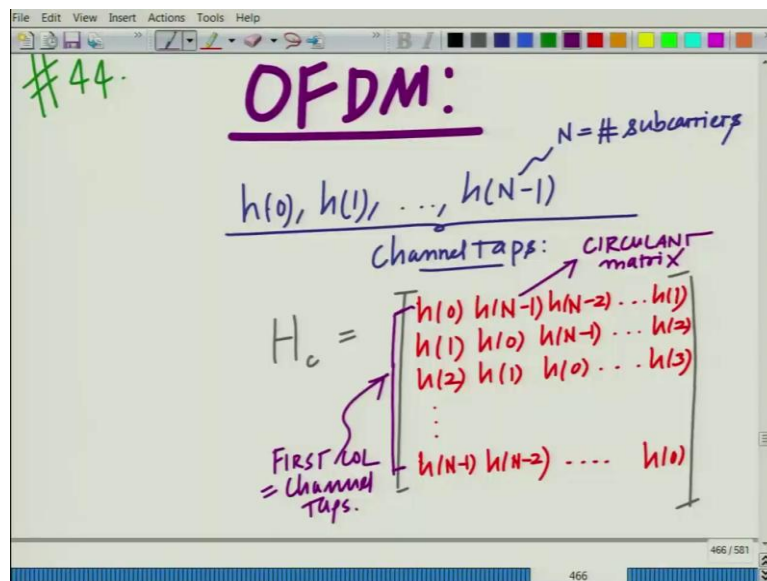


Applied Linear Algebra for Signal Processing, Data Analytics and Machine Learning
Professor Aditya K. Jagannatham
Department of Electrical Engineering
Indian Institute of Technology, Kanpur
Lecture 44
OFDM System: Circulant Matrices and Properties

Hello, welcome to another module in this massive open online course. So, we are looking at OFDM and OFDM system and the specific application of linear algebra and the modelling and analysis of this system. So, let us continue our discussion.

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So, we are looking at OFDM and where we are left in the last module is the following thing: considering an ISI channel with the channel taps h_0, h_1, h_{N-1} where N equals number of sub carriers, where N equals the number of sub carriers. What we had said is that these are the channel taps and the channel matrix if you look at the OFDM system after the cyclic prefix and so on, the channel matrix can be has a very interesting structure.

So, this will be h_0, h_{N-1}, h_{N-2} so on h_1 and then h_1 comes over here h_0 pushes to the right h_{N-1} also pushes to the right you have h_2 then h_2 comes rotates here, you have 0, you have h_3 so on and the last row will be h_{N-1}, h_{N-2} so on until h_0 .

And what we had said is very interestingly if you look at each row, each row is obtained by circularly shifting the previous row at the same time each column is also obtained by circularly shifting the previous column and therefore, is a very interesting matrix such a matrix is known as a circulant matrix and this is the characteristic feature of OFDM circulant.

This is the circulant matrix with if you look at the first column, if you specifically look at this is the first column, the first column basically comprises of the channel taps. So, how to construct this matrix? Simply take the right, the first column as the channel taps h_0, h_1 up to h_{N-1} and then each subsequent column you shift it down shifted by 1 and the 1 that pushes down you bring it back onto the top of the next column.

So, the second column will be h_{N-1}, h_0, h_1 up to h_{N-2} . Third column will be h_{N-2}, h_{N-1}, h_0 so on up to h_{N-3} and so on and let us now look at interesting properties of this matrix.

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PROPERTY OF H_c : $W = e^{-j\frac{2\pi}{N}}$

EIGENVECTORS OF H_c ARE OF THE FORM

$$\bar{f}_k = \frac{1}{N} \begin{bmatrix} 1 \\ W^{-k} \\ W^{-2k} \\ \vdots \\ W^{-(N-1)k} \end{bmatrix}$$

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\bar{f}_k is the k^{th} col of IFFT matrix.

$$\Rightarrow \frac{1}{N} \cdot F_{FFT}^H \cdot F_{FFT} = I$$
 ~ Roughly behaves like a unitary matrix

$$N = 2 \quad W = e^{-j\frac{2\pi}{N}} = -e^{-j\frac{\pi}{2}} = e^{-j\pi} = -1$$

$$F_{IFFT} = \frac{1}{2} \begin{bmatrix} 1 & 1 \\ 1 & W^{-1} \end{bmatrix} = \frac{1}{2} \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix}$$

Now, what are the interesting properties? The most interesting properties now, there are several the most interesting property of this matrix is in fact the property of any circulant matrix, property of the matrix H_c is that the Eigen vector, this is very interesting; the Eigen vectors H_c are of the form that is if you look at the k th Eigen vector, I can write this as $f_{bar k}$ which is $\frac{1}{N}$ times 1 W raised to minus k , W raised to minus $2k$, W raised to 2 minus N minus 1 k .

