / (1 - p_K z^{-1})$, z to the power minus 1 okay; K varying from 1 to N . Now, from here see from the polynomial in the denominator, we have evaluated the routes so the routes have to be evaluated here and then you can write this one, for each one of them you know the solution.

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$$x[n] = \sum [C_k] z^k u[n].$$

$$C_k = \frac{X(z) \cdot (z - p_k)}{z} \Big|_{z=p_k}.$$

$X(z) \rightarrow$ multiple roots $m..$

$$X(z) = \sum_{i=1}^m \frac{D_i}{(z - p_k)^i}$$

$$\frac{1}{(z - 0.5)^3} = \frac{D_1}{z - 0.5} + \frac{D_2}{(z - 0.5)^2} + \frac{D_3}{(z - 0.5)^3}.$$

So, $X[n]$ will consist of this sequence C_k to the power n summation C_k , okay. C_k , C_1 , C_2 etcetera all to the power n , $u[n]$ all right. And all we evaluate C_k , it will be $X(z)$ by z into z minus p_k evaluated at z is equal to p_k , as you have calculated residuals in case of Laplace transform. If, $X(z)$ has multiple roots, multiple roots say; m number of multiple roots m number of multiple roots then that is say, $x(z)$ contains a term like this, D_i by z minus p_k to the power i , i is varying to m varying from 1 to m , okay.

There is m number of routes are there, at the same place p_k then the corresponding residues; so if it is 1 by z minus say 0.5 to the power 3, then I will have a term say D_1 by z minus 0.5 plus D_2 by z minus, sorry z minus 0.5 squared plus D_3 by z minus 0.5 cube. So, these D_i 's, the general term D_i can be evaluated m minus i factorial, sorry excuse me m minus i $d z$, m minus i , z minus p_k , exactly identical method as you have done in case of Laplace transform, okay.

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The image shows a handwritten derivation on a notepad. At the top, the residue formula is written as $a_i = \frac{1}{(m-i)!} \frac{d^{m-i}}{dz^{m-i}} \left[(z-p_k)^m \cdot \frac{x(z)}{z} \right]_{z=p_k}$. Below this, an example is given: $x(z) = \frac{z^{-1}}{1 - 0.25z^{-1} - 0.375z^{-2}}$. This is then simplified to $x(z) = \frac{z}{(1 - 0.75z)(1 + 0.5z)}$. Finally, it is expressed as a sum of partial fractions: $x(z) = \frac{A_1}{1 - 0.75z} + \frac{A_2}{1 + 0.5z}$.

Let us take one or two small examples. Suppose, $X(z)$ is given whereas z inverse by $1 - 0.25z^{-1} - 0.375z^{-2}$, I can evaluate the residues; this is factorable you can this is quadratic, you can find out the factors, okay.

So, $1 - 0.75z^{-1}$; okay let me write in the factor factored form first, $1 + 0.5z^{-1}$ to the power minus 1, okay, the factors are actually like this. So, this I write as A_1 by $1 - 0.75z^{-1}$ plus A_2 by $1 + 0.5z^{-1}$ to the power minus 2, okay. So A_1 and A_2 , one may try very simple combinations of 0.75 and 0.5.

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$$A_1 = 0.8 \quad A_2 = -0.8$$
$$X(z) = 0.8 \left[\frac{1}{1 - 0.75z^{-1}} - \frac{1}{1 + 0.5z^{-1}} \right]$$
$$x[n] = 0.8 \left[(0.75)^n - (-0.5)^n \right] u[n]$$

You can see A_1 is equal to 0.8; A_2 is minus 0.8. So, $X(z)$ becomes 0.8 into $\frac{1}{1 - 0.75z^{-1}}$ minus $\frac{1}{1 + 0.5z^{-1}}$. So, what will be $x[n]$, very simple 0.8 corresponding to this it is 0.75 to the power n minus, minus 0.5 to the power n , whole thing multiplied by $u[n]$.

So, these are closed. This is minus 0.5 this was 1 plus 0.5 thank you, thank you 1 plus 0.5 z^{-1} . Now, if $B(z)$ has the quadratic with complex routes what will you do; if you remember in your early classes, in Laplace transform we studied the complex routes. If the denominator contains complex routes when you try to resolve in say, s plus α Whole Square plus β square form, remember?

