

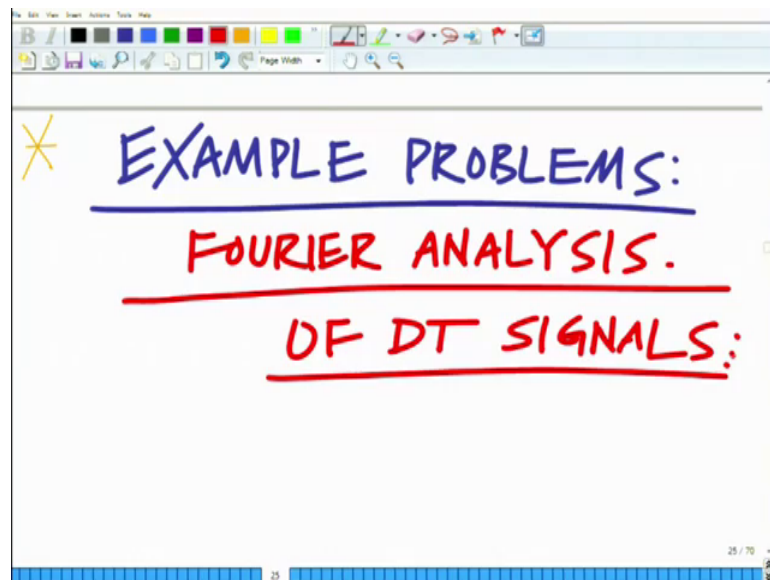
Principles of Signals and Systems
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Lecture - 62

Example Problems: DFS Analysis of Discrete Time Signals, Problems on DTFT

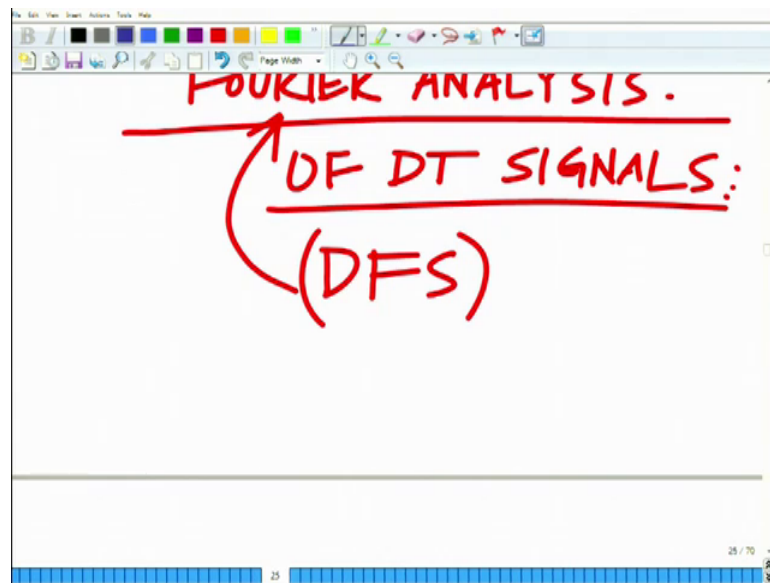
Hello, welcome to another module in this massive open online course. So, we are looking at example problems right, we have started looking at example problems for the Fourier analysis the Fourier analysis of discrete time signals and we have started with example problems for the discrete Fourier series, all right.