This is the k th Eigen vector where you remember W equal to e raised to minus j 2π over N , that is the definition of W . Let me just quickly check how we define W ? W equal to if you remember W equal to, we have W equal to e raise to minus j 2π over N indeed and now, if you look at this, this quantity this is nothing but what is this $f_{bar k}$? If you look at this this $f_{bar k}$ is nothing but if you observe this closely this is the k th column of the IFFT matrix.

This is the k th column of the IFFT matrix and that is the interesting property of this. It is very interesting property and remember that is what we said so, it is naturally because it is circular that it gives the impression that it is periodic. So, it naturally it has to have relation to sinusoids at complex exponentials and what is that relation? That relation is essentially if you look at this Eigen vector $f_{bar k}$ the Eigen vector of this H_c is essentially given by the k th column of the IFFT matrix, this is the interesting property and let us prove this.

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Channel taps: $h(0) h(1) h(2) \dots h(N-1)$

CIRCULAR matrix

$$H_c = \begin{bmatrix} h(0) & h(N-1) & h(N-2) & \dots & h(1) \\ h(1) & h(0) & h(N-1) & \dots & h(2) \\ h(2) & h(1) & h(0) & \dots & h(3) \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ h(N-1) & h(N-2) & \dots & \dots & h(0) \end{bmatrix}$$

r th row: $h(r-0) \quad h(r-1) \quad \dots \quad h(r-(N-1))$
mod N .

FIRST COL = Channel Taps.

PROPERTY OF H_c : $W = e^{-j2\pi/N}$

EIGENVECTORS OF H_c ARE OF THE FORM

$\begin{bmatrix} 1 \\ 1 \\ \vdots \\ 1 \end{bmatrix}$

Consider $H_c \bar{F}_k$

r th row in this product.

$$\frac{1}{N} \sum_{n=0}^{N-1} h(r-n) W^{-nk}$$

$r-n=m \Rightarrow r-m=n$

$$= \frac{1}{N} \sum_{m=0}^{N-1} h(m) W^{-(r-m)k}$$

modulo N

$$= \frac{1}{N} \left(\sum_{m=0}^{N-1} h(m) W^{mk} \right) W^{-rk}$$

$$= \frac{1}{N} \left(\sum_{m=0}^{N-1} h(m) e^{-j2\pi mk/N} \right) W^{-rk}$$

k th DFT coefficient of channel taps: $H(k)$

$$\left[H_c \bar{F}_k \right] = H(k) \cdot \frac{1}{N} W^{-rk}$$

r th element.

So, let us try to see how this is, this is as follows can be seen as follows: if you look at this now, first let us realize that the r th column, r th row if you look at the r th row that will be $h_{r,0}$, $h_{r,1}$ so on $h_{r,N-1}$ can, the only thing I need to do here is I can add modulo N . So, this will be $h_{r,0} \text{ modulo } N$, $h_{r,1} \text{ modulo } N$ so on and $h_{r,N-1} \text{ modulo } N$.

So, this small N here this indicates modulo N and therefore, if you look at the r th row and if you look at the inner product with this that is let us consider, consider the product $H \mathbf{c}$ times \mathbf{f}^H and let us look at what happens to the r th row in this product. That is output is a vector and we are asking the question what is our element of that vector and that will be $\frac{1}{N} \sum_{n=0}^{N-1} h_{r,n} W^{-nk}$.

This is once again to remind you this indicates that this is modulo which is equal to $\frac{1}{N} \sum_{n=0}^{N-1}$ or now substitute $n = r - m$, now make $r - m = m$, substitute $r - m = m$ So, m also goes from 0 to $N - 1$ this will become h_m there is no need for modulo N because m anywhere goes to 0 to $N - 1$.

