

Principles of Signals and Systems
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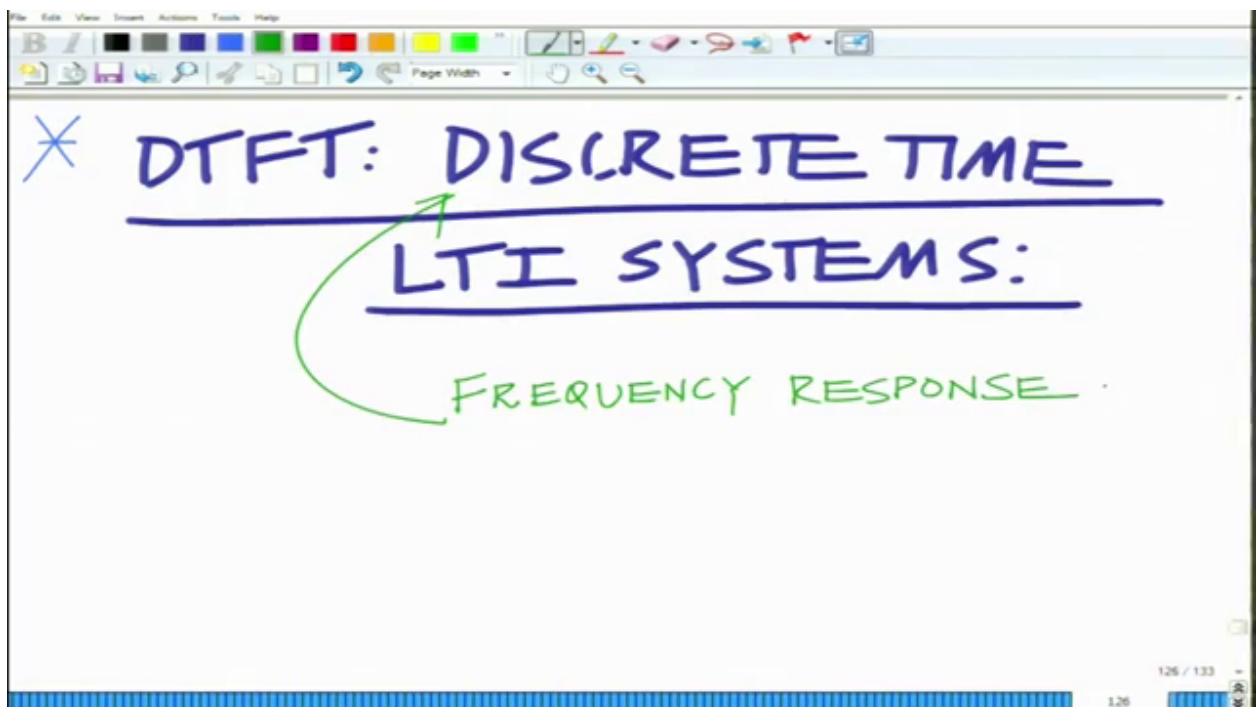
Lecture – 59

DTFT: Discrete Time LTI Systems- LTI Systems Characterized by Difference Equations

[noise]