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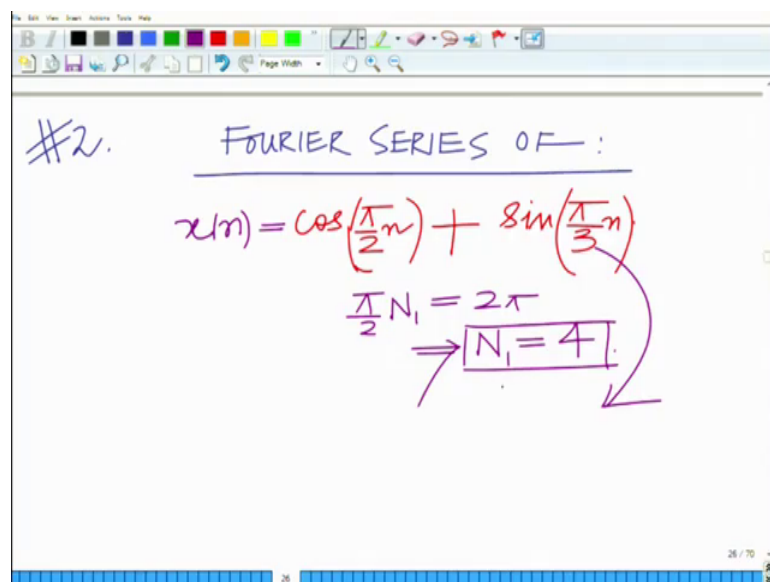
So, let us continue our discussion. We are looking at example problems for the Fourier analysis the Fourier analysis of DT signal discrete type signal and in particular we have started with the discrete Fourier series all right, for discrete time periodic signals all right.

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So, let us continue our discussion and let us look at the next problem. So, if you looked at problem number 1 yesterday if I remember correctly in the previous module. So, let us look at the second problem that is problem number 2 which is the following.

(Refer Slide Time: 01:46)



So, we want to find the Fourier series the Fourier series of cosine pi by 2 n plus cosine pi by 2 n plus sin pi by 3 and naturally you can see that these are periodic in the period of this. So, this is a periodic signal all right.

So, this is your $x[n]$ and as you let cosine πn by 2 the period of this can be found as follows πn by 2 into k or πn by 2 into N_1 let us call this N_1 equals 2π implies because cosine x is cosine x plus 2π . So, N_1 equals basically 4 that is the period of the first signal. Similarly for the sin signal we have πn by 3 N_2 equals 2π which implies N_2 equals 6.

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$$\frac{\pi}{3} N_2 = 2\pi$$

$$\Rightarrow N_2 = 6$$

$$N_0 = \text{Period of } x(n)$$

$$= \text{LCM}(4, 6) = 12$$

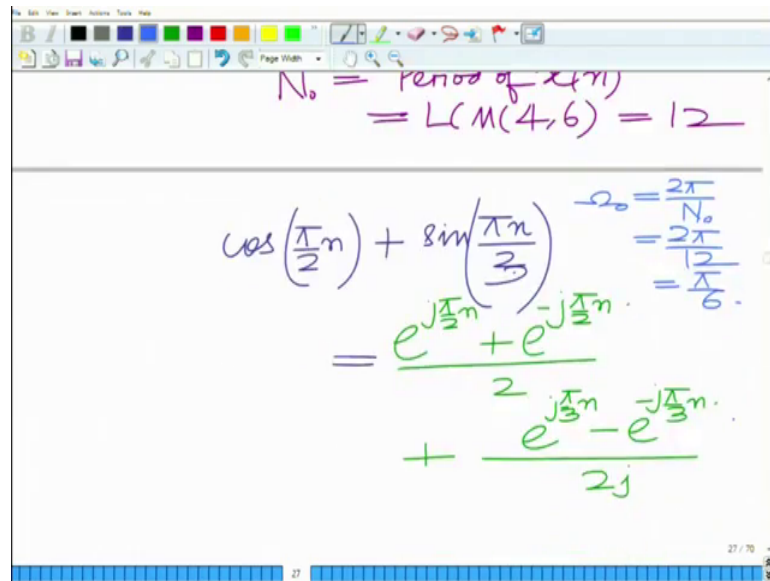
$$\cos\left(\frac{\pi n}{2}\right) + \sin\left(\frac{\pi n}{3}\right)$$

So, period is of the first component that is cosine πn by 2 n is 4 and period of the second component that is sin πn by 3 n is 6 and the period of the sum correct is therefore, the least common multiple because it has to be a multiple of both and therefore, the minimum possible size length is the least common multiple of these.

So, therefore, the period fundamental period of this sum signal is N_0 which is the least common multiple of 4 and 6 which is 12. So, N_0 equals period of $x[n]$ equals LCM of 4 and 6 the least common multiple of 4 and 6 which is 12 ok.