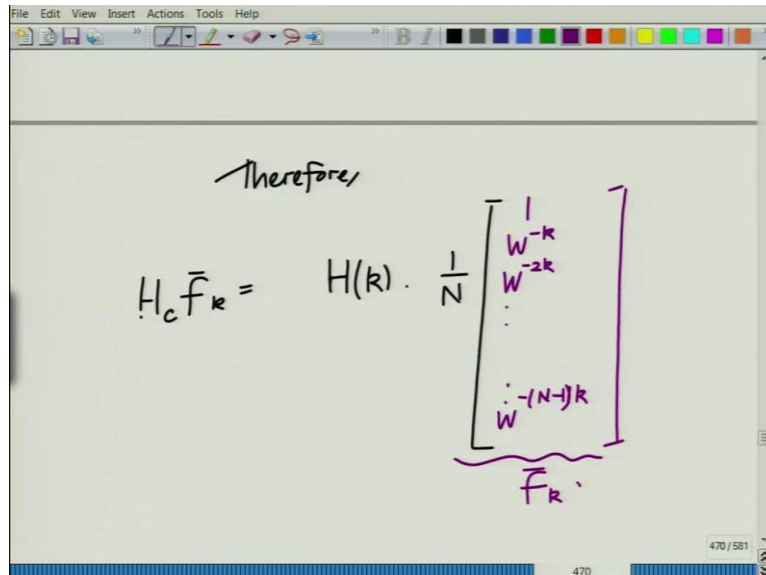
So, modulo $m \text{ modulo } N$ will simply be m because m is going from 0 to $N - 1$. So, therefore, this will be $h_m W^{-nk}$. So, this implies $r - m = m$. This implies $r - m = m$. So, this will be $r - m = m$ or in this case plus $m k$ times this other quantity will be common W^{-rk} and now, if you look at this quantity, this is interesting.

This is summation, this is nothing but if substitute for W this will be $\frac{1}{N} \sum_{m=0}^{N-1} h_m e^{-j 2\pi m k / N}$ times W^{-rk} and if you look at this quantity this is nothing but the FFT or the DFT coefficient of the channel taps. This is nothing but the in fact this is the k th DFT coefficient.

In fact, there is a k th DFT coefficient of the channel tap. So, I can write this so, this is essentially your H_k . So, this is basically your H_k times $\frac{1}{N} W^{-rk}$ and if you look at $H \mathbf{c}$ times \mathbf{f}^H and if you ask the question, what is the r th element that is what is our element of this? That is simply the H_k that is the k th DFT coefficient H_k times $\frac{1}{N} W^{-rk}$. This is the interesting aspect.

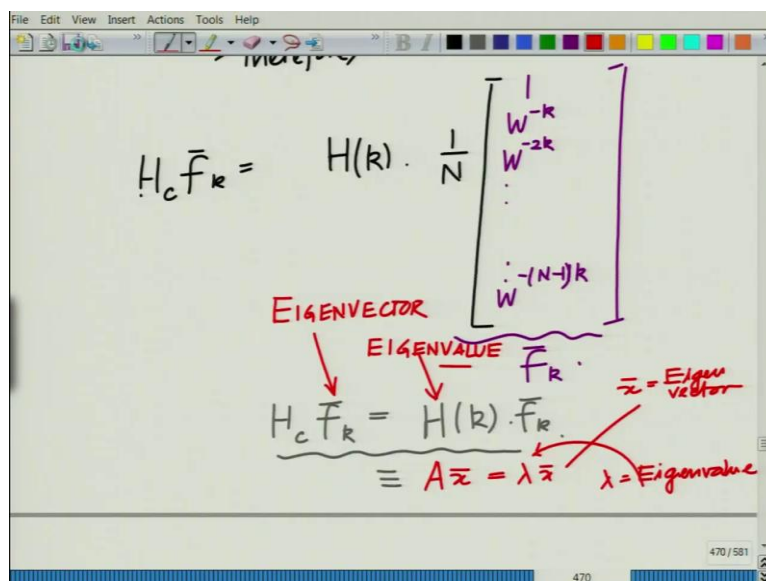
So, $H_c \bar{f}_k$ where H_c is the k th DFT quotient that is summation m equal to 0 to $N-1$ $h(m) e^{-j 2 \pi m k / N}$. That is if you take the channel taps into the k th DFT coefficient that is what it is. Now, write all this, so therefore, if you write $H_c \bar{f}_k$.