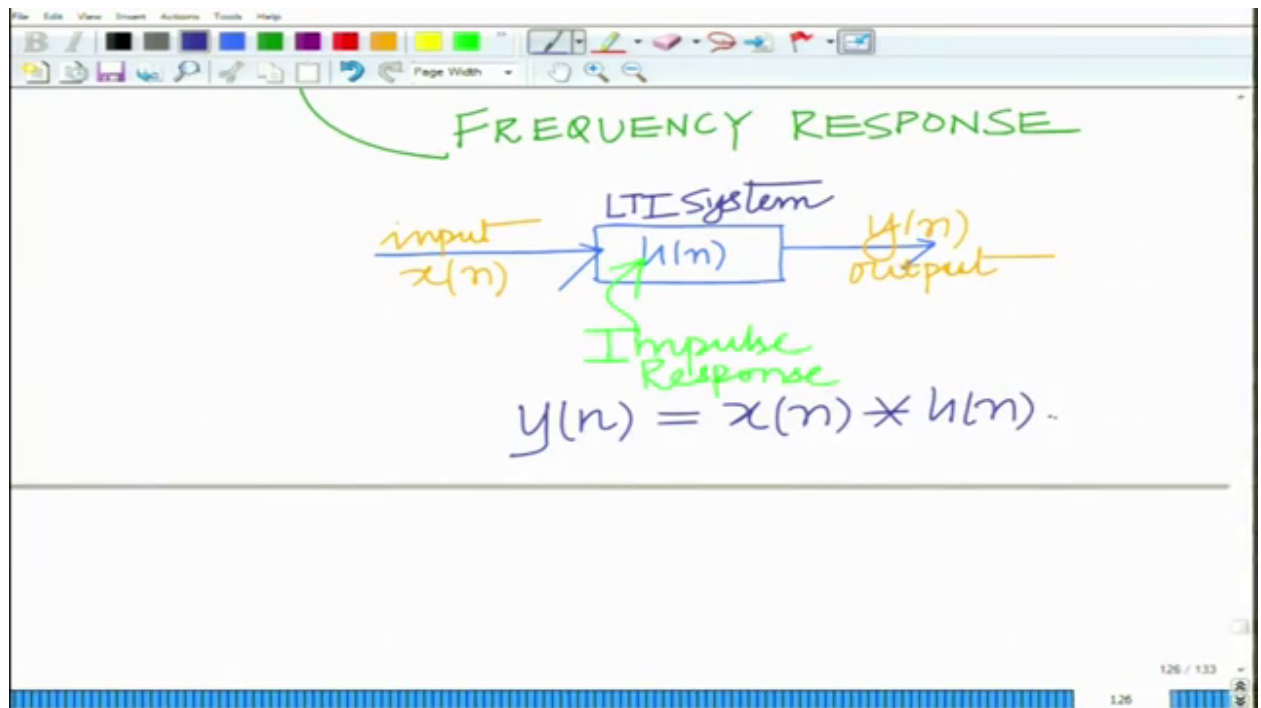
Hello welcome to another module in this massive open online course. So, we are looking at the d d t f t the discrete time fourier transform ah for the analysis of discrete time signals and systems and in this module ah we will look at the d t f t for the analysis of discrete time l t i systems ok. So, we [noise] intend to discuss [noise] the d t f t [noise]

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application of the d t f t for ah [noise] for discrete time [noise] [noise] in particular ah we would like to [vocalized-noise] ah look at it in terms of the frequency response.

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Of these discrete time l t i systems ok what is the [noise] frequency response [noise] ok. So, consider a discrete time l t i system with impulse response given by $h(n)$ ok [noise]. So, the input [noise] to this [noise] l t i system is $x(n)$ [noise] output is $y(n)$ [noise] ok and $h(n)$ denotes the impulse response of this system [noise] it denotes the impulse response of this system [vocalized-noise] and the input output relationship therefore,. So, this is your l t i system [noise] ok [vocalized-noise] the discrete time l t i system with input given by $x(n)$ output given by $y(n)$ ok $h(n)$ denotes the discrete time impulse response of this l t i system ok [vocalized-noise]

And now naturally we know that the output $y(n)$ is given as the convolution of $x(n)$ [noise] times [noise] $h(n)$ or $x(n)$ is the input $h(n)$ is the impulse response are now taking.

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$$y(n) = x(n) * h(n)$$

$$\xrightarrow{\text{DTFT}} Y(-\Omega) = X(-\Omega) H(-\Omega)$$

$$\Rightarrow \boxed{H(-\Omega) = \frac{Y(-\Omega)}{X(-\Omega)}}$$

Frequency Response Transfer Function

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The d t f t [noise] discrete time fourier transform we have y omega equals x omega [noise] [noise] times h omega which implies [noise] as usual the transfer function h omega [noise] equals y omega [noise] divided by [noise] x omega this is the [noise] transfer function ok there is a frequency response of the l t i system of the transfer function [noise] [vocalized-noise] frequency response of the transfer function [vocalized-noise] and now ah what happens if x n is periodic now

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Frequency Response Transfer Function

$$x(n) = \text{Periodic}$$

$$y(n) = ? \text{ Given } H(-\Omega)$$

$$x(n) = e^{j-\Omega_0 n}$$

$$y(n) = h(n) * x(n)$$

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lets say x_n the input signal is [noise] periodic in that case how do you obtain [noise] y_n ah given the impulse response that is given the [vocalized-noise] given h of ω [noise] [vocalized-noise].