And now, I can write using the properties of cosine πn by 2 n using the properties of complex exponentials πn by 3. I can write this as this can be written as well $e^{j\pi n/2}$ plus $e^{-j\pi n/2}$ divided by 2 plus $e^{j\pi n/3}$ and sin πn by 3 n is $e^{j\pi n/3}$ and minus $e^{-j\pi n/3}$ divided by $2j$.

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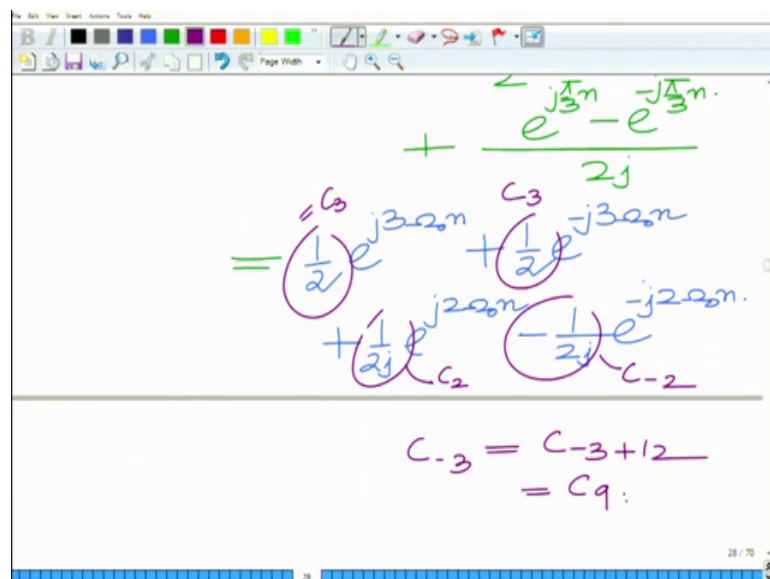
$$N_0 = \text{period of } x(n) = \text{LCM}(4, 6) = 12$$

$$\cos\left(\frac{\pi n}{3}\right) + \sin\left(\frac{\pi n}{3}\right) = \frac{e^{j\frac{\pi n}{3}} + e^{-j\frac{\pi n}{3}}}{2} + \frac{e^{j\frac{\pi n}{3}} - e^{-j\frac{\pi n}{3}}}{2j}$$

$\omega_0 = \frac{2\pi}{N_0} = \frac{2\pi}{12} = \frac{\pi}{6}$

So, therefore, this is equal to well half. Now, look at this pi by so omega naught, so if you look at omega naught. Now, the lcm N naught equals 12. So, we have omega naught equals 2 pi over N naught equals 2 pi over 12 equals pi by 6. So, pi by 2 is basically 3 omega naught. So, omega naught is basically pi by 6. So, pi by 2 is 3 omega naught and pi by 3 is 2 omega naught ok.

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$$= \frac{1}{2} e^{j3\omega_0 n} + \frac{1}{2} e^{-j3\omega_0 n} + \frac{1}{2j} e^{j2\omega_0 n} - \frac{1}{2j} e^{-j2\omega_0 n}$$

$C_{-3} = C_{-3+12} = C_9$

So, I can write this as half $e^{j3\omega_0 n}$ plus half $e^{-j3\omega_0 n}$ over $2j$ plus $e^{j2\omega_0 n}$ minus $e^{-j2\omega_0 n}$ over $2j$.

And clearly you can see this is the coefficient of $e^{j3\omega_0 n}$ that is half. So, this is equal to C_3 , this is C_{-3} , this is C_2 the DFS and this is C_{-2} basically, but C_{-3} everything is modulo N . So, C_{-3} is C_{N-3} that is C_9 and C_{-2} is C_{N-2} that is C_{10} .