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Therefore,

$$H_c \bar{f}_k = H(k) \cdot \frac{1}{N} \begin{bmatrix} 1 \\ W^{-k} \\ W^{-2k} \\ \vdots \\ W^{-(N-1)k} \end{bmatrix} \bar{f}_k$$



Therefore,

$$H_c \bar{f}_k = H(k) \cdot \frac{1}{N} \begin{bmatrix} 1 \\ W^{-k} \\ W^{-2k} \\ \vdots \\ W^{-(N-1)k} \end{bmatrix} \bar{f}_k$$

EIGENVECTOR

EIGENVALUE

$$H_c \bar{f}_k = H(k) \bar{f}_k$$

$\bar{z} = \text{Eigen vector}$

$\lambda = \text{Eigenvalue}$

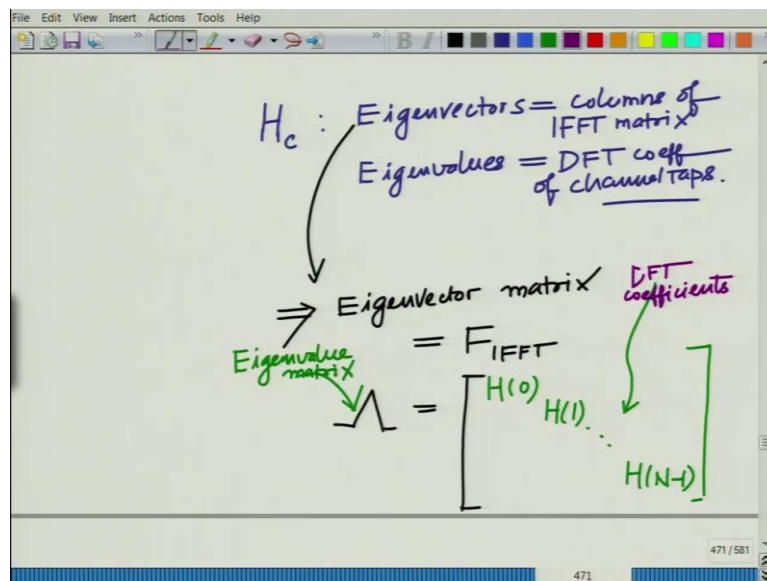
$$\equiv A \bar{z} = \lambda \bar{z}$$

So therefore, write all the elements $H_c \bar{f}_k$ equals H_c . This will be H_k or this will be H_k times $1/N$. What are the elements? 1 first element will be W , each element is W raised to minus $r k$. So, W raised to minus k , W raised to minus $2 k$ so on, W raised to minus $N-1 k$ and in fact, if you look at this now, interestingly if you look at this, this is nothing but essentially the vector \bar{f}_k . So, that is the interesting thing. So, essentially you have the relation $H_c \bar{f}_k$ equals $1/N$ are equals. In fact, H_k times \bar{f}_k .

So, this is the form $A \bar{x}$. If you look at this, this is similar to $A \bar{x} = \lambda \bar{x}$, where \bar{x} equals the you know very well now, \bar{x} equals Eigen vector λ , equals the Eigen value. So, this implies that in this \bar{f}_k , this is the Eigen vector of H_c and this DFT coefficient is the Eigen value. So, that is the interesting aspect of this.

So, the H_k is that is a DFT coefficient, the Eigen vectors are the columns of the IFFT matrix and Eigen values are given by the FFT or the DFT of the channel taps. So, this is the interesting property of the circulant matrix.

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So, H_c has the following very interesting property. H_c Eigen vectors equals columns of IFFT matrix and Eigen values equals the DFT coefficient. Eigen values are the DFT coefficients of the channel taps and this also implies that the Eigen value matrix, implies the Eigen value matrix if you look Eigen value matrix, this is equal to simply the column of the matrix which contains the IFFT columns of the IFFT matrix.

So, this is simply the IFFT matrix. So, this is the Eigen, I am sorry this is the Eigen vector matrix and Eigen value matrix λ , the Eigen value matrix is contains the DFT coefficients. So, this is the Eigen value matrix, this is the Eigen value matrix. So, this is the Eigen value matrix basically contains the DFT coefficients and we know what is Eigen value decomposition?