Now, first observe that if x_n is a complex exponential [noise] [vocalized-noise] if x_n equals e raised to $j \omega$ naught n ok now the output y_n equals h [noise] n convolved with x_n [noise] ok which is basically equal to

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$$x(n) = e^{j\omega_0 n}$$

$$y(n) = h(n) * x(n)$$

$$= \sum_{k=-\infty}^{\infty} h(k) e^{j\omega_0(n-k)}$$

$$= \sum_{k=-\infty}^{\infty} h(k) \cdot e^{j\omega_0 n} \cdot e^{-j\omega_0 k}$$

or which [noise] equals the summation of [noise] k equals [noise] minus infinity to infinity [noise] [noise] h k e raised to j omega naught n minus k which is summation k equals minus infinity to infinity [noise] h k e raised to j omega naught n e raised to j omega naught k now this e raised to j omega naught n which is common i take that outside. So, that becomes

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The image shows a whiteboard with handwritten mathematical equations. At the top, the output $y(n)$ is expressed as a convolution sum: $y(n) = e^{j\omega_0 n} \sum_{k=-\infty}^{\infty} h(k) e^{-j\omega_0 k}$. The sum is identified as the frequency response $H(\omega_0)$. Below this, the input $x(n)$ is shown as $x(n) = e^{j\omega_0 n}$, which is labeled as an "Eigenfunktion" (eigenfunction). The resulting output is $y(n) = H(\omega_0) e^{j\omega_0 n}$.

$e^{j\omega_0 n}$ raised to $j\omega_0 n$ [noise] summation k equals minus infinity to infinity [noise] $h(k)$ [noise] $e^{-j\omega_0 k}$ which is simply [noise] $H(\omega_0)$ of ω_0 that is the dft at ω_0 that summation $h(k) e^{-j\omega_0 k}$ [vocalized-noise]. So, the output is simply given as $y(n)$ is simply given as $H(\omega_0)$ for the complex exponential at frequency ω_0 the output is simply $H(\omega_0) e^{j\omega_0 n}$. So, $y(n)$ [noise]. So, we have [noise] [vocalized-noise].

And this is interesting and we already seen this [noise]. So, if $x(n)$ [noise] is $e^{j\omega_0 n}$ then the output signal $y(n)$ is simply [noise] $H(\omega_0)$ multi plus scaled version of the input signal $(())$ $H(\omega_0) e^{j\omega_0 n}$ and therefore, this is also known as the eigen function ok or the eigen signal [noise] [vocalized-noise] this is the eigen function of this lti system because the output is simply a scaled version of scalar multiple times the input ok [vocalized-noise] now.

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$$x(n) = e^{j\Omega_0 n}$$

$$y(n) = H(\Omega_0) e^{j\Omega_0 n}$$

If $x(n)$ is Periodic.

$$x(n) = \sum_{k=0}^{N_0-1} C_k e^{jk\Omega_0 n}$$

$\Omega_0 = \frac{2\pi}{N_0}$
 DFS of $x(n)$
 Fundamental Freq.

If $x(n)$ is periodic [noise] then we realize that $x(n)$ can be expressed as a discrete Fourier series [noise] then we have [noise] $x(n)$ equal to summation [noise] k equal to zero to $N_0 - 1$ $C_k e^{jk\Omega_0 n}$ [noise] where Ω_0 equals [noise] $\frac{2\pi}{N_0}$ [noise] ok this is $C_k e^{jk\Omega_0 n}$ [noise] (()). So, what is this this is basically your discrete Fourier series [noise] [noise].

And this Ω_0 is basically the fundamental frequency [noise] ok [vocalized-noise]. So, it is expressed as a multiple is expressed as a sum of the complex explanation at the fundamental frequency and its harmonics ok k equal to zero to $N_0 - 1$ this is a discrete Fourier series ok now from linearity now we know that corresponding to each $e^{jk\Omega_0 n}$ the output of the LTI system is simply $H(k\Omega_0) e^{jk\Omega_0 n}$ ok

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DFS of $x(n)$ N_0
 Fundamental Freq.

Input = $e^{jk\Omega_0 n}$
 Output = $H(k\Omega_0) e^{jk\Omega_0 n}$
 \Rightarrow $c_k \cdot e^{jk\Omega_0 n}$ $\xrightarrow{\text{output}}$ $c_k H(k\Omega_0) e^{jk\Omega_0 n}$

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. So, for each $e^{jk\Omega_0 n}$ raised to input [noise] $k\Omega_0$ naught n ok now the corresponding output [noise] we have seen [noise] equals $H(k\Omega_0)$ of $k\Omega_0$ [noise] naught [noise] times $e^{jk\Omega_0 n}$ raised to j [vocalized-noise] $k\Omega_0$ naught n which implies for input of c_k [noise] times $e^{jk\Omega_0 n}$ gives the corresponding output $c_k H(k\Omega_0)$. So, the input is scaled by linearity output is also scaled [noise] i am sorry this has to be [noise] $k\Omega_0$ naught n [noise] [vocalized-noise] and therefore, if input is $x(n)$ [noise] equals summation [noise]

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$$x(n) = \sum_{k=0}^{N_0-1} c_k e^{jk\Omega_0 n}$$

FROM LINEARITY:

$$\text{output} = \sum_{k=0}^{N_0-1} c_k H(k\Omega_0) e^{jk\Omega_0 n}$$

output for periodic signal $x(n)$.