(Refer Slide Time: 06:28)

Handwritten notes on a whiteboard showing the derivation of coefficients C_3 , C_{-3} , C_2 , and C_{-2} using modulo N arithmetic. The notes include the following equations:

$$C_{-3} = C_{-3+12} = C_9$$

$$C_{-2} = C_{-2+12} = C_{10}$$

$$C_3 = \frac{1}{2j} \quad C_9 = \frac{1}{2j}$$

$$C_2 = \frac{1}{2j} \quad C_{10} = -\frac{1}{2j}$$

So, this is basically. So, what we have is from the above discrete Fourier series is C_3 equals half C_{-3} that is C_9 is also equal to half C_2 equals $\frac{1}{2j}$ that is $\frac{1}{2j}$ which is also you can write it as $\frac{j}{2}$ and C_{10} is equal to minus $\frac{1}{2j}$.

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$$C_3 = \frac{1}{2} \quad C_9 = \frac{1}{2}$$

$$C_2 = \frac{1}{2j} \quad C_{10} = -\frac{1}{2j}$$

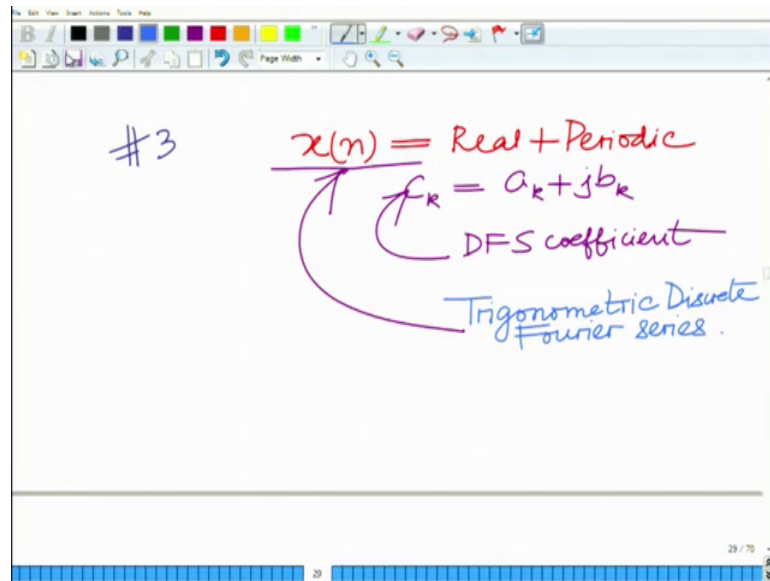
$$\text{DFS } x(n) = \frac{1}{2j} e^{j2\omega n} + \frac{1}{2} e^{j3\omega n} + \frac{1}{2} e^{j9\omega n} - \frac{1}{2j} e^{j10\omega n}$$

↑
DFS Representation of signal $x(n)$

And therefore, I can write the signal as the discrete Fourier series representation $x(n)$ as $C_3 e^{j2\omega n} + C_2 e^{j\omega n} + C_9 e^{j9\omega n} + C_{10} e^{j10\omega n}$. That is $\frac{1}{2j} e^{j2\omega n} + \frac{1}{2} e^{j3\omega n} + \frac{1}{2} e^{j9\omega n} - \frac{1}{2j} e^{j10\omega n}$.

So, this is the discrete Fourier series representation of the signal $x(n)$. So, this is the DFS. This is the DFS representation of the signal $x(n)$ all right. Discrete Fourier series representation of the signal $x(n)$ so that completes problem number 2 let us look at problem number 3.

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And problem number 3 as follows let x_n be a real periodic sequence. So, let us write it as this let us write it in compact form, x_n is real plus periodic ok. So, it is a real periodic sequence.

And the DFS coefficient, so given that the DFS coefficient is $a_k + j b_k$ DFS coefficient C_k can be expressed as $a_k + j b_k$ and we want to find the trigonometric discrete Fourier series. Now, for this we want to find the trigonometric discrete Fourier series, the discrete Fourier series coefficients of C_k we want to find the trigonometric discrete.

So, we want to find the trigonometric Fourier discrete Fourier series given the discrete Fourier series the complex discrete Fourier series all though you can also see the complex exponential discrete Fourier series coefficients C_k ok. And let us also consider the scenario for simplicity where N is odd ok. So, let us consider the scenario where N is odd which also implies that $N - 1$ is even ok.