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$$(s+\alpha)^2 + \beta^2 \quad \frac{A(s)}{B(s)}$$

$$A \quad \frac{A(s)}{B(s)} = \frac{K_1(s+\alpha)}{(s+\alpha)^2 + \beta^2} + \frac{K_2 \beta}{(s+\alpha)^2 + \beta^2}$$

$$\frac{A(s)}{B(s)} = \frac{K_1(s+\alpha)}{(s+\alpha)^2 + \beta^2} + \frac{K_2 \beta}{(s+\alpha)^2 + \beta^2}$$

$$X(z) = \frac{A(z)}{B(z)}$$

$$= \frac{A(z)}{1 - 2e^{-\alpha} \cos(\beta)z^{-1} + e^{-2\alpha} z^{-2}}$$

And we tried to say, if you are given some function $A(s)$ by $B(s)$, all right and $B(s)$ contains a factor like this then you try to dissolve it into; some A into some K_1 into $s + \alpha$ by $s + \alpha$ whole squared plus β square plus K_2 times β by $s + \alpha$ whole plus β square, all right. Corresponding to this, you get a cosine function as inverse and for this you get a sine function, exactly similar procedure we shall follow here.

So, $X(z)$ if it is given as $A(z)$ by $B(z)$ and $B(z)$ is having a quadratic then we can write $B(z)$ as $1 - 2e^{-\alpha} \cos \beta z^{-1} + e^{-2\alpha} z^{-2}$ into some e to the power minus α , $\cos \beta z^{-1}$ plus e to the power minus twice αz to the power minus 2, okay.

So, the quadratic that will be given to you, quadratic that will be given to you will tell you the last term; z to the power minus 2 whatever is a coefficient that will be corresponding to e to the power minus 2 α s. Take that square root e to the power of minus 2 α then the middle term should be equally 2 into e to the power minus α into $\cos \beta$, so you know $\cos \beta$, okay you can calculate $\sin \beta$ also.

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The slide shows the following handwritten equations:

$$= K_1 \cdot \frac{1 - z^{-1} \cos \beta \cdot e^{-\alpha}}{1 - 2e^{-\alpha} \cos \beta z^{-1} + e^{-2\alpha} z^{-2}}$$

$$+ K_2 \cdot \frac{z^{-1} \sin \beta e^{-\alpha}}{1 - 2e^{-\alpha} \cos \beta z^{-1} + e^{-2\alpha} z^{-2}}$$

$$x[n] = K_1 e^{-\alpha n} \cos n\beta + K_2 e^{-\alpha n} \sin n\beta$$

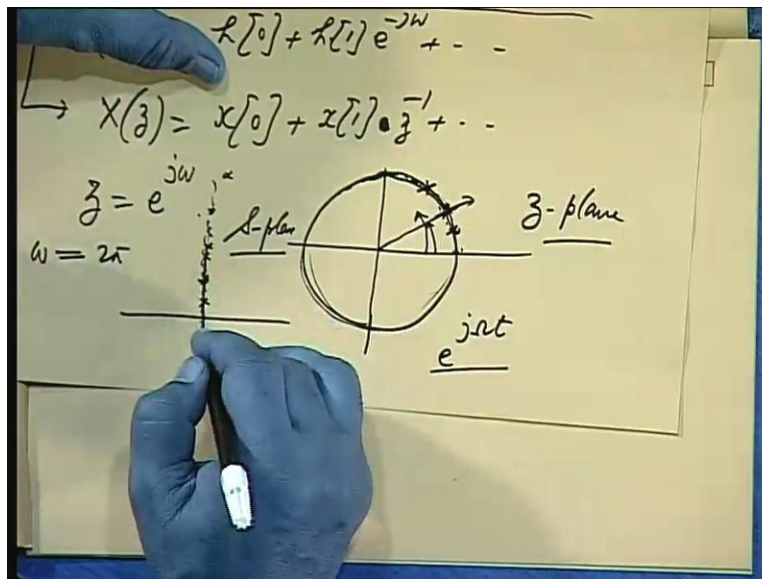
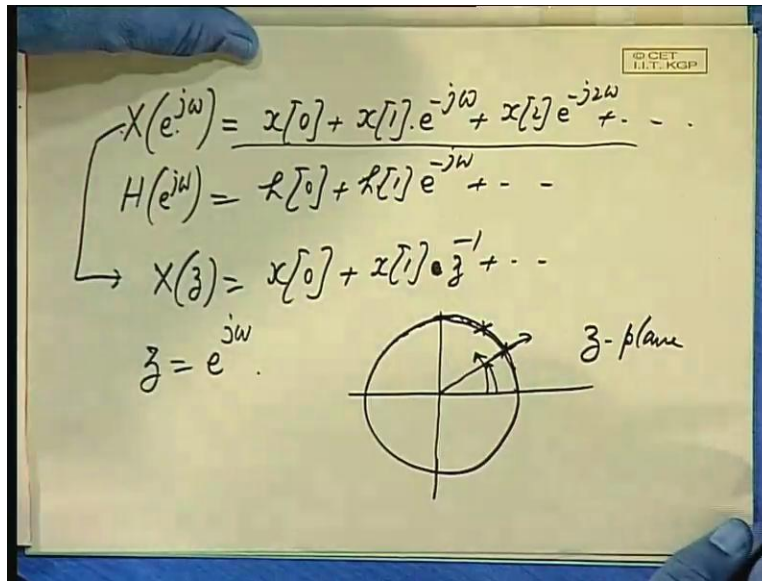
$$= e^{-\alpha n} [K_1 \cos n\beta + K_2 \sin n\beta]$$