The Eigen value decomposition of a square matrix is nothing but the Eigen value matrix, I am sorry the Eigen vector matrix times the diagonal matrix of Eigen values times the inverse of the Eigen vector matrix. Eigen vector matrix is the IFFT matrix inverse of the Eigen vector

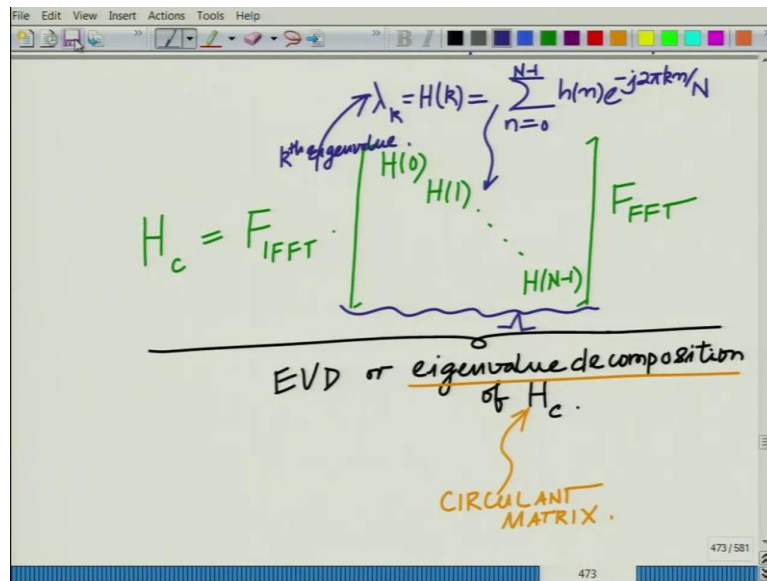
matrix is nothing but the inverse of the IFFT matrix which is nothing but the FFT matrix and that essentially completes this analysis.

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$H_c = \text{Eigenvector matrix} \times \text{Diagonal matrix of Eigenvalues.}$
 $\times \text{inverse of Eigenvector matrix}$
 $\Rightarrow H_c = F_{\text{IFFT}} \cdot \Lambda \cdot F_{\text{IFFT}}$
DFT coefficients

$H_c = F_{\text{IFFT}} \begin{bmatrix} H(0) & & \\ & \dots & \\ & & H(N-1) \end{bmatrix} F_{\text{FFT}}$
DFT coefficients

$H_c = F_{\text{IFFT}} \begin{bmatrix} H(0) & & \\ & H(1) & \\ & & \dots \\ & & & H(N-1) \end{bmatrix} F_{\text{FFT}}$



So, you have now very interestingly if you look at it, H_c any matrix let me write the general property. This can be written as the product of Eigen vector matrix times diagonal matrix of Eigen values, Eigen values times inverse of Eigen vector matrix which in this case implies H_c equal to Eigen vector matrix is the IFFT matrix, F_{IFFT} times the diagonal matrix of Eigen values which contains the DFT coefficients times F_{IFFT} inverse.

But F_{IFFT} inverse equals F_{FFT} . So, therefore, this is writing it explicitly. F_{IFFT} times your H_0 up to H_{N-1} times F_{FFT} . Let us see and this essentially, so if you look at this this is the Eigen value decomposition. Let me write this thing once again H_c equal to F_{IFFT} times H_0, H_1 up to H_{N-1} times F_{FFT} and this is essentially the what we call as the EVD.

This is the Eigen value decomposition of H_c which is now. So, for the circulant matrix in fact for any circle in matrix Eigen value decomposition is given by the IFFT matrix times the diagonal matrix of Eigen values which are nothing but the DFT or FFT coefficients of the channel taps followed by the FFT matrix and in fact, let us also write the formula for the Eigen value you have λ_k equals $H(k)$ which is basically summation m equal to 0 or you can write it as n equal to 0 to $N-1$ $h(n) e^{-j2\pi kn/N}$.

So, this is your kth Eigen value and this is your matrix Λ that is the diagonal matrix of Eigen values. So, this is essentially the interesting aspect or this is the most interesting property of the circulant matrix and now, what we are going to do is we are going to use this matrix in the OFDM system model and we are going to simplify the OFDM system model and observe something very interesting and that is what we are going to do in the next and the

final module. So, let us stop here and we will continue this discussion in the next module.

Thank you very much.