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$N_0 = \text{odd}$
 $\Rightarrow N_0 - 1 = \text{even}$

$$x(n) = \sum_{k=0}^{N_0-1} C_k e^{jk\omega_0 n}$$

$$= C_0 + \sum_{k=1}^{N_0-1} C_k e^{jk\omega_0 n}$$

Now, the discrete Fourier series is given as $x(n) = \sum_{k=0}^{N_0-1} C_k e^{jk\omega_0 n}$ which is remember N_0 is odd, so $N_0 - 1$ is even..

So, this is C_0 plus summation k equal to 1 to $N_0 - 1$ $C_k e^{jk\omega_0 n}$. So, N_0 is odd $N_0 - 1$ is even.

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$$= C_0 + \sum_{k=1}^{N_0-1} C_k e^{jk\omega_0 n}$$

$$= C_0 + \sum_{k=1}^{N_0-1} C_k e^{jk\omega_0 n} + C_{N_0-k} e^{j(N_0-k)\omega_0 n}$$

Since $x(n) = \text{real}$.
 $\Rightarrow C_k = C_{-k}^*$
 $= C_{N_0-k}^*$

So, I can write this as C_0 plus summation k equal to 1 to $N_0 - 1$ by 2 $C_k e^{jk\omega_0 n} + C_{N_0-k} e^{j(N_0-k)\omega_0 n}$

omega naught n. So, I am splitting it into 2. I am writing at and varying the index only from k equal to 1 to N naught minus 1 by 2 and writing it as a sum of C k e raise to j k omega naught n plus C n minus k e raise to j k n N naught minus k times n.

Now, if you look at this since this is a real signal this implies C k and C naught n minus k which is C of minus k or conjugates of each other. So, C k equals C minus k conjugate equals C N naught minus k conjugate, ok.

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$$\begin{aligned}
 & e^{j(N_0-k)\Omega_0 n} \\
 &= e^{j2\pi n} \cdot e^{-jk\Omega_0 n} \\
 &= \underbrace{1}_{e^{j2\pi n}} e^{-jk\Omega_0 n} \\
 &= e^{-jk\Omega_0 n}
 \end{aligned}$$

And further if you look at e raise to j N naught e raise to j N naught minus k omega naught n. This is e raise to j N naught into omega naught is 2 pi e raise to j 2 pi n into e raise to minus j k omega naught n and e raise to j 2 pi n is basically unity ok. So, this is unity and therefore, this is e raise to minus j. This is e raise to minus j k omega naught n, ok.

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The image shows a whiteboard with handwritten mathematical equations. The top equation is:

$$x(n) = C_0 + \sum_{k=1}^{N/2} C_k e^{jk\omega n} + C_k^* e^{-jk\omega n}$$

The second equation shows the sum being rewritten as twice the real part of the first term:

$$= C_0 + \sum_{k=1}^{N/2} 2 \operatorname{Re} \{ C_k e^{jk\omega n} \}$$

The third equation defines the real part of the complex exponential term:

$$\operatorname{Re} \{ C_k e^{jk\omega n} \} = \operatorname{Re} \{ (a_k + jb_k)(\cos k\omega n + jsin k\omega n) \}$$

So, basically what this tells us is that this is equal to, so I can write rather recast $x(n)$ as C_0 plus summation k equal to 1 to $N/2$ minus 1 divided by 2, $C_k e^{jk\omega n}$ plus $C_k^* e^{-jk\omega n}$, ok.

Now, C_k , now, you can see this $C_k^* e^{-jk\omega n}$ is nothing but $C_k e^{jk\omega n}$ conjugate ok. So, $C_k^* e^{-jk\omega n}$ is basically $C_k e^{jk\omega n}$ conjugate ok. So, now, we have $C_k e^{jk\omega n}$ plus its conjugate. So, basically what remains is what it what is this leads to its twice its real part twice the real part of $C_k e^{jk\omega n}$ ok.