So, this finally you reduce into the same form, some K_1 times $1 - z^{-1} \cos \beta$ e to the power minus α divided by listed denominator plus K_2 times, $z^{-1} \sin \beta$ e to the power minus α by $1 - 2e^{-\alpha} \cos \beta z^{-1} + e^{-2\alpha} z^{-2}$ plus e to the power minus 2 αz to the power minus 2.

So, corresponding $x[n]$ will be after evaluating K_1 and K_2 , it is just a matching that you have to do for the numerator function. So, you get K_1 times, what is the inverse of this? e to the power minus α , $\cos \beta$, okay. So $\cos n\beta$ and this is minus αn plus K_2 times e to the

power minus alpha n, sin n beta, okay. So that, gives me K 1 into e to the power minus alpha, sorry e to the power minus alpha n I can take out; K 1 into cos n beta plus K 2 into sin n beta. It is better to write in separate forms, sin n, beta n, cosine n beta. Somebody may write cosine n beta plus theta, whatever you want okay.

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Now, sometimes we try to link up the discrete time Fourier transforms; that is if you remember earlier we are talking about, frequency response of the system or frequency domain

representation of a signal, that was x_0 plus $x_1 e^{-j\omega}$ plus $x_2 e^{-j2\omega}$ and so on, is it not?

If you remember, we are writing $H e^{-j\omega}$ frequency response of the system whose time domain response, impulse response was H_n . So, this was h_0 plus $h_1 e^{-j\omega}$ and so on, okay. You see, in $X(z)$ the expression is x_0 plus $x_1 z^{-1}$ and so on; so from the z transform you can directly get the frequency domain representation of the sequence, just put z is equal to $e^{-j\omega}$, you get the same function.

So, in the z domain what is $e^{-j\omega}$, it is a unit circle as ω increases sorry, ω increases along this; $e^{-j\omega}$, ω is equal to say two radians, means I will go here by two radians.

So these are the points on the unit circle in the z plane; this is a z plane, z plane is a complex plane okay. That means some specific values of z which will be defined on this unit circle, if I take then $X(z)$ represents nothing but the discrete time Fourier transform of the sequence, we collect DTFT, this is known as DTFT; discrete time Fourier transform of a sequence X_n , is it all right?

So just take the unit circle, take the points on this so that presents $x e^{-j\omega}$. Now, when I come back that is when is complete two π , $\omega = 2\pi$, again I come back to the same point, same valuation, all right. So, I keep on making circles as ω tends to infinity from zero; if I keep on increasing the value of ω on the in the z plane on this circle, on this perimeter, I am just describing the same points, all right which is one another same thing. This is the point that you shall take up later on, when I talk about stability of systems and well in the filter design.

As in the case of continuous domain, you have varied along the imaginary axis of s plane; along the imaginary axis of s plane you have varied the frequency from minus infinity to plus infinity

or from 0 to infinity, if you take 0 to infinity. Basically, points on this represent different sinusoids or exponential periodic functions of this nature, is it not?

You consider the frequencies of the imaginary axis. We will try to establish a relation between s plane and z plane parameters, that is if I shift a point in the continuous domain along this, what does it mean in the z plane? Now that mapping of a particular zone in the x plane onto z plane that means from the continuous domain, how to translate it into discrete domain, is that all right. Thank you very much, we will continue with this in the next class.