Period = $\frac{N_0 \cdot 2\pi}{\Omega_0}$.

k equal to zero to n naught n minus k equal to zero to n naught minus one [noise]

C k e raised to j k omega naught n the output by linearity [noise] from linearity [noise] the output is simply [noise] the output is [noise] [vocalized-noise] k equal to zero to n naught minus one c [noise] k h of k omega naught e raised to j k omega naught n ok. So, this is the corresponding output ok. So, this is your output to the [noise] periodic [noise] signal x n ok this is the output for periodic signal x n with period n naught [noise] equals two pi by omega naught all right you can see the if input is periodic the output is also ah the output is also basically periodic ok [vocalized-noise] all right [vocalized-noise] and finally, for l t i systems characterized by difference equations

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LTI SYSTEMS CHARACTERIZED
BY DIFFERENCE EQUATIONS.

$$\sum_{k=0}^N a_k y(n-k) = \sum_{k=0}^M b_k x(n-k).$$

DIFFERENCE EQUATION

let us now look at similar to what we have seen before [noise] discrete time l t i systems [noise] [noise] characterized [noise] characterized by [noise] difference equations ok we have summation k equal to zero to n minus one lets say the difference equation is summation k equal to zero to n minus one y of n minus k equals summation k equal to zero ah sorry k equal to zero to n equals summation k equal to zero to m [noise] [noise] b k x n minus k this is the difference equation ok [noise]

This is the reference equation now taking the d t f t

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The image shows a screenshot of a software application window with a toolbar at the top. The main content area contains handwritten mathematical work. At the top, there are three horizontal lines with arrows pointing to the left, labeled $k=0$, $k=0$, and $k=0$ respectively. Below these is a horizontal line with the text "DIFFERENCE EQUATION" written in purple and underlined. The main equation is written in purple ink:

$$\sum_{k=0}^N a_k Y(\Omega) e^{-jk\Omega} = \sum_{k=0}^M b_k X(\Omega) e^{-jk\Omega}$$

In the bottom right corner of the application window, there is a status bar showing "131 / 133" and "131".

what we have a summation $\sum_{k=0}^{N-1} a_k y(\Omega) e^{-jk\Omega}$ that must be equal to summation $\sum_{k=0}^M b_k x(\Omega) e^{-jk\Omega}$ this means now taking the [noise]

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$$\sum_{k=0}^N a_k Y(\Omega) e^{-jk\Omega} = \sum_{k=0}^M b_k X(\Omega) e^{-jk\Omega}$$

$$Y(\Omega) \sum_{k=0}^N a_k e^{-jk\Omega} = X(\Omega) \sum_{k=0}^M b_k e^{-jk\Omega}$$

of ω common on the left i have y of ω summation k equal to zero to n a k e raised to minus $j k \omega$ equals

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$$H(\Omega) = \frac{Y(\Omega)}{X(\Omega)} = \frac{\sum_{k=0}^M b_k e^{-jk\Omega}}{\sum_{k=0}^N a_k e^{jk\Omega}}$$

Taking inverse DFT gives impulse response.

taking x of ω common on the right $x \omega^k$ equal to zero to m b_k [noise] e raised to minus $k \omega$ which means h of ω equals $y \omega$ by $x \omega$ [noise] equal summation k equal to zero to m [noise] $b_k e$ raised to minus $j \omega$ divided by summation k equal to zero to n a_k [noise] e raised to minus $j k \omega$ ok.

So, that is the transfer function [vocalized-noise] and taking the inverse d t f t of this gives the impulse response ok [noise] taking the inverse d t f t of this gives the impulse response [vocalized-noise] all right. So, basically that completes the analysis of the application on the properties of the d t f t that is the discrete time fourier transform for l t i systems with respect to the transfer function the output output for an periodic signal output for an a periodic signal and also ah how to derive the impulse response and the transfer function of a d t of an l t i system [vocalized-noise] described by a constant coefficient difference equation alright. So, well stop here and continue in the subsequent module.

Thank you very much [noise]