So, this will be C_0 just simplify this will be C_0 plus summation k equal to 1 to $N/2$ minus 1 by 2 twice real part of $C_k e^{jk\omega n}$. And the real part of $C_k e^{jk\omega n}$ equals basically the real part of $a_k \cos k\omega n + j b_k \sin k\omega n$. Which is equal to the real part is well $a_k \cos k\omega n + b_k \sin k\omega n$ ok.

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The image shows a whiteboard with handwritten mathematical notes. At the top, there is a partial equation: $-j e^{jk\omega_0 n} + j e^{jk\omega_0 n}$. Below it, the expression is equated to $a_k \cos(k\omega_0 n) - b_k \sin(k\omega_0 n)$. A horizontal line separates this from the main equation: $x(n) = C_0 + 2 \sum_{k=1}^{N_0-1} a_k \cos(k\omega_0 n) - b_k \sin(k\omega_0 n)$. The text "Trigonometric DFS" is written above the summation, with "N₀ = odd" written below it. Below the main equation, two definitions are given: $a_k = \text{Re}\{C_k\}$ and $b_k = \text{Im}\{C_k\}$. Arrows point from these definitions to the corresponding terms in the summation.

And therefore, substituting this now, you have $x(n)$ the trigonometric discrete Fourier series can be obtained as C_0 plus twice summation k equals 1 to $N_0 - 1$ by 2, $a_k \cos(k\omega_0 n) - b_k \sin(k\omega_0 n)$. Whereas I already told you that this a_k and b_k are obtained from the DFS coefficient C_k such that C_k is $a_k + j b_k$.

So, these are given from these are basically the real part a_k is the real part and b_k is the imaginary part of the DFS coefficient C_k , ok. So, a_k equal to real part of C_k and b_k equals basically the imaginary part of C_k all right, ok. And this is basically your trigonometric Fourier series.

But remember when also remember this is the trigonometric discrete Fourier series, but remember this is when N_0 equals N_0 is even sorry or N_0 is odd sorry N_0 is odd all right. Similarly you can find the discrete Fourier series the trigonometric discrete Fourier series when N_0 is even in that case $N_0 - 1$ be odd. So, you cannot apply the same technique all right.

So, basically that brings us to the next problem which is problem number 4, if I remember correctly. So, this brings, this brings us to problem number 4.

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The image shows a presentation slide with a white background and a blue border. At the top, the title "PROBLEMS ON DTFT" is written in blue, underlined. Below the title, the text "#4) $x(n) = u(n) - u(n-N)$ " is written in purple. A purple arrow points from the expression to "DTFT = ?". Below this, a red piecewise function is defined: $x(n) = \begin{cases} 1 & 0 \leq n \leq N-1 \\ 0 & \text{otherwise} \end{cases}$. The slide also features a toolbar at the top with various icons and a footer at the bottom with the number "33 / 70".

Find the discrete time Fourier transform. So, now, we are entering problems on discrete time Fourier transform that you just you completed the Fourier series these are the problems on DTFT ok, the problem on DTFT. This is number 4. Let us say $x[n]$ equals $u[n] - u[n - N]$ what is the corresponding DTFT of this signal.

So, this is a $u[n] - u[n - N]$, this implies if you can see this is clearly $x[n]$ equals 1, 0 less than n less than equal to $n - 1$ and 0 and 0 otherwise ok.

This is your $x[n]$ that is 1 only in a window of 0 to capital $N - 1$. And outside this window that is for n less than 0 strictly less than 0 and n strictly greater than $n - 1$ or basically n greater than or equal to capital N this is 0 ok.

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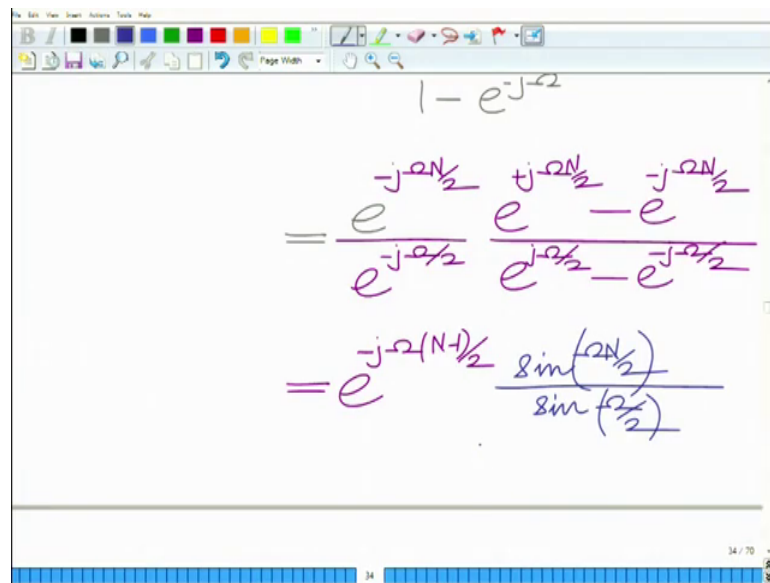
The image shows a handwritten derivation of the Discrete-Time Fourier Transform (DTFT) of a rectangular pulse signal. The derivation is written on a whiteboard with a toolbar at the top and a page number '34' at the bottom right. The equations are as follows:

$$X(\omega) = \sum_{n=-\infty}^{\infty} x(n) e^{-j\omega n}$$
$$= \sum_{n=0}^{N-1} e^{-j\omega n}$$
$$= \frac{1 - e^{-j\omega N}}{1 - e^{-j\omega}}$$

And therefore, $x(\omega)$, now, if you find the DTFT that is given as $x(\omega)$ equals summation n equals minus infinity to infinity $x(n) e^{-j\omega n}$.

Now, there is only nonzero from 0 to $n-1$. So, this is and in that window it is one. So, this is simply summation n equal to 0 to $n-1$ $e^{-j\omega n}$ which is a geometric series and the sum of this is $1 - e^{-j\omega n}$ divided by $1 - e^{-j\omega}$. Which is now, if you write take $e^{-j\omega N/2}$ common in the numerator and $e^{-j\omega/2}$ common in the denominator.

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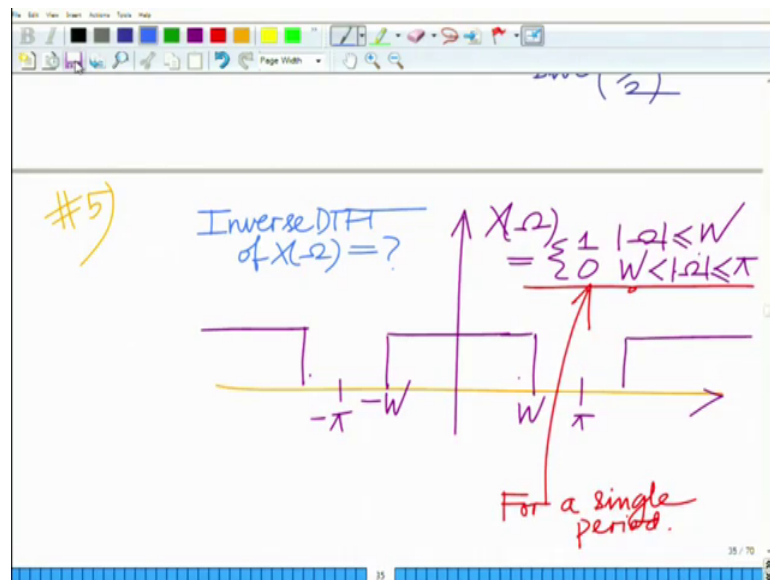
$$\begin{aligned}
 & 1 - e^{j\omega N} \\
 &= \frac{e^{-j\omega N/2}}{e^{-j\omega/2}} \frac{e^{j\omega N/2} - e^{-j\omega N/2}}{e^{j\omega/2} - e^{-j\omega/2}} \\
 &= e^{-j\omega(N-1)/2} \frac{\sin(\omega N/2)}{\sin(\omega/2)}
 \end{aligned}$$

So, this will be e raise to minus j omega N by 2 minus or in that e raise to plus j omega N by 2 minus e raise to minus j omega N by 2.

In the numerator and the denominator it will be e raise to j omega by 2 minus e raise to minus j omega by 2 which is equal to e raise to minus j omega n minus 1 divided by 2 times e raise to j omega n by 2 minus e raise to minus j omega capital N over 2 this is twice j sin omega N over 2.

So, and you can see that the 2 j factor in the numerator and denominator cancels. So, what you are left with a sin n omega by 2 in the numerator and sin omega by 2 in the denominator ok, sin omega 2 by 2 in the denominator.

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Let us find the DTFT of the inverse DTFT rather of another signal. So, this is problem number 5, you have x of ω .

So, this is remember the DTFT is periodic. So, this is your π periodic with periodicity 2π . So, we have x of ω this is your x of ω which is equal to 1, as shown magnitude of ω less than or equal to W and 0 otherwise. This is in one period or let us write it is 0 W less than more ω less than less than or equal to less than or equal to less than or equal to π , ok.

So, basically this is this thing this describes it in one period, this is remember this is also because this is always a periodic signal. So, this is for a single period this is for a single period. So, this is x of ω the inverse DTFT.

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The image shows a whiteboard with handwritten mathematical equations. At the top, it says "Inverse DTFT". Below that, the first equation is $x(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\omega) e^{j\omega n} d\omega$. The second equation is $= \frac{1}{2\pi} \int_{-W}^W e^{j\omega n} d\omega$. The third equation is $= \frac{1}{2\pi} \cdot \frac{e^{j\omega n}}{jn} \Big|_{-W}^W$. The whiteboard has a toolbar at the top and a status bar at the bottom showing "36 / 70".

And what we need to do is we have to find the inverse DTFT of x of ω . The problem is; what is the inverse DT of this DTFT of this x of ω which is 1 in this interval of width $2W$ from $-W$ to W .

So, you can say this is discrete Fourier transform which is the bandwidth of $2W$ all right. It is one in this band W to W , ok. And the DTFT is given as inverse DTFT is given as remember the inverse DTFT formula is $x(n)$ equals $\frac{1}{2\pi}$ times integral from $-W$ to W .

Student: ω .

$X(\omega) e^{j\omega n}$ which is equal to $\frac{1}{2\pi}$. In fact, this is simply 1 from $-\pi$ to π therefore, this is one only in the interval $-W$ to W . So, this reduces to integral from $-W$ to W and in this interval it is 1.

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The image shows a whiteboard with a software interface at the top. The derivation is as follows:

$$\begin{aligned} &= \frac{1}{2\pi} \int_{-W}^W jn \, d\omega \\ &= \frac{1}{2\pi} \frac{e^{jWn} - e^{-jWn}}{jn} \\ &= \frac{1}{2\pi} \frac{2j \sin Wn}{jn} \\ &= \frac{\sin Wn}{\pi n} \end{aligned}$$

So, this is raise to $j \omega n$ $d \omega$ which is one over 2π e raised to $j \omega n$ over $j n$ evaluated between the limits minus W to W equals 1 over 2π e raise to $j \omega n$ minus e raise to $-j \omega n$ divided by $j n$ which is 1 over 2π e raise to $j W n$. So, this is $2 j \sin W n$ divided by $j n$. So, removing the j is the factor of 2 is also go away. So, what we have is $\sin W n$ by πn . So, basically that gives us $x n$ the inverse DTFT, $x n$ equals.

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The image shows a whiteboard with a software interface at the top. The final result is boxed and annotated:

$$x(n) = \frac{\sin Wn}{\pi n}$$

Inverse DTFT of given $X(\Omega)$.

So, this is basically the inverse DTFT of the, inverse DTFT of the inverse DTFT of the given signal ok, the corresponding in fact, a periodic time domain signals, all right.

So, we have looked at problems pertaining to the pertaining to the discrete Fourier series and also the DTFT started looking at problems pertaining to the DTFT in this module. Let us continue this discussion in the subsequent modules.

Thank you very much.